

Automotive IPD Series

Built-in current sensing function 1ch High Side Switch

BV1HB045EFJ-C

General Description

The BV1HB045EFJ-C is a 1ch high-side switch for automotive application. It has a built-in overcurrent limit function, thermal shutdown protection function, open load detection function, low power output-OFF function. It has a current sensing function of output load current.

Key Specifications

■ Power Supply Operating Range	6 V to 28 V
■ ON-Resistance ($T_j = 25^\circ\text{C}$)	45 m Ω (Typ)
■ Overcurrent Limit	21.0 A (Min)
■ Standby Current ($T_j = 25^\circ\text{C}$)	0.5 μA (Max)
■ Active Clamp Tolerance ($T_j = 25^\circ\text{C}$)	130 mJ

Features

- Built-in current sensing function
- Built-in Dual TSD (Note 1)
- AEC-Q100 Qualified (Note 2)
- Built-in Overcurrent Protection Function (OCP)
- Built-in thermal shutdown protection function (TSD)
- Built-in open load detection function
- Built-in Low-Voltage Output-OFF Function (UVLO)
- Built-in diagnostic output
- Low On-Resistance Single Nch MOSFET Switch
- Monolithic power management IC with control unit (CMOS) and power MOSFET mounted on a single chip

(Note 1) Two type of built-in temperature protection:

Junction temperature, and ΔT_j protection that detects sudden temperature rise of the Power-MOS

(Note 2) Grade 1

Package
HTSOP-J8

W (Typ) x D (Typ) x H (Max)
4.9 mm x 6.0 mm x 1.0 mm



Application

- Resistance load, inductance and capacitance load

Typical Application Circuit

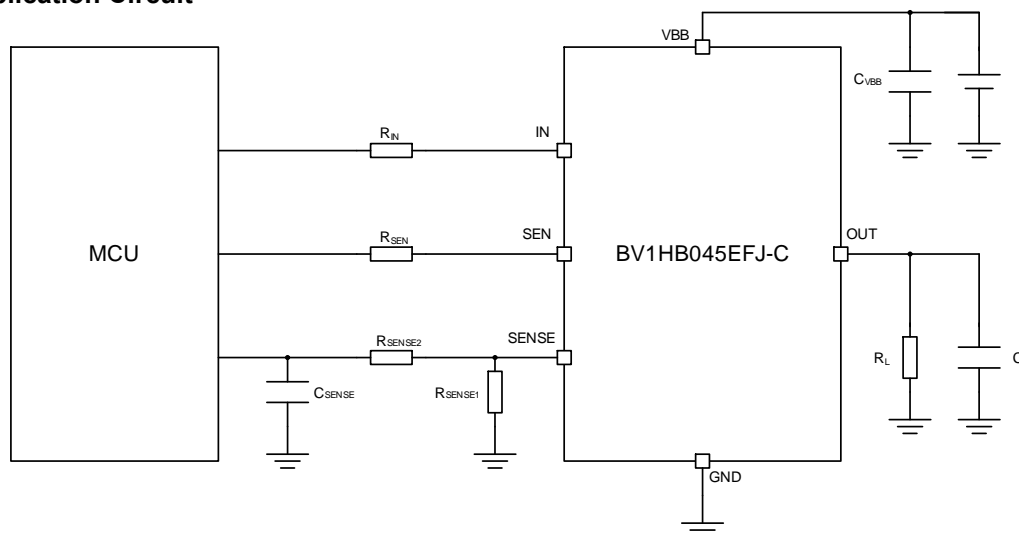


Figure 1. Application Circuit

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Pin Configuration

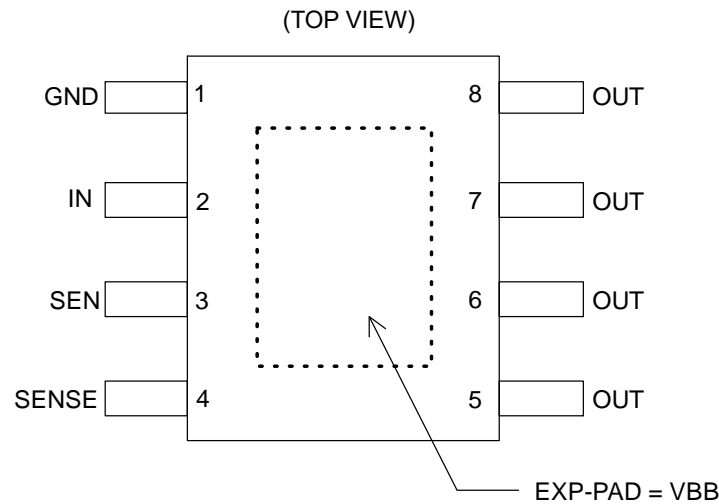


Figure 2. Pin Configuration

Pin Description

Pin No.	Pin Name	Function
1	GND	Ground pin
2	IN	Input pin. Pull-down resistor is connected internally. Active High to turn on the switch.
3	SEN	Current Sense and Diagnostic Function Enable Terminal.
4	SENSE	Current Sense output terminal.
5	OUT	Switch output pin
6	OUT	Switch output pin
7	OUT	Switch output pin
8	OUT	Switch output pin
EXP-PAD	VBB	Power input pin, switch input pin

Block Diagram

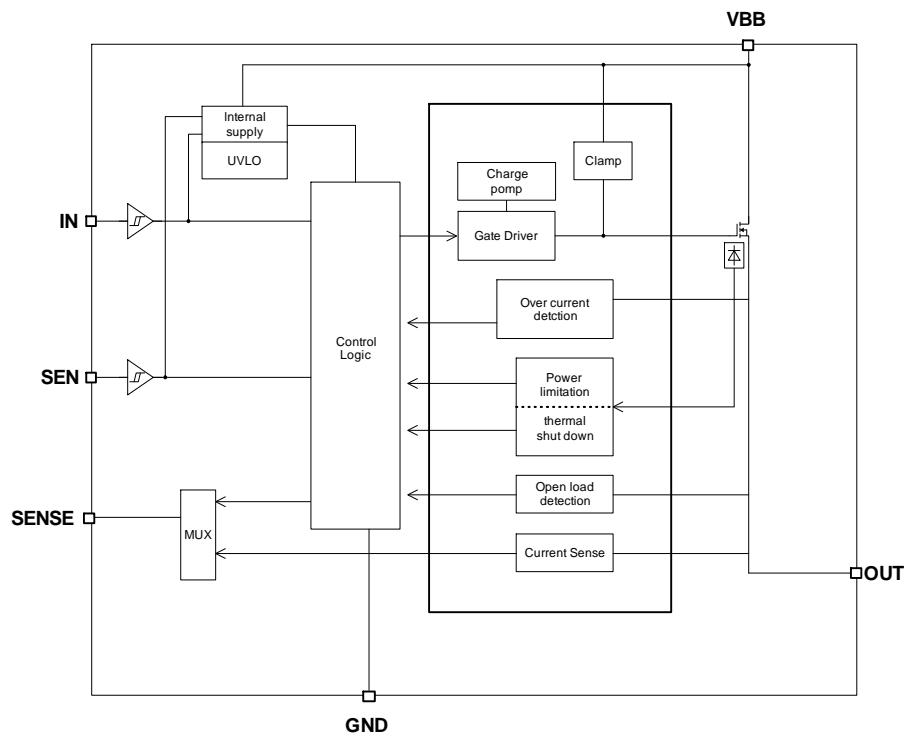


Figure 3. Block Diagram

Definition

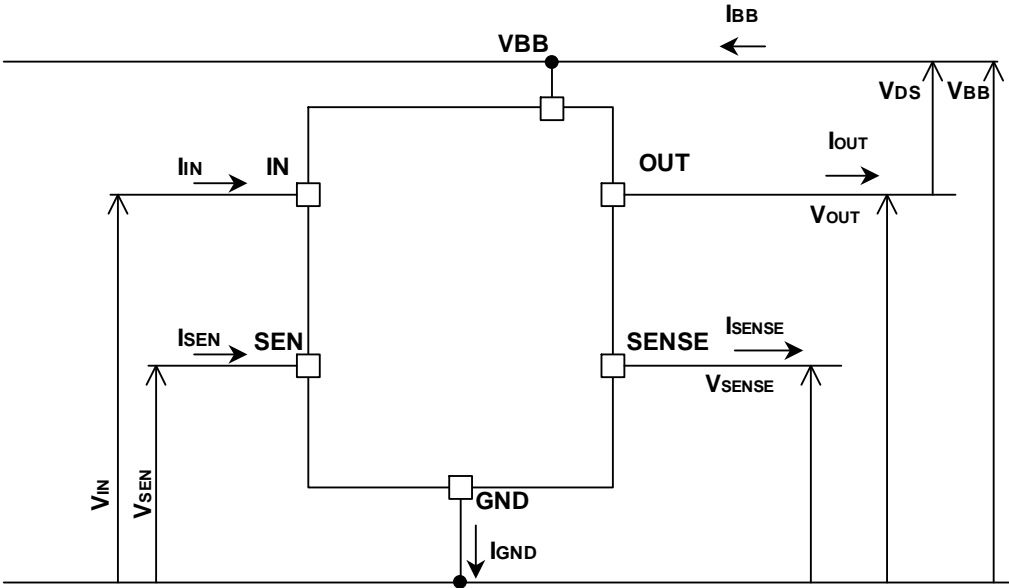


Figure 4. Voltage and Current Definition

Absolute Maximum Ratings (Ta = 25 °C)

Parameter	Symbol	Rating	Unit
V _{BB} - OUT Voltage	V _{DS}	-0.3 to Internal clamp ^(Note 1)	V
Power Supply Voltage	V _{BB}	-0.3 to +40	V
Input Voltage	V _{IN} , V _{SEN}	-0.3 to +7.0	V
Diagnostic Output Voltage	V _{SENSE}	-0.3 to +7.0	V
Output Current	I _{OUT}	Internal limit ^(Note 2)	A
Diagnostic Output Current	I _{SENSE}	20	mA
Junction Temperature Width	T _J	-40 to +150	°C
Storage Temperature Range	T _{stg}	-55 to +150	°C
Maximum Junction Temperature	T _{Jmax}	+150	°C
Active Clamp Energy (Single Pulse) T _{J(START)} = 25 °C, I _{OUT} = 2 A ^{(Note 3)(Note 4)}	E _{AS} (25 °C)	130	mJ
Active Clamp Energy (Single Pulse) T _{J(START)} = 150 °C, I _{OUT} = 2 A ^{(Note 3)(Note 4)}	E _{AS} (150 °C)	70	mJ
Supply Voltage for Short Circuit Detection ^(Note 5)	V _{BBLIM}	28	V

(Note 1) Internally limited by output clamp voltage.

(Note 2) Internally limited by fixed over current limit.

(Note 3) Maximum active clamp energy using single pulse of I_{OUT(START)} = 2 A and V_{BB} = 14 V.

(Note 4) Not 100% tested.

(Note 5) Maximum power supply voltage that can detect short circuit protection.

Caution 1: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Caution 2: Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, design a PCB with thermal resistance taken into consideration by increasing board size and copper area so as not to exceed the maximum junction temperature rating.

Caution 3: When IC turns off with an inductive load, reverse energy is generated. This energy can be calculated by the following equation:

$$E_L = \frac{1}{2} \times L \times I_{OUT(START)}^2 \times \left(1 - \frac{V_{BB}}{V_{BB} - V_{DS}}\right)$$

Where:

L is the inductance of the inductive load.

I_{OUT(START)} is the output current at the time of turning off.

The BV1HB045EFJ-C integrates the active clamp function to internally absorb the reverse energy E_L which is generated when the inductive load is turned off. When the active clamp operates, the thermal shutdown function does not work. Decide a load so that the reverse energy E_L is active clamp energy E_{AS} (refer to Figure 5. Active Clamp Energy vs Output Current) or under when inductive load is used.

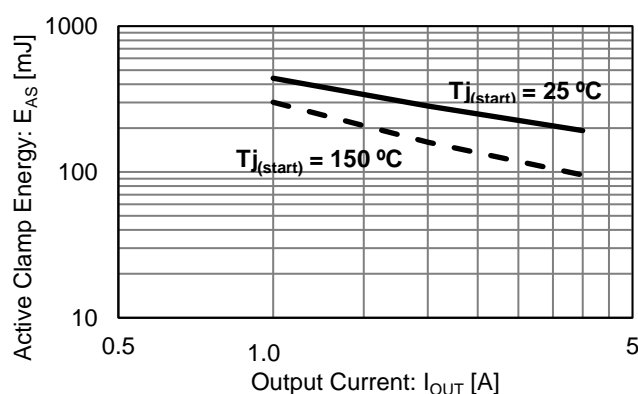


Figure 5. Active Clamp Energy vs Output Current

Recommended Operating Conditions

Parameter	Symbol	Min	Typ	Max	Unit
Power Supply Voltage Operating Range	V _{BB}	6	14	28	V
Operating Temperature	T _{opr}	-40	-	+150	°C
Input Frequency	f _{IN}	-	-	1	kHz

Thermal Resistance (Note 1)

Parameter	Symbol	Typ	Unit	Condition
HTSOP-J8				
Between Junction and Surroundings Temperature Thermal Resistance	θ_{JA}	130.3	°C/W	1s (Note 2)
		36.8	°C/W	2s (Note 3)
		25.9	°C/W	2s2p (Note 4)
Between Junction and the top center of the outside surface of the component package Thermal Characterization Parameter (Note 5)	Ψ_{JT}	20	°C/W	1s (Note 2)
		8	°C/W	2s (Note 3)
		6	°C/W	2s2p (Note 4)

(Note 1) The thermal impedance is based on JESD51-2A (Still-Air) standard. It is used the chip of BV1HB045EFJ-C.

(Note 2) JESD51-3 standard FR4 114.3 mm x 76.2 mm x 1.57 mm 1-layer (1s)

(Top copper foil: ROHM recommended Footprint + wiring to measure, 2 oz. copper.)

(Note 3) JESD51-5 standard FR4 114.3 mm x 76.2 mm x 1.60 mm 2-layers (2s)

(Top copper foil: ROHM recommended Footprint + wiring to measure/

Copper foil area on the reverse side of PCB: 74.2 mm x 74.2 mm,

copper (top & reverse side) 2 oz)

(Note 4) JESD51-5/-7 standard FR4 114.3 mm x 76.2 mm x 1.60 mm 4-layers (2s2p)

(Top copper foil: ROHM recommended Footprint + wiring to measure/

2 inner layers and copper foil area on the reverse side of PCB: 74.2 mm x 74.2 mm,

copper (top & reverse side/inner layers) 2 oz/1 oz)

(Note 5) The thermal characterization parameter to report the difference between junction temperature and the temperature at the top center of the outside surface of the component package.

■ PCB Layout 1 layer (1s)

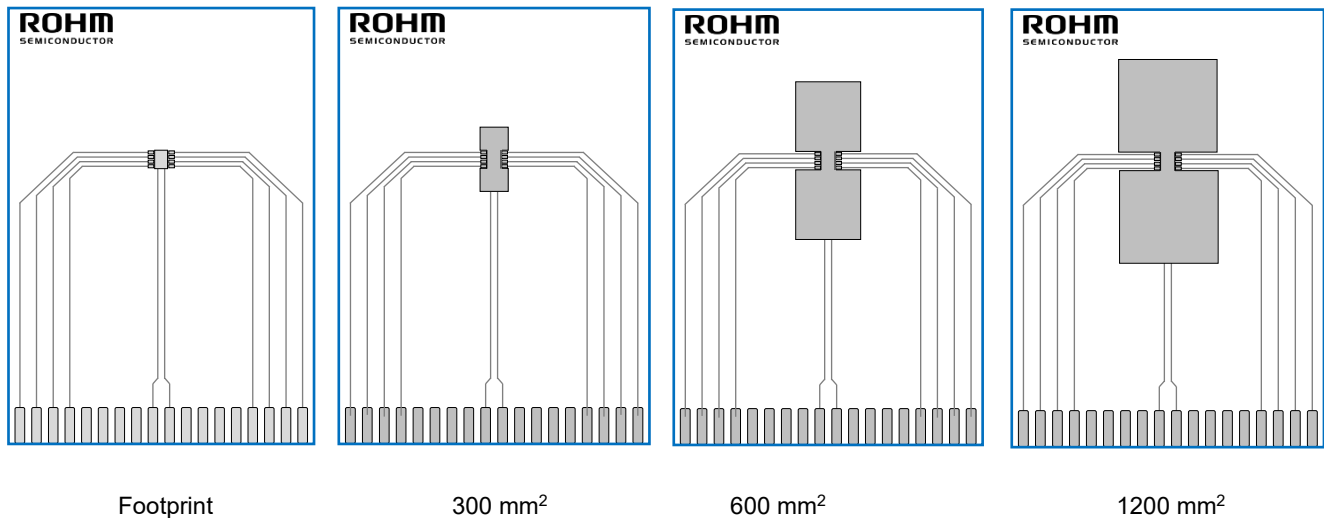


Figure 6. PCB Layout 1 Layer (1s)

Dimension	Value
Board Finish Thickness	1.57 mm ± 10 %
Board Dimension	76.2 mm x 114.3 mm
Board Material	FR4
Copper Thickness (Top Layer)	0.070 mm (Cu: 2 oz)
Copper Foil Area Dimension	Footprint/100 mm ² /600 mm ² /1200 mm ²

Thermal Resistance – continued

- PCB Layout 2 layers (2s)

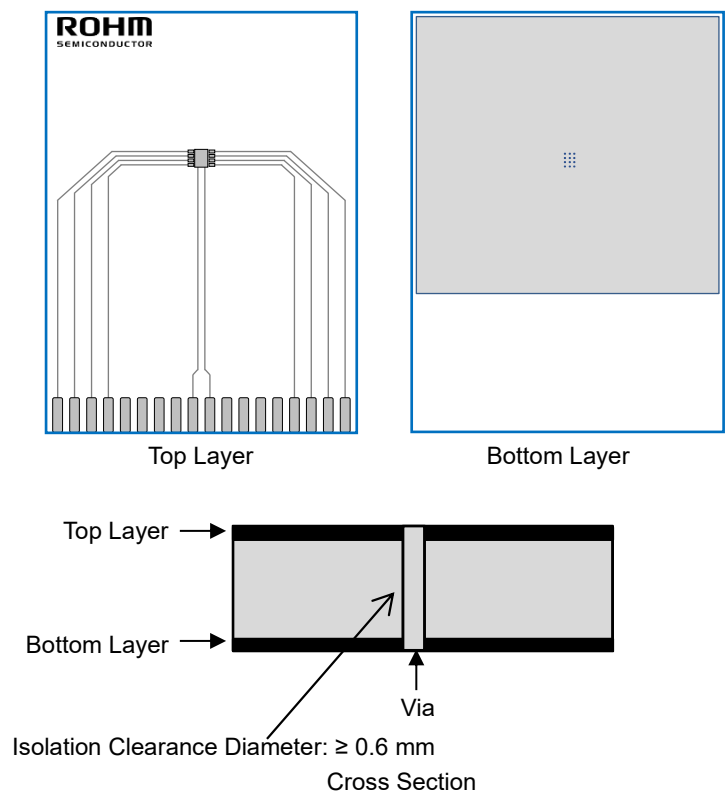


Figure 7. PCB Layout 2 Layers (2s)

Dimension	Value
Board Finish Thickness	1.60 mm ± 10 %
Board Dimension	76.2 mm x 114.3 mm
Board Material	FR4
Copper Thickness (Top/Bottom Layers)	0.070 mm (Cu + Plating)
Thermal Vias Separation/Diameter	1.2 mm/0.3 mm

Thermal Resistance – continued

- PCB Layout 4 layers (2s2p)

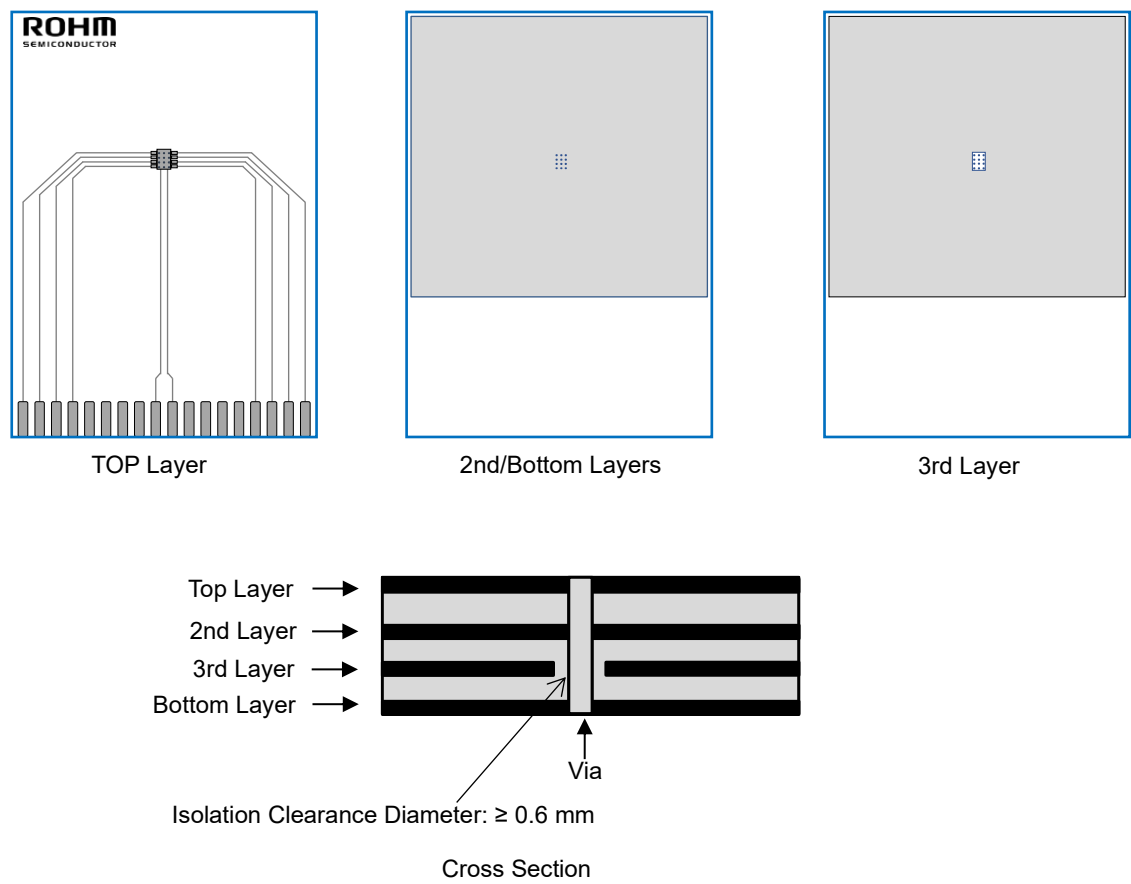


Figure 8. PCB Layout 4 Layers (2s2p)

Dimension	Value
Board Finish Thickness	1.60 mm \pm 10 %
Board Dimension	76.2 mm x 114.3 mm
Board Material	FR4
Copper Thickness (Top/Bottom Layers)	0.070 mm (Cu + Plating)
Copper Thickness (Inner Layers)	0.035 mm
Thermal Vias Separation/Diameter	1.2 mm/0.3 mm

Thermal Resistance – continued

- Transient Thermal Resistance (Single Pulse)

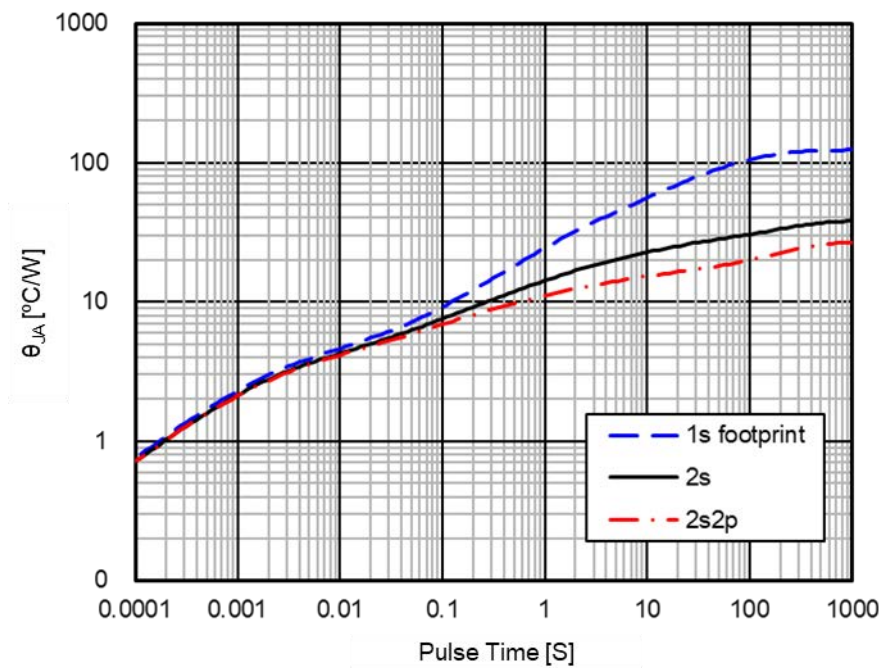


Figure 9. Transient Thermal Resistance

- Thermal Resistance (θ_{JA} vs Copper foil area- 1s)

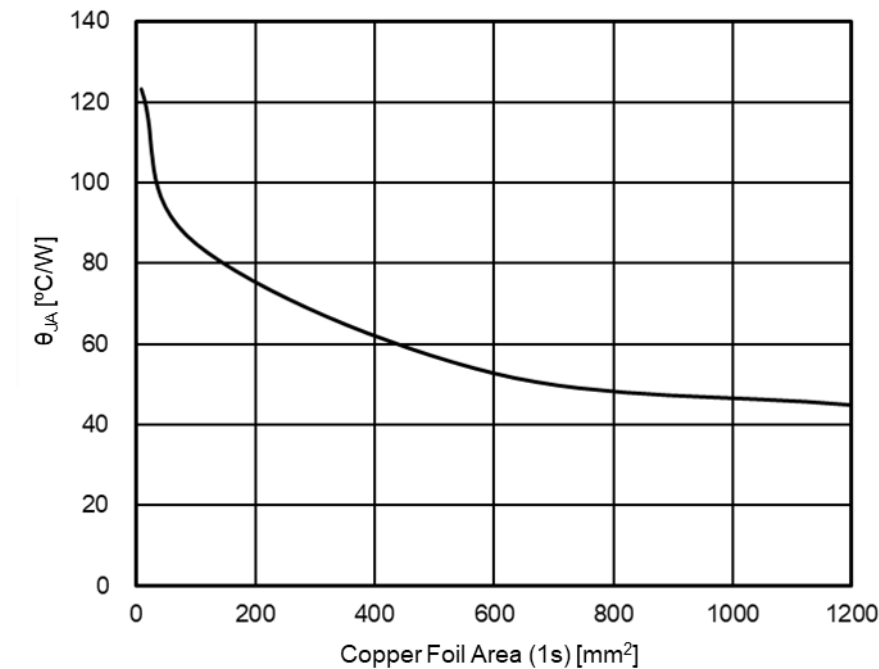


Figure 10. Thermal Resistance

Electrical Characteristics (Unless otherwise specified $V_{BB} = 6\text{ V}$ to 28 V , $T_j = -40\text{ }^{\circ}\text{C}$ to $+150\text{ }^{\circ}\text{C}$)

Parameter	Symbol	Limit			Unit	Condition
		Min	Typ	Max		
Power Supply						
Standby current	I _{BBL1}	-	-	0.5	μA	V _{BB} = 14 V, V _{IN} = V _{SEN} = 0 V, V _{OUT} = 0 V, T _j = 25 °C
		-	-	10	μA	V _{BB} = 14 V, V _{IN} = V _{SEN} = 0 V, V _{OUT} = 0 V, T _j = 150 °C
	I _{BBL2}		1.0		mA	V _{BB} = 14 V, V _{IN} = 0 V, V _{SEN} = 5 V, V _{OUT} = 0 V, T _j = 25 °C
			1.2		mA	V _{BB} = 14 V, V _{IN} = 0 V, V _{SEN} = 5 V, V _{OUT} = 0 V, T _j = 150 °C
Operating Current	I _{BBH}	-	3	5	mA	V _{BB} = 14 V, V _{IN} = V _{SEN} = 5 V, V _{OUT} = open
UVLO Detection Voltage	V _{UVLO}	-	-	5	V	V _{IN} = 5 V, R _L = 10 kΩ R _L : Output Load Resistor
UVLO Hysteresis Voltage	V _{UVHYS}	-	-	0.9	V	
Input (V _{IN})						
High Level Input Voltage	V _{INH}	2.1	-	-	V	
Low Level Input Voltage	V _{INL}	-	-	0.9	V	
Input Hysteresis Voltage	V _{IN_HYS}	-	0.4	-	V	
High Level Input Current	I _{INH}	-	50	150	μA	V _{IN} = 5 V
Low Level Input Current	I _{INL}	-10	-	+10	μA	V _{IN} = 0 V
Input (V _{SEN})						
H-level input voltage	V _{SENH}	2.1	-	-	V	
L-level input voltage	V _{SENL}	-	-	0.9	V	
Input hysteresis	V _{SEN_HYS}	-	0.4	-	V	
H-level input current	I _{SENH}	-	50	150	μA	V _{IN} = 5 V
L-level input current	I _{SENL}	-10	-	+10	μA	V _{IN} = 0 V
Power MOS Output						
Output ON Resistance	R _{ON1}	-	45	60	mΩ	V _{BB} = 8 V ~ 28 V, T _j = 25 °C
	R _{ON2}	-	-	90	mΩ	V _{BB} = 8 V ~ 28 V, T _j = 150 °C
	R _{ON3}	-	-	75	mΩ	V _{BB} = 6 V, T _j = 25 °C
Output Leak Current	I _{OUTL1}	-	-	0.5	μA	V _{IN} = 0 V, V _{OUT} = 0 V, T _j = 25 °C
	I _{OUTL2}	-	-	10	μA	V _{IN} = 0 V, V _{OUT} = 0 V, T _j = 150 °C
Output Slew Rate	SR _{ON}	-	0.3	1.0	V/μs	V _{BB} = 14 V, R _L = 6.5 Ω
	SR _{OFF}	-	0.3	1.0	V/μs	V _{BB} = 14 V, R _L = 6.5 Ω
Output voltage drop limitation at small load currents	V _{DS(SL)}	-	10	25	mV	I _{OUT} = 50 mA
Propagation Delay when ON	t _{OUTON}	-	90	140	μs	V _{BB} = 14 V, R _L = 6.5 Ω
Propagation Delay when OFF	t _{OUTOFF}	-	40	100	μs	V _{BB} = 14 V, R _L = 6.5 Ω
Output Clamp Voltage	V _{DSCLP}	45	50	55	V	V _{IN} = 0 V, I _{OUT} = 10 mA

Electrical Characteristics (Unless otherwise specified $V_{BB} = 6V$ to $28V$, $T_j = -40^{\circ}C$ to $150^{\circ}C$)

Parameter	Symbol	Limit			Unit	Condition
		Min	Typ	Max		
Current sensing unit						
Current Sense Ratio 1	K ₁	-50%	1500	+50%	-	V _{IN} = V _{SEN} = 5 V, I _{out} = 50mA
Current Sense Ratio 2	K ₂	-30%	1450	+30%		V _{IN} = V _{SEN} = 5 V, I _{out} = 0.25 A
Current Sense Ratio 3	K ₃	-20%	1450	+20%	-	V _{IN} = V _{SEN} = 5 V, I _{out} = 0.5 A
Current Sense Ratio 4	K ₄	-10%	1450	+10%	-	V _{IN} = V _{SEN} = 5 V, I _{out} = 1 A
Current Sense Ratio 5	K ₅	-7%	1450	+7%	-	V _{IN} = V _{SEN} = 5 V, I _{out} = 2 A
Current Sense Ratio 6	K ₆	-5%	1450	+5%	-	V _{IN} = V _{SEN} = 5 V, I _{out} = 4 A
K _{ILIS} derating	ΔK _{ILIS}	-5%	-	+5%	%	K ₄ vs K ₅
SENSE terminal leakage current	I _{SENSEL}	-	-	0.5	μA	V _{SEN} = 0 V, V _{SENSE} = 0 V
Output voltage of SENSE terminal in abnormal condition	V _{SENSEH}	4.0	5.5	6.5	V	V _{BB} = 8 V to 28 V, R _{SENSE} = 1 kΩ
Diagnostic output delay time when input (IN) is ON	t _{INON}	-	130	300	μs	V _{BB} = 14 V, R _L = 6.5 Ω, T _j = 25 °C
Diagnostic output delay time when input (IN) is off	t _{INOFF}	-	40	100	μs	V _{BB} = 14 V, R _L = 6.5 Ω, T _j = 25 °C
Diagnostic output delay time when input (SEN) is ON	t _{SENON}	-	10	50	μs	V _{BB} = 14 V, R _L = 6.5 Ω, T _j = 25 °C
Diagnostic output delay time when input (SEN) is off	t _{SENOFF}	-	10	50	μs	V _{BB} = 14 V, R _L = 6.5 Ω, T _j = 25 °C
SENSE Settling Time after Load Change	t _{SENON(CL)}	-	-	20	μs	R _{SENSE} = 1 kΩ, I _{out} = 1 A to 2 A
Protection Circuit						
Overcurrent Limit Value	I _{LIMH}	21	30	40	A	V _{DS} = 5 V
Open Load Detection Voltage	V _{OLD}	V _{BB} -3.0	V _{BB} -2.0	V _{BB} -1.0	V	V _{BB} = 8 V to 28 V
Open Load Detection Source Current	I _{OLD}	-	10	30	μA	V _{IN} = 0 V, V _{OUT} = 5 V
Open Load Detection Diagnostic Output Mask Time	t _{OLD}	100	250	400	μs	V _{BB} = 14 V, V _{IN} = 5 to 0 V
Thermal Shutdown ^(Note 1)	T _{TSD}	150	175	200	°C	
Thermal Shutdown Hysteresis ^(Note 1)	T _{TSDHYS}	-	15	-	K	
ΔT _j Protection Temperature ^(Note 1)	T _{DTJ}	-	90	-	K	

(Note 1) Not 100% tested.

Typical Performance Curves

(Unless otherwise specified $V_{BB} = 14\text{ V}$, $V_{IN} = 5\text{ V}$, $T_j = 25\text{ }^{\circ}\text{C}$)

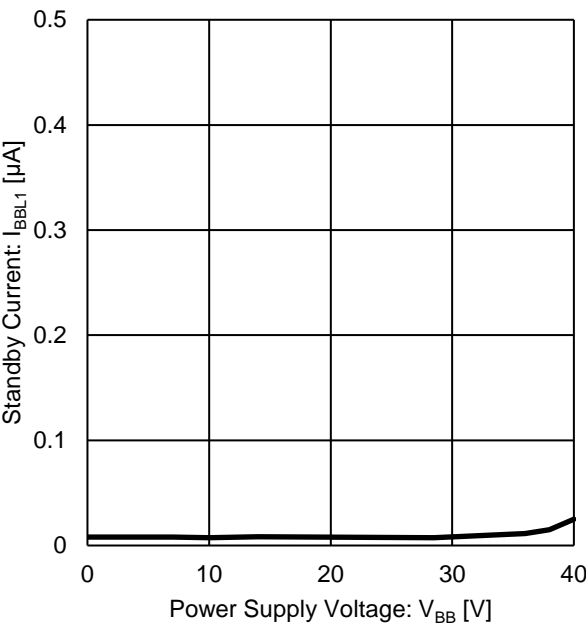


Figure 11. Standby Current vs Power Supply Voltage

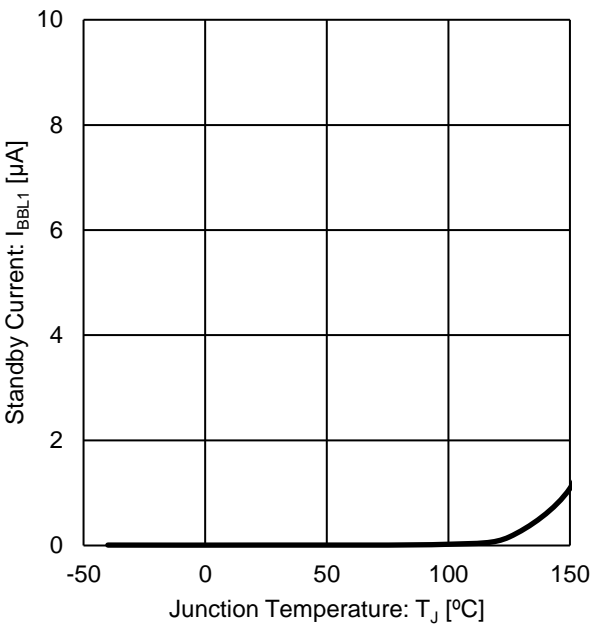


Figure 12. Standby Current vs Junction Temperature

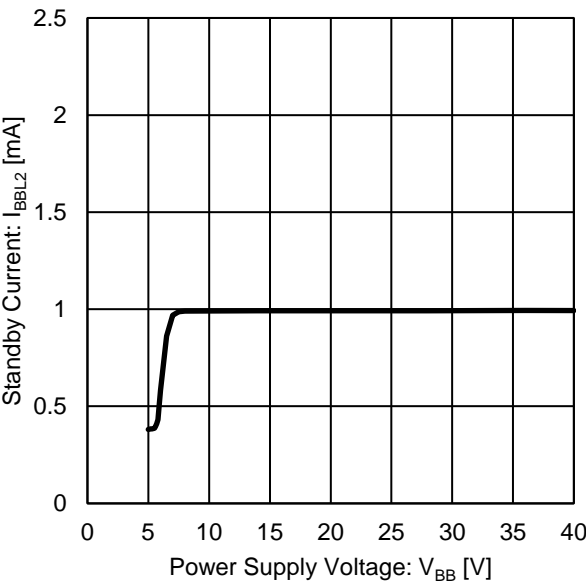


Figure 13. Standby Current vs Power Supply Voltage

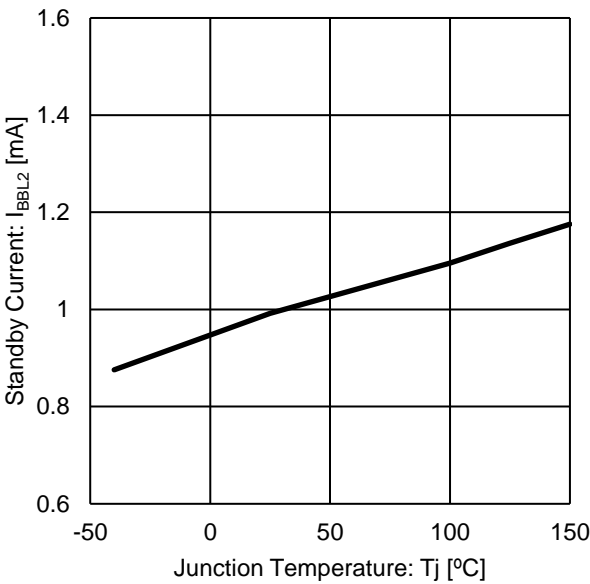


Figure 14. Standby Current vs Junction Temperature

Typical Performance Curves - continued

(Unless otherwise specified $V_{BB} = 14\text{ V}$, $V_{IN} = 5\text{ V}$, $T_j = 25\text{ }^{\circ}\text{C}$)

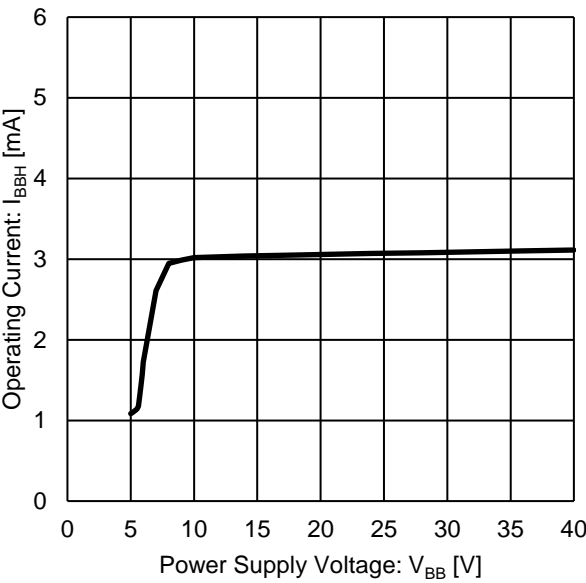


Figure 15. Circuit Current vs Power Supply Voltage

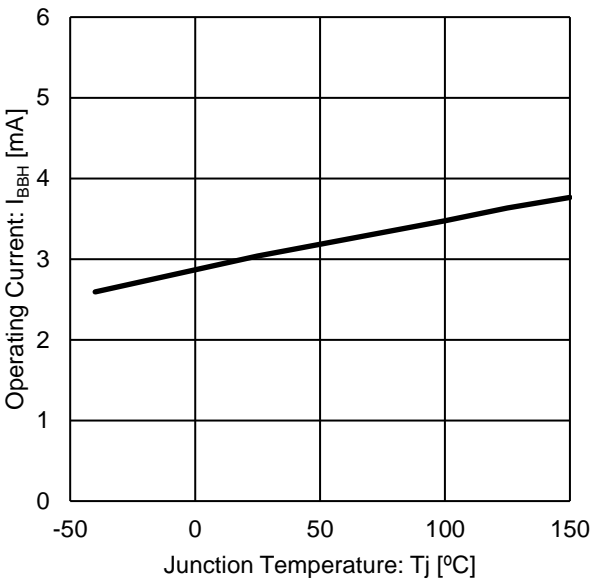


Figure 16. Circuit Current vs Junction Temperature

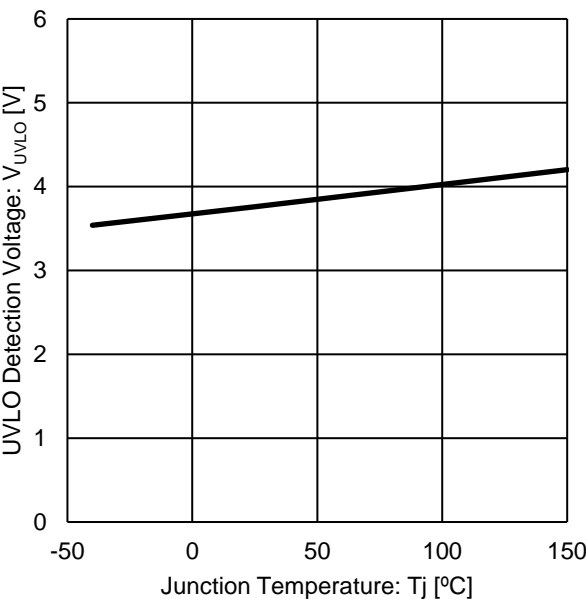


Figure 17. UVLO Detection Voltage vs Junction Temperature

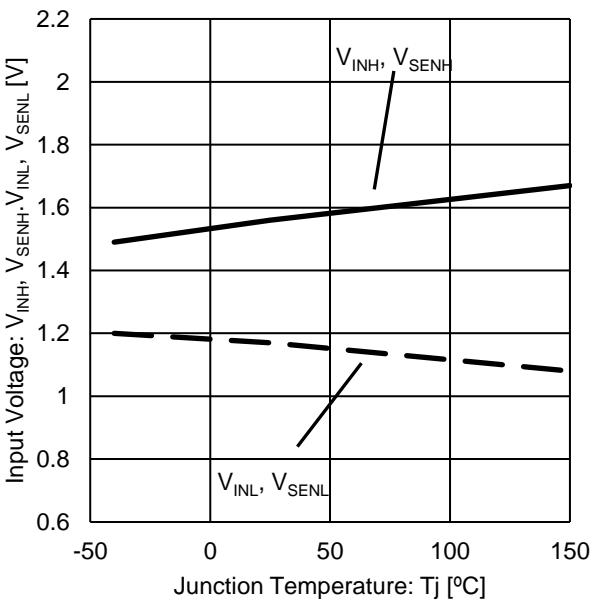


Figure 18. Input Voltage vs Junction Temperature

Typical Performance Curves - continued

(Unless otherwise specified $V_{BB} = 14\text{ V}$, $V_{IN} = 5\text{ V}$, $T_j = 25\text{ }^{\circ}\text{C}$)

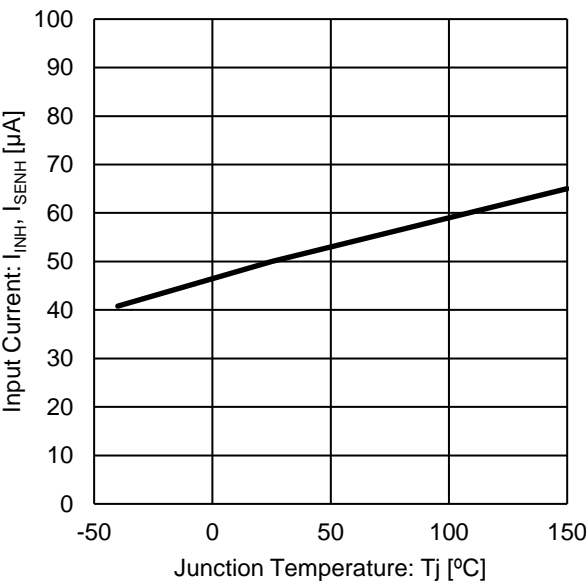


Figure 19. Input Current vs Junction Temperature

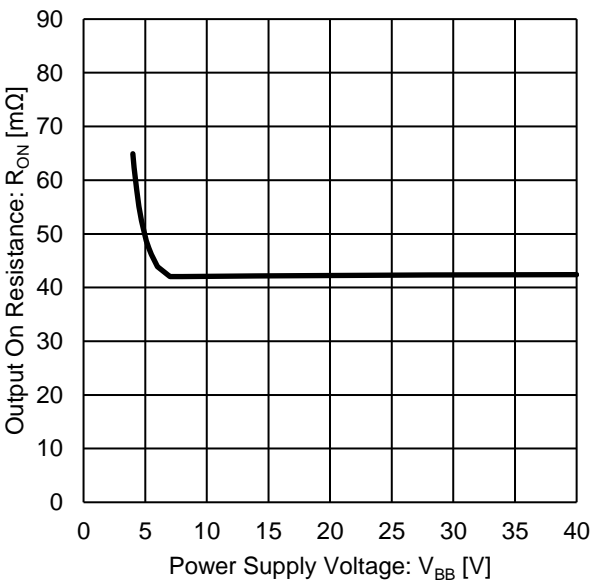


Figure 20. Output ON Resistance vs Power Supply Voltage

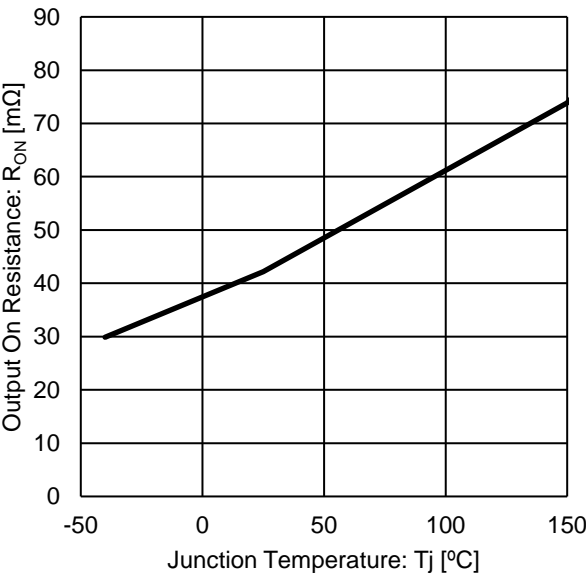


Figure 21. Output ON Resistance vs Junction Temperature

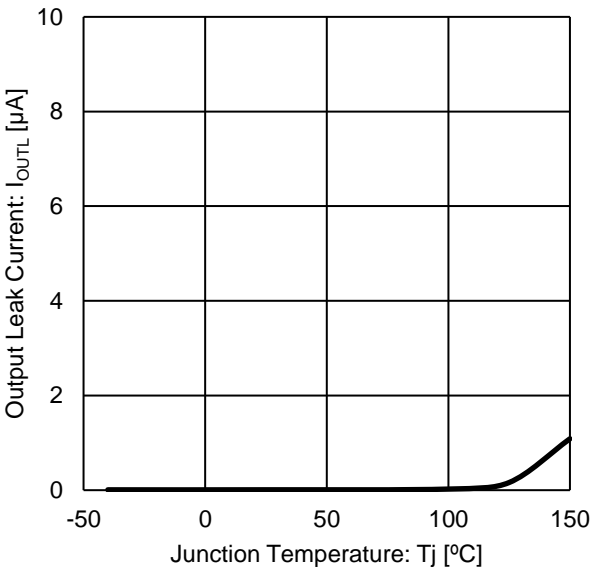


Figure 22. Output Leak Current vs Junction Temperature

Typical Performance Curves - continued

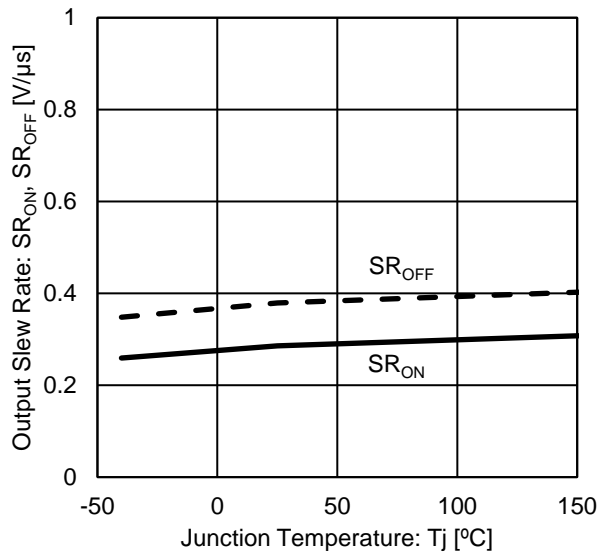
(Unless otherwise specified $V_{BB} = 14\text{ V}$, $V_{IN} = 5\text{ V}$, $T_j = 25\text{ }^{\circ}\text{C}$)

Figure 23. Output Slew Rate vs Junction Temperature

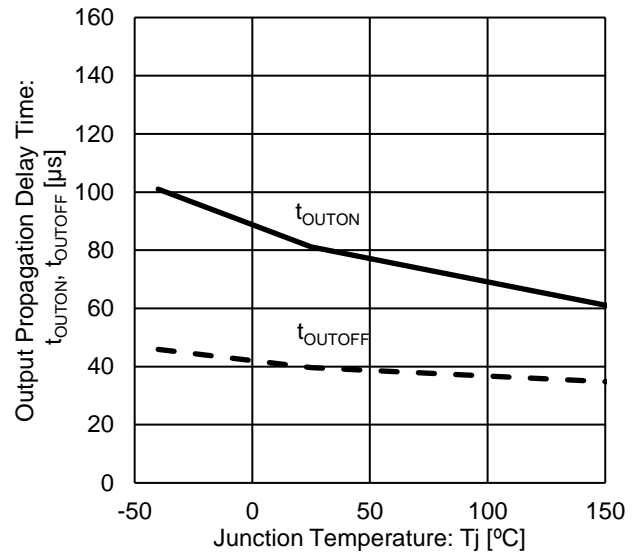


Figure 24. Output Propagation Delay Time vs Junction Temperature

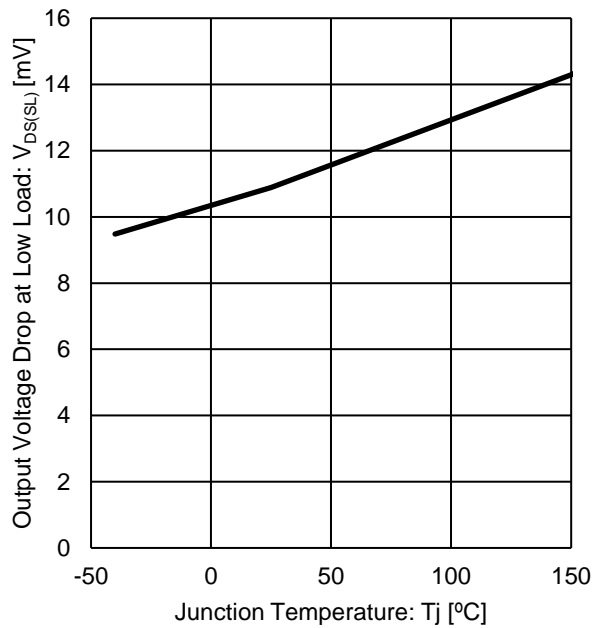


Figure 25. Output Voltage Drop at Low Load vs Junction Temperature

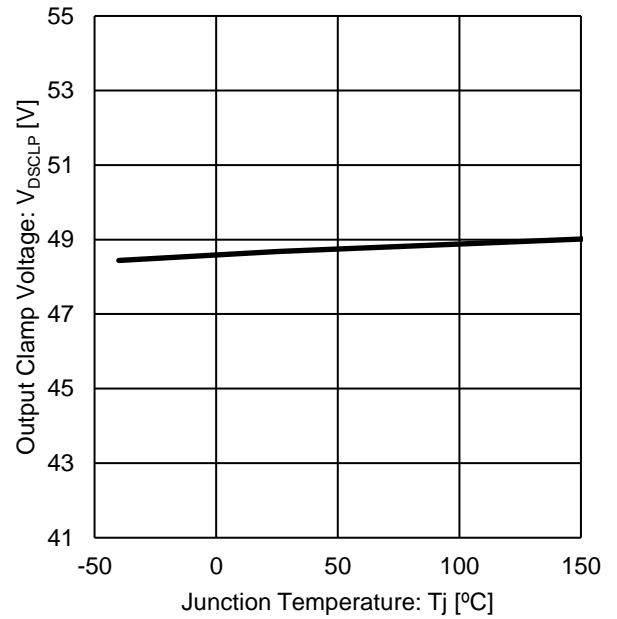


Figure 26. Output Clamp Voltage vs Junction Temperature

Typical Performance Curves - continued

(Unless otherwise specified $V_{BB} = 14\text{ V}$, $V_{IN} = 5\text{ V}$, $T_j = 25\text{ }^{\circ}\text{C}$)

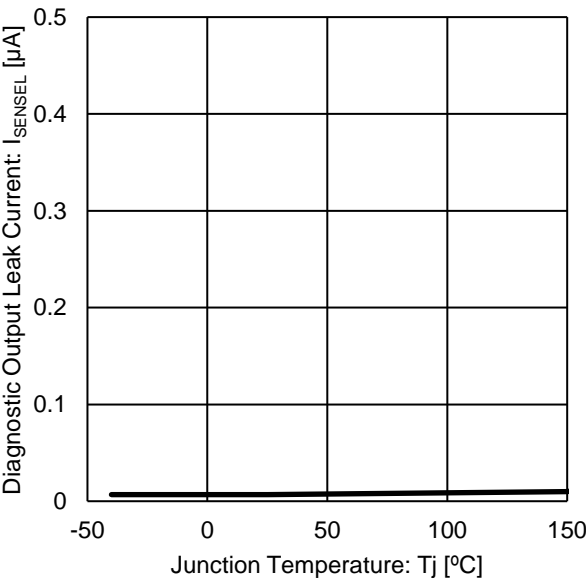


Figure 27. Diagnostic Output Leak Current vs Junction Temperature

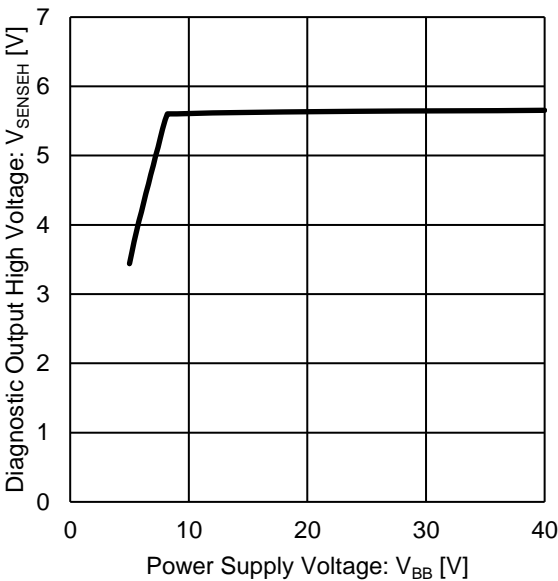


Figure 28. Diagnostic Output High Voltage vs Power Supply Voltage

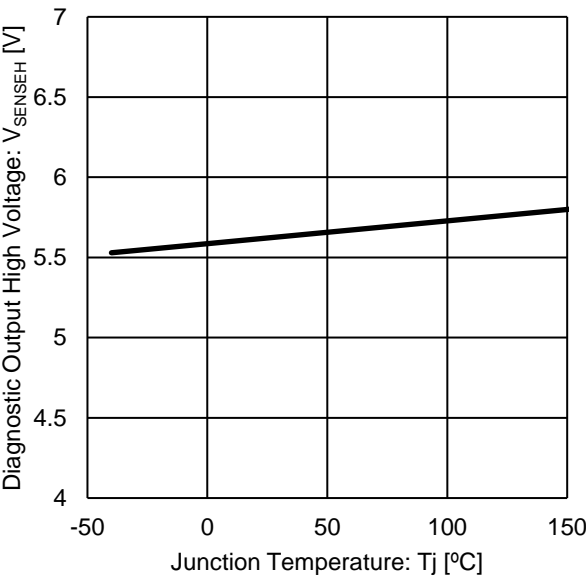


Figure 29. Diagnostic Output High Voltage vs Junction Temperature

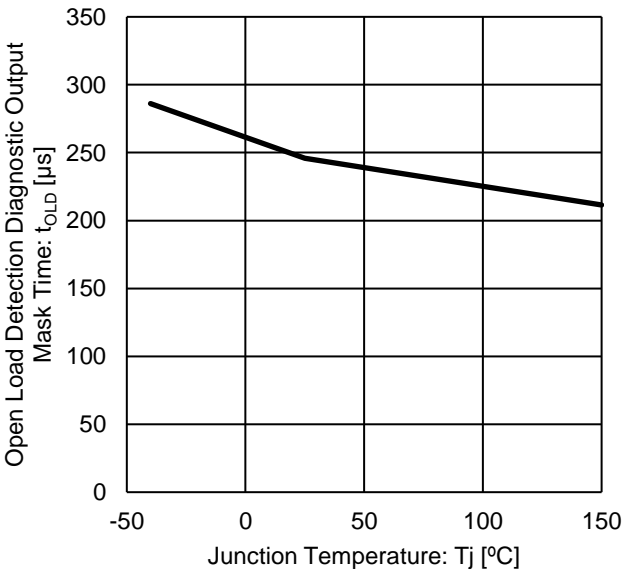


Figure 30. Open Load Detection Diagnostic Output Mask Time vs Junction Temperature

Typical Performance Curves - continued

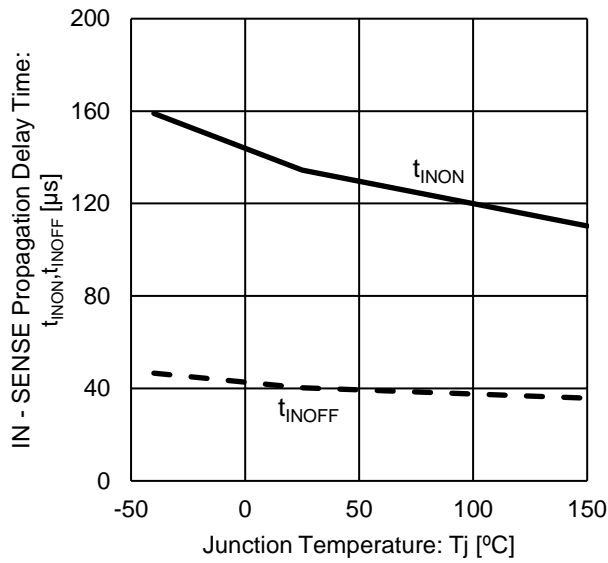
(Unless otherwise specified $V_{BB} = 14\text{ V}$, $V_{IN} = 5\text{ V}$, $T_j = 25\text{ }^{\circ}\text{C}$)

Figure 31. IN - SENSE Propagation Delay Time vs Junction Temperature

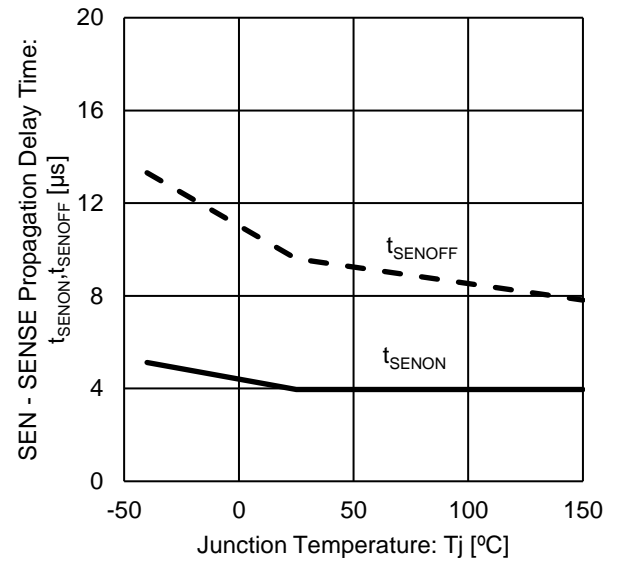


Figure 32. SEN - SENSE Propagation Delay Time vs Junction Temperature

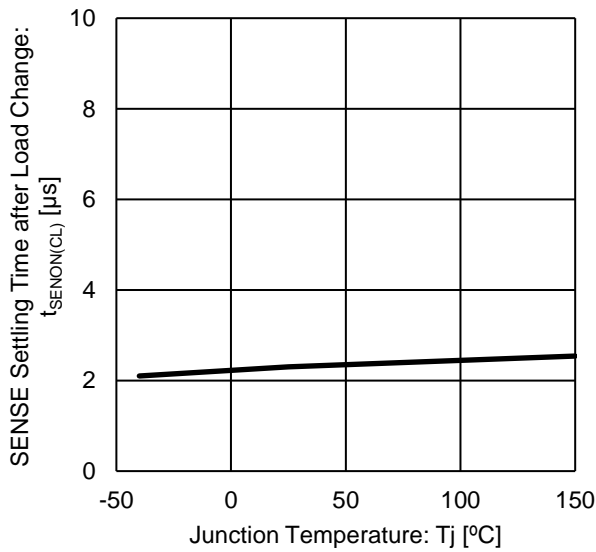


Figure 33. SENSE Settling Time after Load Change vs Junction Temperature

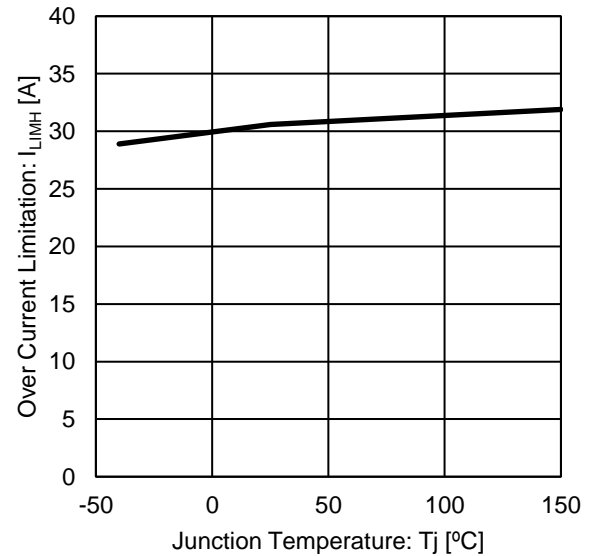


Figure 34. Over Current Limitation vs Junction Temperature

Measurement Circuit

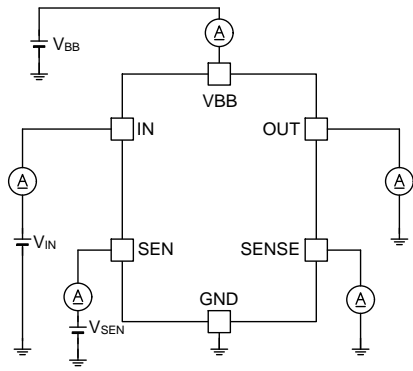


Figure 35. Standby Current
Low-Level Input (V_{IN}) Current
Low-Level Input (V_{SEN}) Current
Output Leak Current
Diagnostic Output Leak Current

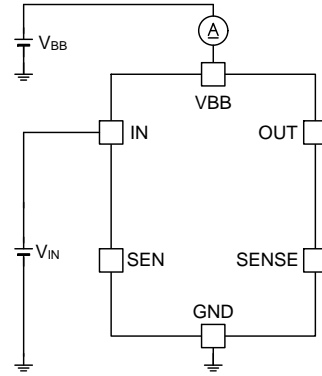


Figure 36. Operating Current

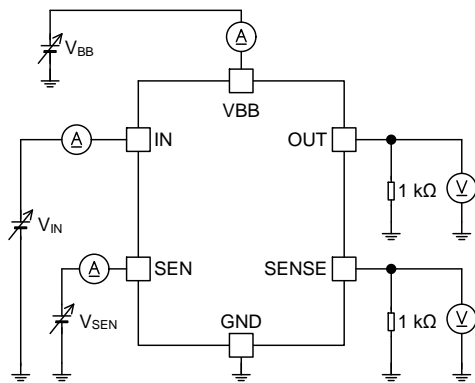


Figure 37. UVLO Detection Voltage
UVLO Hysteresis Voltage
High Level Input Voltage
Low Level Input Voltage
Input Hysteresis Voltage
High Level Input Current
Thermal Shutdown
Thermal Shutdown Hysteresis

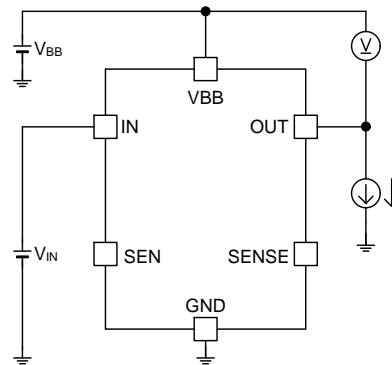


Figure 38. Output ON Resistance
Output Clamp Voltage

Measurement Circuit - continued

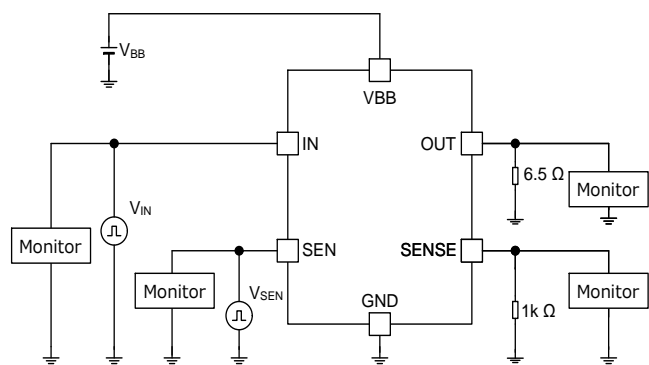


Figure 39. Output ON Slew Rate
Output OFF Slew Rate
Output ON Propagation Delay Time
Output OFF Propagation Delay Time
Diagnostic Output ON Propagation Delay Time
Diagnostic Output OFF Propagation Delay Time

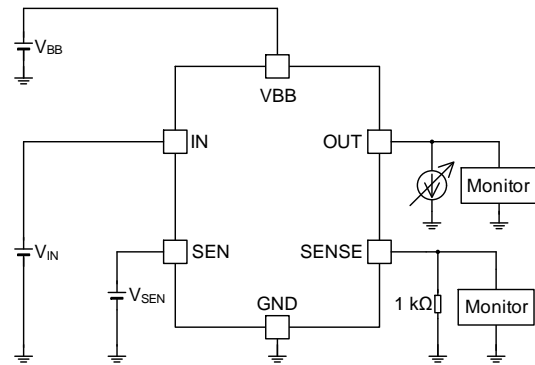


Figure 40. SENSE Settling Time
after Load Change

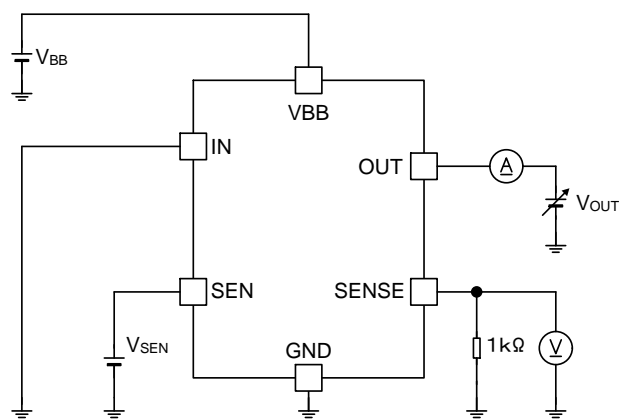


Figure 41. Open Load Detection Voltage
Open Load Detection Sink Current

Timing Chart

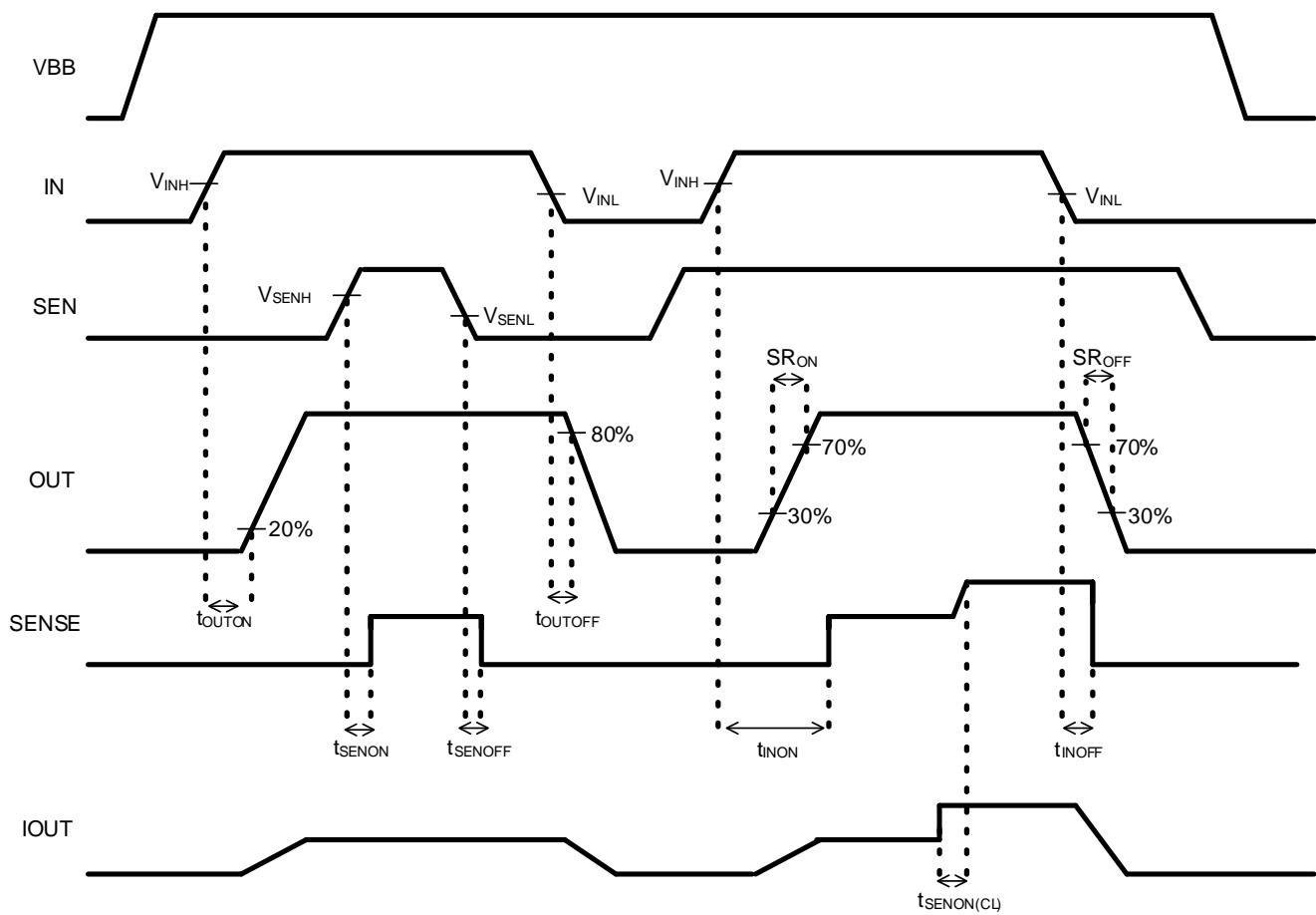


Figure 42. Timing Chart

Function Description

1. Protection Function

Table 1. Detection and Release Conditions of Each Protection Function and Diagnostic Output

Mode	Conditions	IN	SEN	SENSE	OUT
Standby	-	Low	Low	Low	Low
Operating	-	High	High	$I_{SENSE} = I_{OUT} / K$	High
Open Load Detect (OLD)	Detect $V_{OUT} > V_{BB} - 2.0 \text{ V (Typ)}$	Low	High	V_{SENSEH}	-
	Release $V_{OUT} < V_{BB} - 2.5 \text{ V (Typ)}$	Low	High	Hi-Z	-
Low Power Output-OFF (UVLO)	Detect $V_{BB} \leq 5.0 \text{ V (Max)}$	High	-	-	Low
	Release $V_{BB} \geq 5.9 \text{ V (Max)}$	High	-	-	High
Thermal Shutdown (TSD)	Detect $T_j > 175 \text{ }^{\circ}\text{C (Typ)}$	High	High	V_{SENSEH}	Low
	Release $T_j < 150 \text{ }^{\circ}\text{C (Typ)}$	High	High	$I_{SENSE} = I_{OUT} / K$	High
ΔT_j Protection (Note 2)	Detect $\Delta T_j > 90^{\circ}\text{C (Typ)}$	High	High	V_{SENSEH}	Low
	Release $\Delta T_j < 30 \text{ }^{\circ}\text{C (Typ)}$	High	High	$I_{SENSE} = I_{OUT} / K$	High
Over Current Protection (OCP)	Detect $I_{OUT} > I_{LIMH}$	High	High	V_{SENSEH}	High
	Release $I_{OUT} > I_{LIMH}$	High	High	$I_{SENSE} = I_{OUT} / K$	High

(Note1) Thermal shutdown is automatically restored to normal operation.

(Note2) Protect function by detecting PowerMOS sharp increase of temperature difference with control circuit.

This IC incorporates the above-mentioned protection-detection function, and outputs an abnormal condition at the SENSE terminal. Connect a resistor between the SENSE-GND and determine the abnormal condition based on the voltage level. It is self-rest and operation becomes normal when each protection releases after detecting.

Function Description - continued

2. Current sensing function

2.1 SENSE current

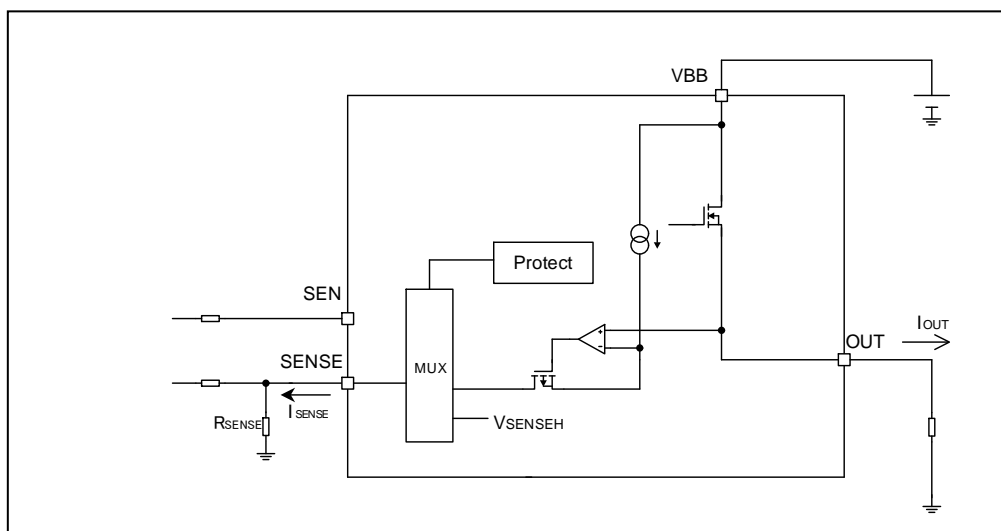


Figure 43. Current Sense Block Diagram

The SENSE terminal of the IC can feed back the current flowing through the IC.
 The SENSE voltage varies linearly according to the load current I_{OUT} during normal operation.
 The V_{SENSE} theoretical equations are shown below.

$$V_{SENSE} = R_{SENSE} \times I_{SENSE}$$

$$I_{SENSE} = \frac{I_{OUT}}{N}$$

$$V_{SENSE} = \frac{R_{SENSE} \times I_{OUT}}{N} = \frac{R_{SENSE} \times I_{OUT}}{1450 (typ)}$$

Where:

V_{SENSE} : SENSE terminal voltages

R_{SENSE} : SENSE resistor

I_{OUT} : Load current

N: Output mirror value

BV1HB045EFJ-C is recommended to use 1 kΩ as the pull-down resistor at SENSE pin.

When R_{SENSE} is 1 kΩ, and I_{OUT} is 2 A, the above formula is summarized as follows.

$$V_{SENSE} = \frac{1000 \times 2}{1450} = 1.379 [V]$$

Function Description - continued

2.2 Variation of Outputs Voltage of SENSE terminals

Diagnostic output current of I_{SENSE} increases linearly with I_{OUT} output current. Figure 44 shows the the variation of current sense ratio. The accuracy of the sense current depends on temperature and load current. To achieve high accuracy requirement, a calibration on the application is possible. To avoid multiple calibration points at different load and temperature conditions, BV1HB045EFJ-C allows limited derating of the k_{LIS} value, at a given point ($I_O = 1\text{ A}$, $T_j = 25\text{ °C}$). An external RC filter between SENSE pin and microcontroller ADC input pin is recommended to reduce signal ripple and oscillations.

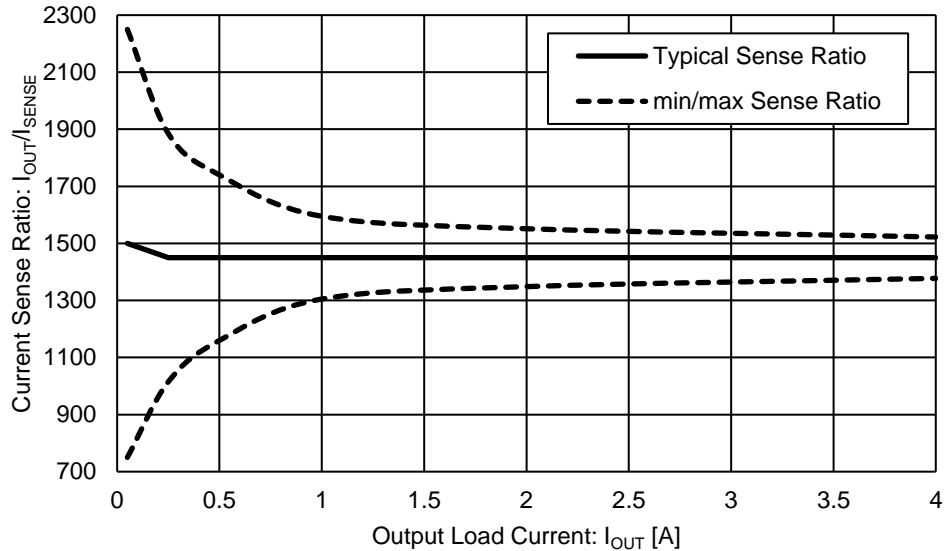


Figure 44. Current Sense Ratio vs Output Load Current

2.3 Outputs of SENSE terminals

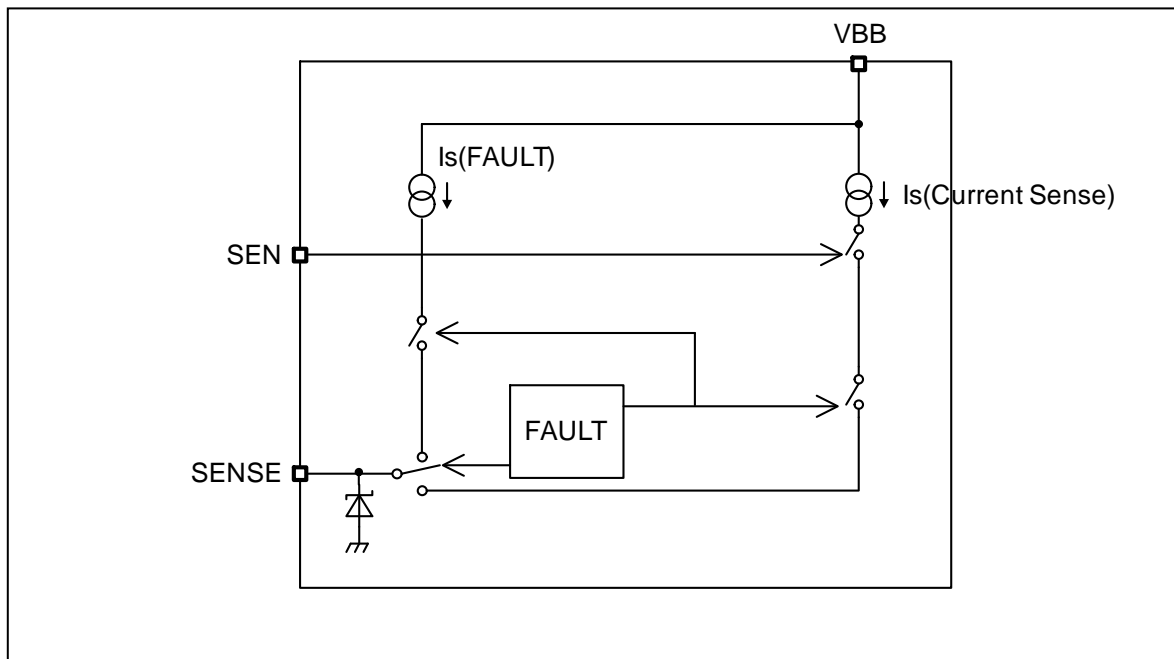


Figure 45. SENSE output-block diagram

The SENSE terminal serves as both the current sense output and the flag signal when an error is detected. When SEN = High, a current approximately 1/1450 of the output current is output to the SENSE terminal.

When overcurrent detection, overheat detection, or load open detection are activated, The FAULT signals of the Figure 45 output the V_{SENSEH} voltage generated internally from the SENSE terminal. When monitoring the V_{SENSEH} , operate within the recommended operating conditions. Refer to Table 1 for more information on SENSE outputs.

Function Description - continued

3. Overcurrent Protection

This IC has a built-in overcurrent protection function. When overcurrent flows in the output, the output current is limited to 30A (Typ) and self-diagnostic output (SENSE) becomes V_{SENSEH} .

4. Thermal Shutdown

4.1 Thermal Shutdown Protection

This IC has a built-in thermal shutdown protection function. When the IC chip temperature exceeds 175°C (Typ), the output is turned OFF and self-diagnostic output (SENSE) becomes V_{SENSEH} . When the temperature goes below 150°C (Typ), output will self-reset and operation becomes normal.

4.2 ΔT_j Protection

This IC has a built-in ΔT_j protection function that turns OFF the output when the temperature difference (T_{DTJ}) between the POWER-MOS unit ($T_{POWER-MOS}$) and the control unit (T_{AMB}) in the IC is 90°C (Typ) or more. ΔT_j protection also has a built-in hysteresis (T_{DTJHYS}) that returns the output to normal state when the temperature difference becomes 30°C (Typ) or less.

Figure 46 shows the timing chart of thermal shutdown protection and ΔT_j protection during output short to GND fault.

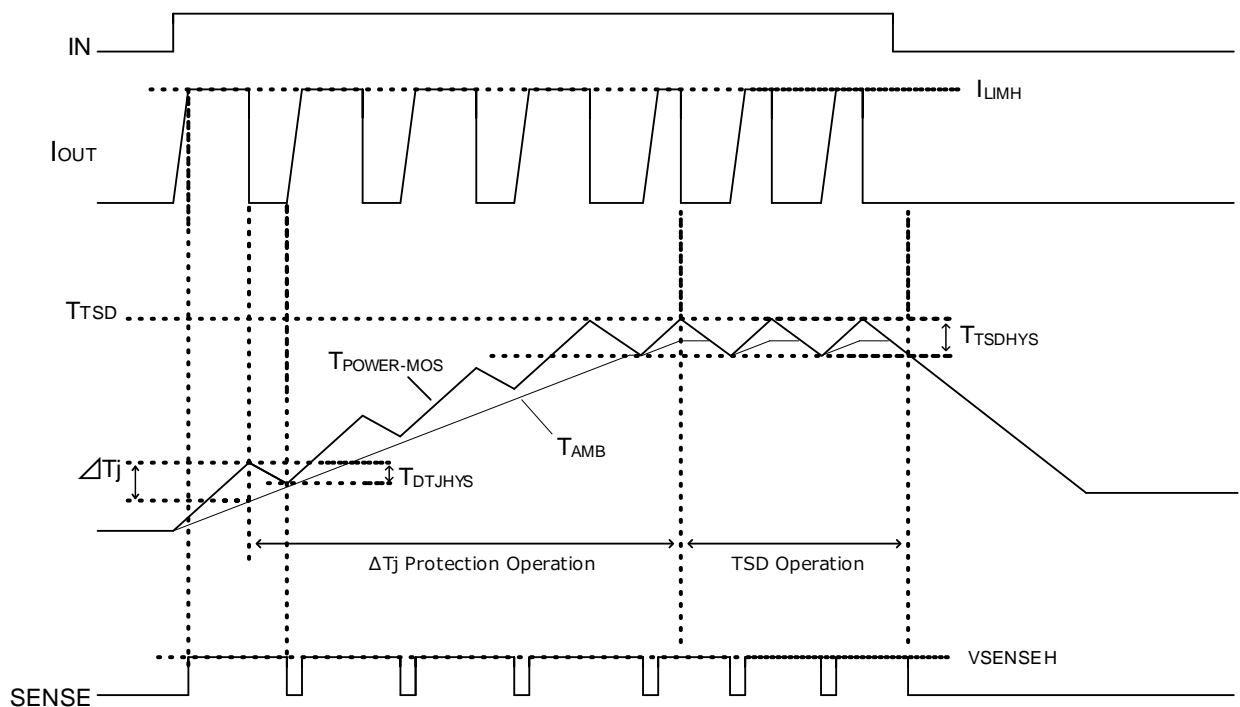


Figure 46. Thermal Shutdown Protection Timing Chart

Function Description - continued

5. Open Load Detection

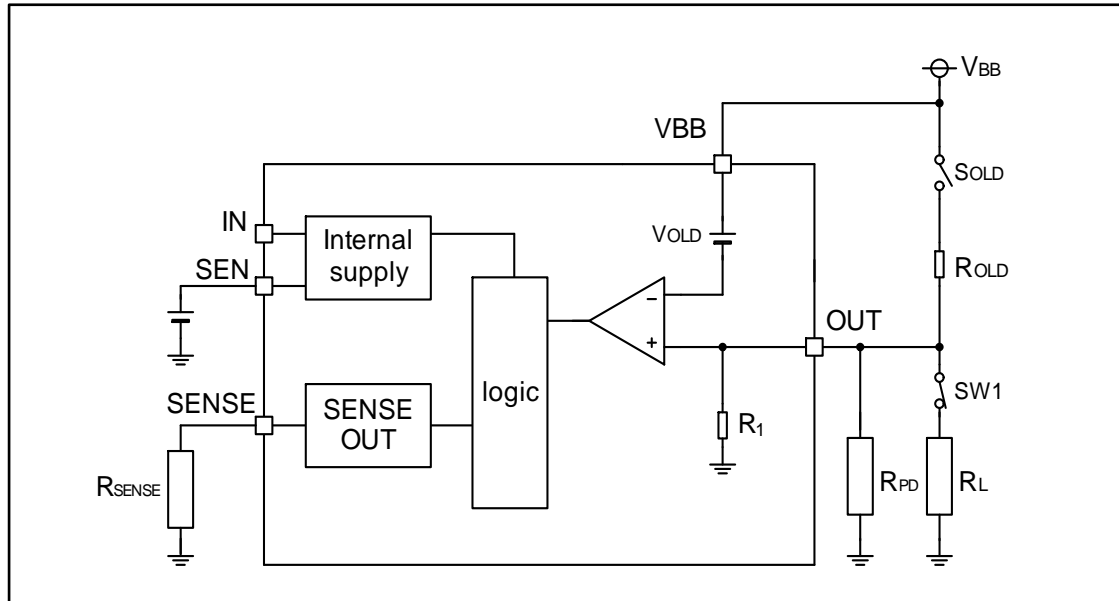


Figure 47. Open Load Detection Block Diagram

Open load can be detected by connecting an external resistance R_{OLD} between power supply V_{BB} and OUT. When output load is disconnected during IN is low, diagnostic output of the SENSE pin is turned to high to indicate abnormality. To reduce the standby current of the system, an open load resistance switch S_{OLD} is recommended. When the SW1 is OFF, voltage of the OUT does not fall to GND level. Because, when the IN pin is low, the voltage of the OUT pin does not become under or equal to the Output ON Detection Voltage. To pull down the OUT pin, insert the pulled down resistance R_{PD} is recommended. The resistance R_{PD} is 4.3 kΩ or less for outflow current from the OUT.

5.1 When the OUT is pulled down by the load (Normal function)

The value of external resistance R_{OLD} is decided based on used minimum power supply voltage (V_{BB}), internal resistance R_1 and open detection voltage V_{OLD} . External resistance R_{PD} is unnecessary. The equation for calculating the R_{OLD} value is shown below.

$$R_{OLD} < \frac{V_{BB} \times R_{1(Min)}}{V_{OLD(Max)}} - R_{1(Min)} \quad [\Omega]$$

The above formula is summarized as follows.

$$R_{OLD} < \frac{V_{BB}}{V_{BB} - 1.0} \times 300 \times 10^3 - 300 \times 10^3 \quad [\Omega]$$

R_{OLD} value is fell below the above calculated result.

5.2 If the SW is OFF, the output is no longer pulled down by the load

The value of external resistance R_{OLD} is decided based on used minimum power supply voltage (V_{BB}), external resistance R_{PD} and open detection voltage V_{OLD} .

The equation for calculating the R_{OLD} value is shown below.

$$R_{OLD} < \left(\frac{V_{BB}}{V_{OLD(Max)}} - 1 \right) \times \frac{R_{1(MIN)} \times R_{PD}}{R_{1(MIN)} + R_{PD}} \quad [\Omega]$$

When R_{PD} is 4.3 kΩ, the above formula is summarized as follows.

$$R_{OLD} < \left(\frac{V_{BB}}{V_{BB} - 1.0} - 1 \right) \times 4.24 \times 10^3 \quad [\Omega]$$

R_{OLD} value is fell below the above calculated result

Function Description - continued

5.3 SENSE output mask time at output falling

This IC diagnoses open load detection after the mask time (t_{OLD} : 250 μ s) inside the IC, when the IN voltage falls from High to Low,

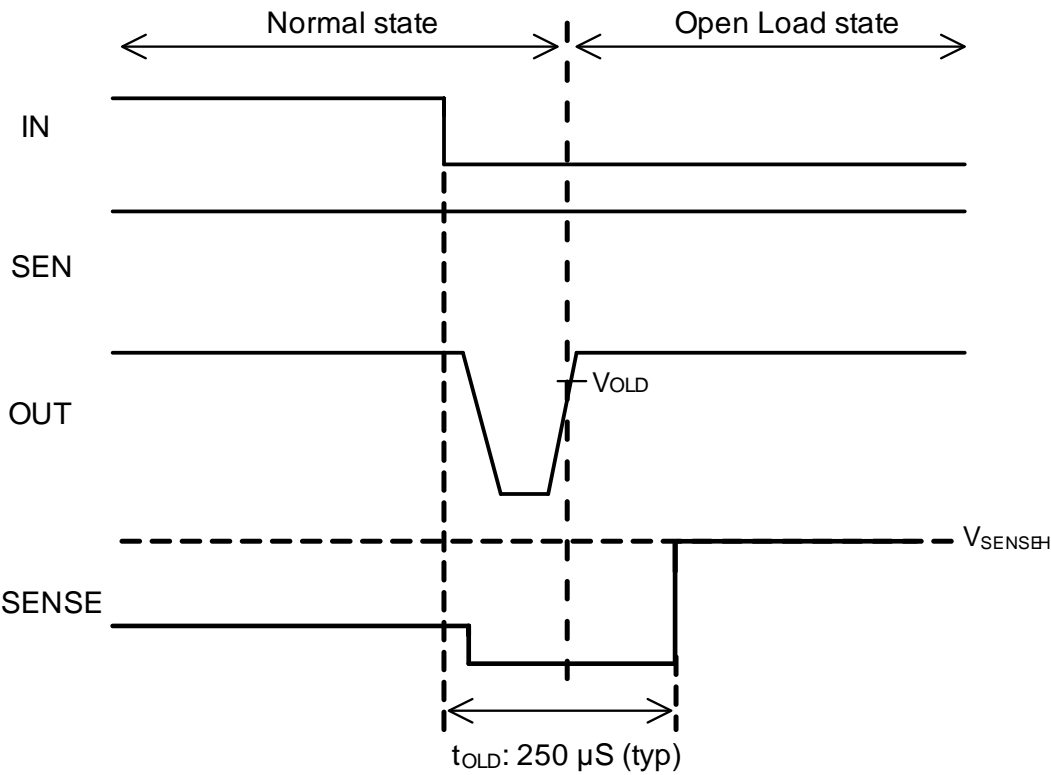


Figure 48. SENSE Output-Mask Timing Chart

Function Description - continued

6. Other Detection

6.1 GND open protection

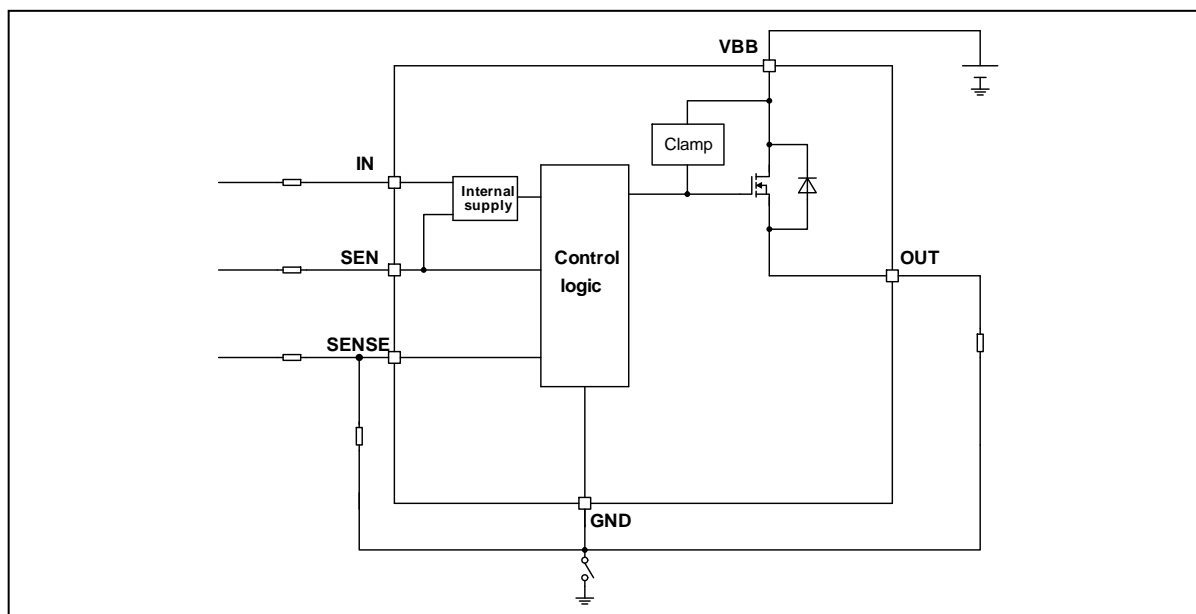


Figure 49. GND Open Detection Block Diagram

When GND of the IC is open, the output is switched OFF regardless of the input voltage. However, self-diagnostic output (SENSE) is not flagged. When an inductive load is connected, the active clamp operates when the GND pin is open

6.2 MCU I/O Protection

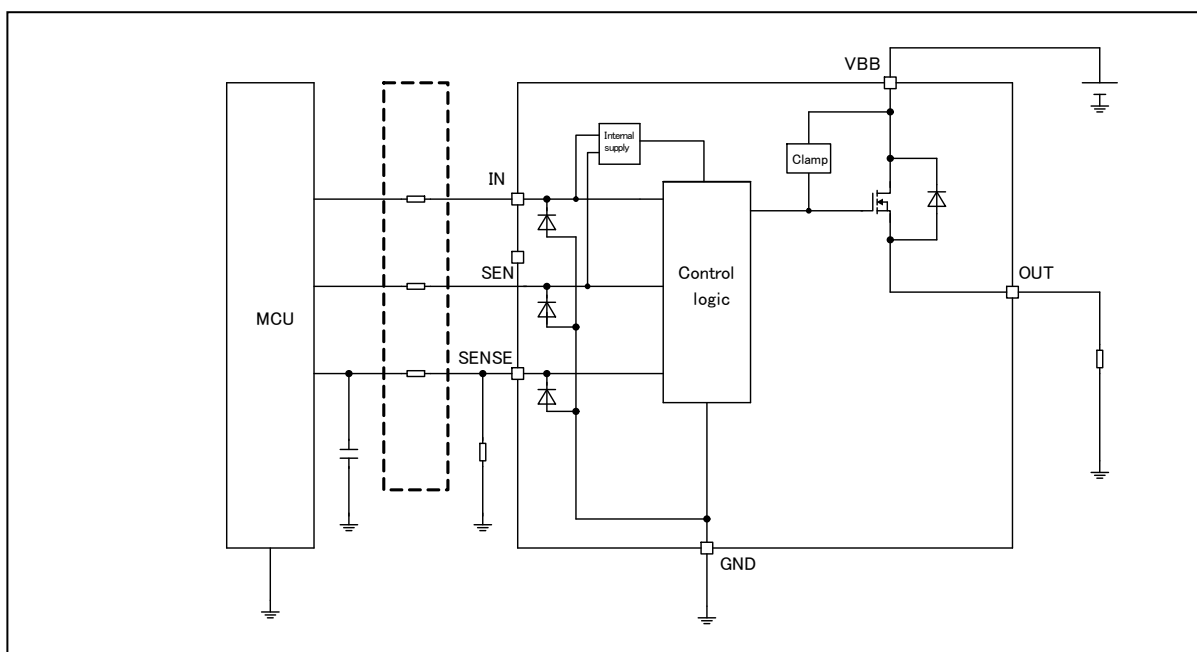


Figure 50. MCU I/O Protection

Negative surge voltage to the input, battery loss, and GND negative voltage may cause damage to the MCU I/O pin. To prevent these problems, a limiting resistor can be inserted between the input terminal and the MCU. 4.7 kΩ to 10 kΩ is recommended as the insert resistor.

Application Circuit Diagram

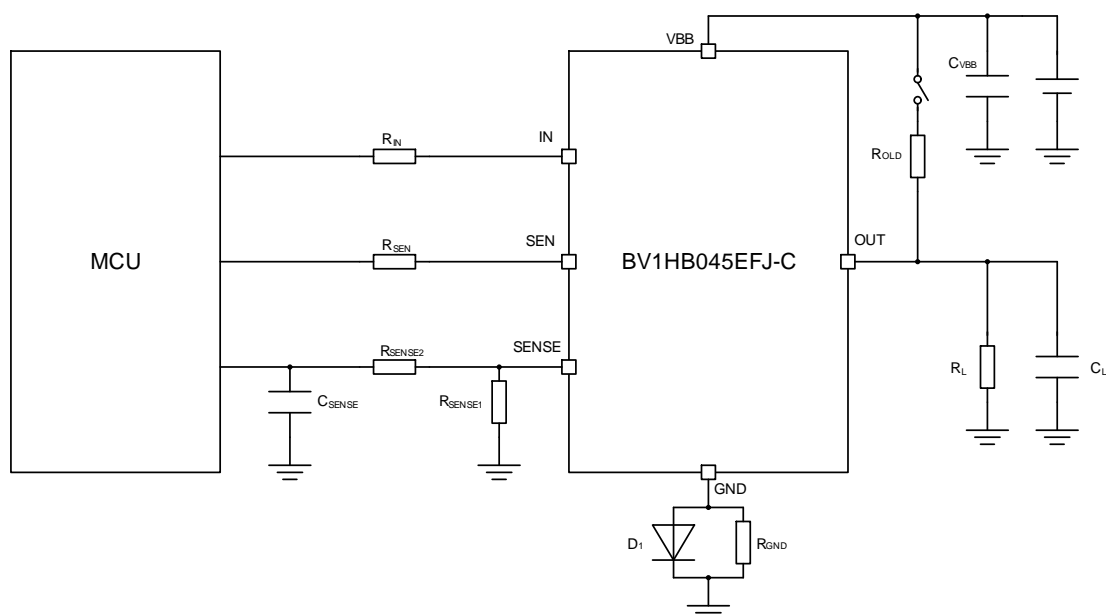
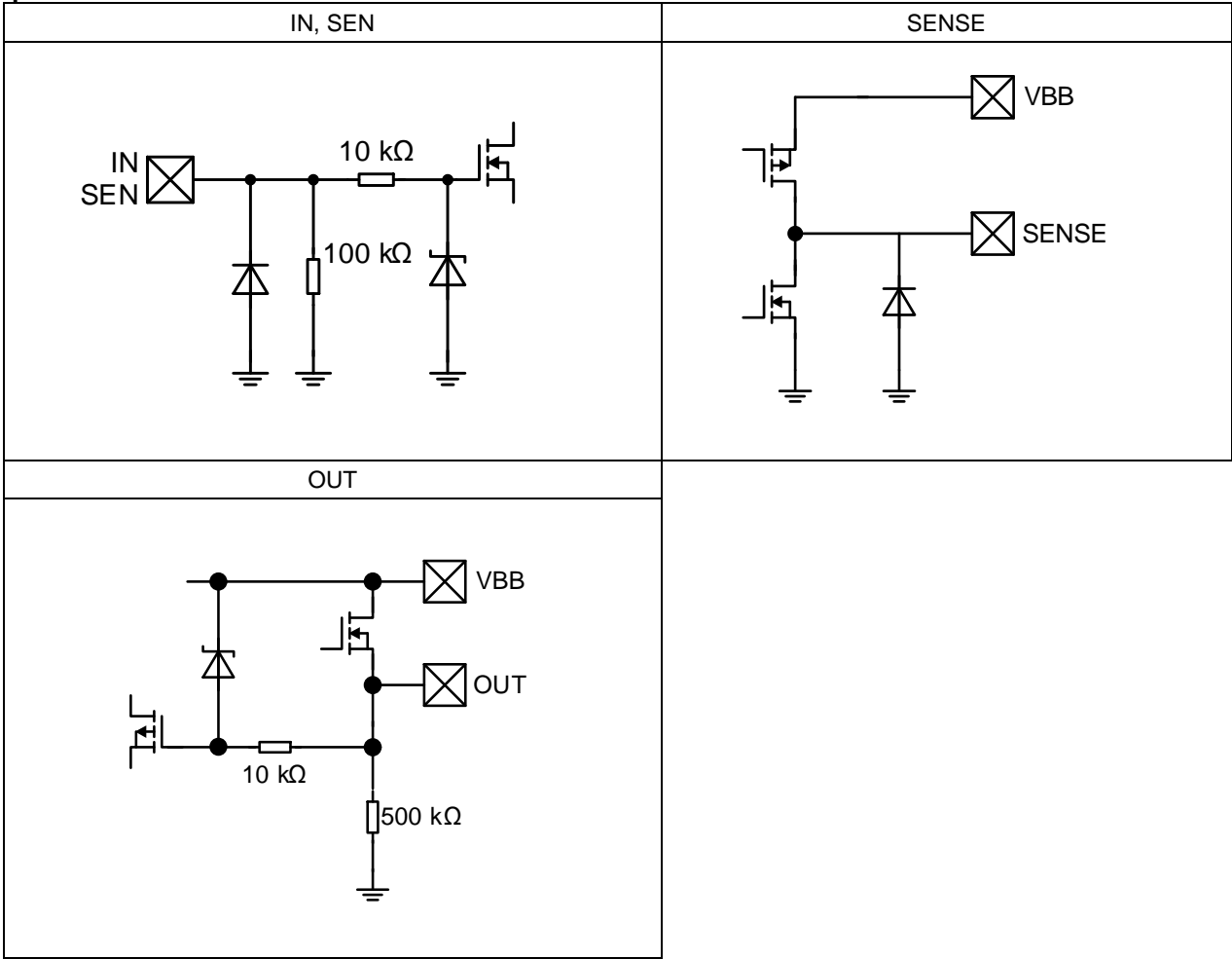


Figure 51. Application Circuit Diagram

Symbol	Value	Purpose
R_{IN}	4.7 k Ω	Limit resistance for negative surge
R_{SEN}	4.7 k Ω	Limit resistance for negative surge
R_{SENSE1}	1 k Ω	Insert the pull-dpwn resistor for using diagnostic function
R_{SENSE2}	10 k Ω	For Noise suppression filter
C_{SENSE}	100 pF	For Noise suppression filter
R_{GND}	100 Ω	Current limit resistance for reverse battery connection
D_{GND}	-	Protection Diode of BV1HB045EFJ-C for reverse battery connection
C_{VBB}	10 μ F	For battery line voltage spike filter
R_{OLD}	2 k Ω	Resistor for open load detection
C_L	1000 pF	Filter for radiation noise from outside
R_L	-	Output Load Resistor

I/O Equivalence Circuits



Resistance values shown in the diagrams above are typical values

Operational Notes

1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Recommended Operating Conditions

The function and operation of the IC are guaranteed within the range specified by the recommended operating conditions. The characteristic values are guaranteed only under the conditions of each item specified by the electrical characteristics.

6. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

7. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

8. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

9. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So, unless otherwise specified, unused input pins should be connected to the power supply or ground line.

10. Ceramic Capacitor

When using a ceramic capacitor, determine a capacitance value considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

11. Thermal Shutdown Function (TSD)

This IC has a built-in thermal shutdown function that prevents heat damage to the IC. Normal operation should always be within the IC's maximum junction temperature rating. If however the rating is exceeded for a continued period, the junction temperature (T_j) will rise which will activate the TSD function that will turn OFF power output pins. When the T_j falls below the TSD threshold, the circuits are automatically restored to normal operation. Note that the TSD function operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD function be used in a set design or for any purpose other than protecting the IC from heat damage.

Operational Notes – continued**12. Over Current Protection Function**

This IC incorporates an integrated overcurrent protection function that is activated when the load is shorted. This protection function is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection function.

13. Active Clamp Operation

The IC integrates the active clamp function to internally absorb the reverse energy E_L which is generated when the inductive load is turned off. When the active clamp operates, the thermal shutdown function does not work. Decide a load so that the reverse energy E_L is active clamp energy (refer to Figure 5. Active Clamp Energy vs Output Current) or under when inductive load is used.

14. Open Power Supply Pin

When the power supply pin (VBB) becomes open at ON (IN = High), the output is switched to OFF regardless of input voltage. If an inductive load is connected, the active clamp operates when VBB is open and becomes the same potential as that on the ground. At this time, the output voltage drops down to -48 V (Typ).

15. Open GND Pin

When the GND pin becomes open at ON (IN = High), the output is switched to OFF regardless of input voltage. If an inductive load is connected, the active clamp operates when the GND pin is open.

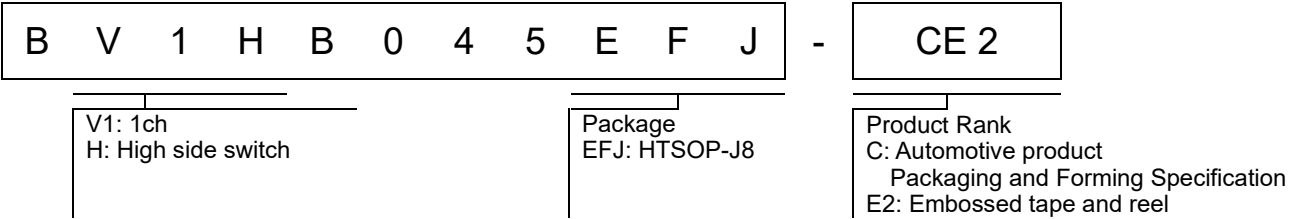
16. OUT Pin Voltage

Ensure that keep OUT pin voltage less than (VBB + 0.3 V) at any time, even during transient condition. Otherwise malfunction or other problems can occur.

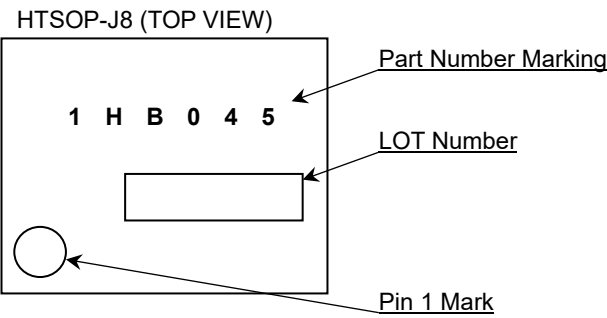
17. Same Pin Connection

Connect all VBB pins, GND pins, OUT pins to same line respectively.

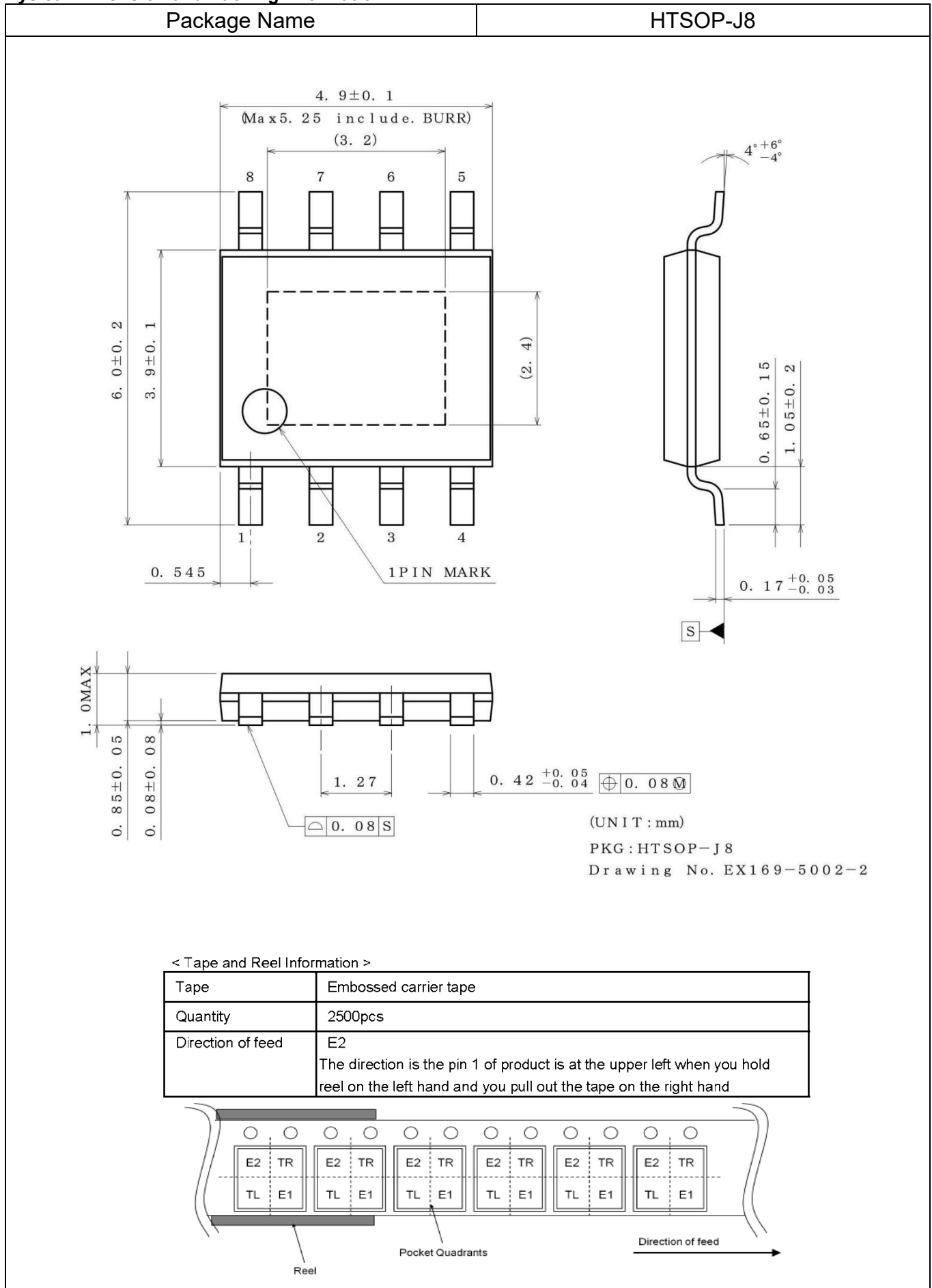
Ordering Information



Marking Diagram



Physical Dimension and Packing Information



Revision History

Date	Revision	Changes
3.Mar.2022	001	New Release

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 - [f] Sealing or coating our Products with resin or other coating materials
 - [g] Use of our Products without cleaning residue of flux (Exclude cases where no-clean type fluxes is used. However, recommend sufficiently about the residue.); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - [h] Use of the Products in places subject to dew condensation
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7. De-rate Power Dissipation depending on ambient temperature. When used in sealed area, confirm that it is the use in the range that does not exceed the maximum junction temperature.
8. Confirm that operation temperature is within the specified range described in the product specification.
9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

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