DATA SHEET

### **General Description**

The IDT8V40817 is a PLL-based clock generator specifically designed for Freescale Semiconductor Microprocessor and Microcontroller applications including the PowerPC and PowerQUICC. This device generates the microprocessor input clock and other microprocessor system and bus clocks at any one of four output frequencies. These frequencies include the popular 33- and 66-MHz PCI bus frequencies. The device offers five low-skew clock outputs plus three reference outputs. The clock input reference is 25 MHz and may be derived from an external source or by the addition of a 25MHz crystal to the on-chip crystal oscillator. The extended temperature range of the IDT8V40817 supports telecommunication and networking requirements.

### **Functional Description**

The IDT8V40817 uses a PLL with a 25MHz input reference frequency to generate a single bank of five configurable LVCMOS output clocks. The output frequency of this bank is configurable to either 25MHz, 33MHz, 50MHz, or 66MHz by two FSEL pins. The 25MHz reference may be either an external frequency source or a 25MHz crystal. The 25MHz crystal is directly connected to the XTAL\_IN and XTAL\_OUT pins with no additional components required. An external reference may be applied to the XTAL\_IN pin with the XTAL\_OUT pin left floating. The input reference, whether provided by a crystal or an external input, is also directly buffered to a second bank of three LVCMOS outputs. These outputs may be used as the clock source for processor I/O applications such as an Ethernet PHY. When FSEL0 and FSEL1 are both configured low, the QA outputs are directly fed from the input reference providing a total of eight low-skew 25MHz outputs. For all other combinations of FSEL0 and FSEL1 the single-ended LVCMOS outputs provide five low-skew outputs for use in driving a microprocessor or microcontroller clock input as well as other system components.

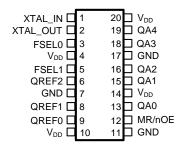
#### **Features**

- Five LVCMOS single-ended outputs for processor and other system circuitry
- Three buffered 25MHz reference clock outputs
- Crystal oscillator or external reference input
- Input reference frequency: 25MHz
- Selectable output frequencies: 25MHz, 33MHz, 50MHz or 66MHz
- Low Cycle-to-Cycle and Period jitter
- Supports Computing, Networking, and Telecommunications Applications
- Full 3.3V supply mode

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- -40°C to 85°C ambient operating temperature
- · Available in Lead-free (RoHS 6) packaging

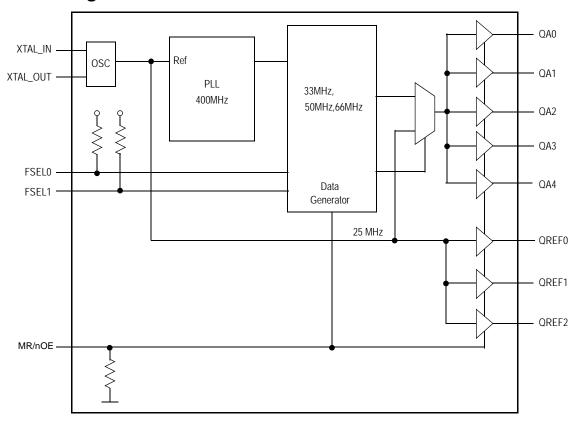
# **Pin Assignment**



IDT8V40817

20 LEAD QSOP 0.194" x 0.236" x 0.058" package body QG Package Top View

# **Block Diagram**



# Pin Description and Pin Characteristic Tables

**Table 1. Pin Descriptions** 

Number	Name	Ty	/pe	Description
1	XTAL_IN	Input		Crystal oscillator input.
2	XTAL_OUT	Output		Crystal oscillator output.
3, 5	FSEL0, FSEL1	Input	Pullup	Select pins. Configures Bank A output frequency.
4, 10, 14, 20	$V_{DD}$	Power		Power supply pins.
6, 8, 9	QREF2, QREF1, QREF0	Output		Single-ended reference clock output pins (25MHz). LVCMOS/LVTTL interface levels.
7, 11, 17	GND	Power		Power supply ground.
12	MR/nOE	Input	Pulldown	Master Reset/Output Enable pin. Enables all outputs.
13, 15, 16, 18, 19	QA0, QA1, QA2, QA3, QA4	Output		Single-ended clock output pins. LVCMOS/LVTTL interface levels.

NOTE: Pullup and Pulldown refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

**Table 2. Pin Characteristics** 

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
C <sub>IN</sub>	Input Capacitance MR/nOE, FSEL0, FSEL1				4		pF
C <sub>PD</sub>	Power Dissipation Capacitance (per output)		V <sub>DD</sub> = 3.465V		8.5		pF
R <sub>PULLUP</sub>	Input Pullup Resistor				75		kΩ
R <sub>PULLDOWN</sub>	Input Pulldown Resis	stor			75		kΩ

## **Function Table**

Table 3. FSELx Function Table)

Control	Default	00	01	10	11
FSEL[0:1]	11	25MHz fed directly from reference input, PLL disabled	33MHz	50MHz	66MHz

## **Absolute Maximum Ratings**

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics or AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Item	Rating
Supply Voltage, V <sub>DD</sub>	-0.3V to 3.8V
DC Input Current, I <sub>IN</sub>	±20mA
DC Output Current, I <sub>OUT</sub>	±75mA
Storage Temperature, T <sub>STG</sub>	-65°C to +125°C
Package Thermal Impedance, $\theta_{JA}$	107.3°C/W

### **DC Electrical Characteristics**

Table 4A. Power Supply DC Characteristics,  $V_{DD} = 3.3V \pm 5\%$ ,  $T_A = -40$ °C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{DD}$	Power Supply Voltage		3.135	3.3	3.465	V
I <sub>DD</sub>	Quiescent Supply Current			12	23	mA

Table 4B. LVCMOS/LVTTL DC Characteristics,  $V_{DD} = 3.3V \pm 5\%$ ,  $T_A = -40^{\circ}C$  to  $85^{\circ}C$ 

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V <sub>IL</sub>	Input Low Voltage					0.8	V
V <sub>IH</sub>	Input High Voltage			2		V <sub>DD</sub> + 0.3	V
$V_{IH}$	Input High Voltage	XTAL_IN	Input Threshold = V <sub>DD</sub> /2	2.4		V <sub>DD</sub> + 0.3	V
I <sub>IN</sub> <sup>(1)</sup>	Input Leakage Curren	t	V <sub>IN</sub> = V <sub>DDL</sub> or GND			150	μΑ
V <sub>OL</sub>	Output Low Voltage		I <sub>OL</sub> = 12mA			0.4	V
V <sub>OH</sub>	Output High Voltage		I <sub>OH</sub> = -12mA	2.4			V
Z <sub>OUT</sub>	Output Impedance				14		Ω

NOTE 1: Inputs have pulldown resistors affecting the input current.

**Table 5. Crystal Characteristics** 

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation Fundamental		ıl			
Resonance	nance Parallel Resonance				
Frequency			25		MHz
Load Capacitance (C <sub>L</sub> )			12		pF
Equivalent Series Resistance (ESR)			20		Ω
Shunt Capacitance		5		7	pF

### **AC Electrical Characteristics**

Table 6. AC Characteristics,  $V_{DD} = 3.3V \pm 5\%$ ,  $T_A = -40^{\circ}C$  to  $85^{\circ}C$ 

Symbol	Parameter	_	Test Conditions	Minimum	Typical	Maximum	Units
Input / O	utput Timing Specifica	tion			ı	-1	
f <sub>REF</sub>	Input Reference Frequ	iency	25MHz Input Crystal		25		MHz
$f_{VCO}$	VCO Frequency Rang	е			400		MHz
		QAx	FSEL[0:1] = 00		25		MHz
		QAx	FSEL[0:1] = 01, PLL Locked		33		MHz
f <sub>OUT</sub>	Output Frequency	QAx	FSEL[0:1] = 10, PLL Locked		50		MHz
		QAx	FSEL[0:1] = 11, PLL Locked		66		MHz
		QREFx			25		MHz
f <sub>REFPW</sub>	Reference Input Pulse Width		25MHz Reference	10			ns
odc	Output Duty Cycle			47	50	53	%
PLL Spe	cifications		1		I	1	
BW	PLL Closed Loop Ban NOTE 2	dwidth;			500		kHz
t <sub>LOCK</sub>	Maximum PLL Lock T	ime				10	ms
Skew an	d Jitter Specifications		I		1	1	
tsk(b)	Bank Skew; NOTE 3,	4				100	ps
tsk(o)	Output Skew; NOTE 3, 5		FSEL[0:1] = 00			200	ps
tjit(cc)	Cycle-to-Cycle Jitter	QAx	25MHz Input Reference			150	ps
tjit(per)	Period Jitter	QAx	25MHz Input Reference			100	ps
t <sub>R</sub> / t <sub>F</sub>	Rise and Fall Time	1	20% to 80%			1	ns

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE: AC Characteristics apply for parallel output termination of 50 $\Omega$  to V<sub>TT</sub>.

NOTE 1: Based upon recommended crystal specifications as outlined in IDT8V40817 Operation.

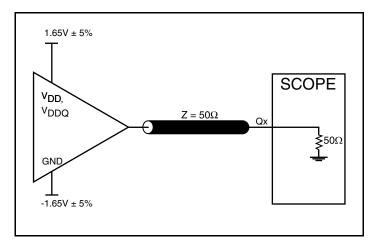
NOTE 2: -3dB point of PLL transfer characteristics.

NOTE 3: This parameter is defined in accordance with JEDEC Standard 65.

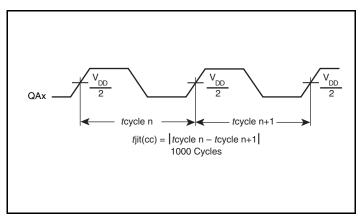
NOTE 4: Defined as skew within a bank of outputs at the same voltage and with equal load conditions. Measured at  $V_{DD}/2$ .

NOTE 5: Defined as skew between outputs at the same supply voltage and with equal load conditions. Measured at  $V_{DD}/2$ .

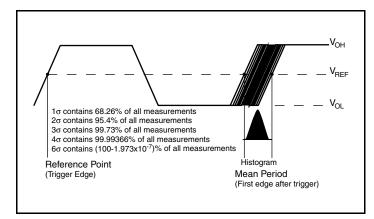
# **Parameter Measurement Information**



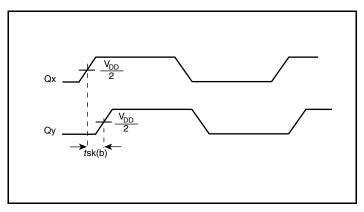
### **LVCMOS Output Load Test Circuit**



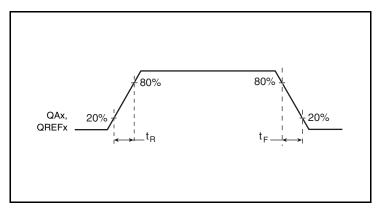
**Output Skew** 



**Cycle-to-Cycle Jitter** 



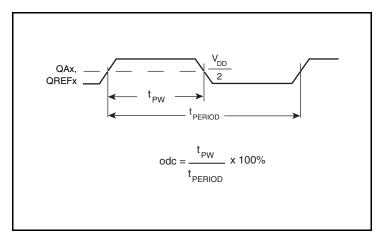
**Period Jitter** 



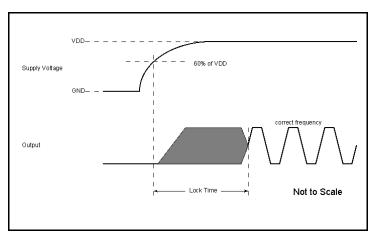
**Bank Skew** 

**Rise/Fall Time** 

# **Parameter Measurement Information, continued**



Output Duty Cycle/Pulse Width/Period



**Lock Time** 

### **Applications Information**

### **Recommendations for Unused Input and Output Pins**

#### Inputs:

#### **LVCMOS Control Pins**

All control pins have internal pullups or pulldowns; additional resistance is not required but can be added for additional protection. A  $1 \mathrm{k}\Omega$  resistor can be used.

#### **Outputs:**

#### **LVCMOS Outputs**

All unused LVCMOS outputs can be left floating. There should be no trace attached.

### **Schematic Example**

Figure 1 (next page) shows an example IDT8V40817 application schematic in which the device is operated at  $V_{DD} = 3.3V$ .

This example focuses on functional connections and is not configuration specific. Refer to the pin description and functional tables in the datasheet to ensure that the logic control inputs are properly set for the application.

The 12pF parallel resonant Fox FX325BS 25MHz crystal is used with tuning capacitors C1 = C2 = 5pF recommended for frequency accuracy. Depending on the parasitic of the printed circuit board layout, these values might require a slight adjustment to optimize the frequency accuracy. Crystals with other load capacitance specifications can be used. This will require adjusting C1 and C2. For this device, the crystal tuning capacitors are required for proper operation.

Crystal layout is very important to minimize capacitive coupling between the crystal pads and leads and other metal in the circuit board. Capacitive coupling to other conductors has two adverse effects; it reduces the oscillator frequency leaving less tuning margin and noise coupling from power planes and logic transitions on signal traces can pull the phase of the crystal resonance, inducing jitter. Routing I<sup>2</sup>C under the crystal is a very common layout error, based on the assumption that it is a low frequency signal and will not affect the crystal oscillation. In fact, I<sup>2</sup>C transition times are short enough to capacitively couple into the crystal if they are routed close enough to the crystal traces.

In layout, all capacitive coupling to the crystal from any signal trace is to be minimized, that is to the XTAL\_IN and XTAL\_OUT pins, traces to the crystal pads, the crystal pads and the tuning capacitors. Using a crystal on the top layer as an example, void all signal and power layers under the crystal connections between the top layer and the

ground plane used by the IDT8V40817. Then calculate the parasitic capacity to the ground and determine if it is large enough to preclude tuning the oscillator. If the coupling is excessive, particularly if the first layer under the crystal is a ground plane, a layout option is to void the ground plane and all deeper layers until the next ground plane is reached. The ground connection of the tuning capacitors should first be made between the capacitors on the top layer, then a single ground via is dropped to connect the tuning cap ground to the ground plane as close to the IDT8V40817 as possible as shown in the schematic.

As with any high speed analog circuitry, the power supply pins are vulnerable to random noise. To achieve optimum jitter performance, power supply isolation is required. The IDT8V40817 provides separate power supplies to isolate any high switching noise from coupling into the internal PLL.

In order to achieve the best possible filtering, it is recommended that the placement of the filter components be on the device side of the PCB as close to the power pins as possible. If space is limited, the 0.1uF capacitor in each power pin filter should be placed on the device side. The other components can be on the opposite side of the PCB.

Power supply filter recommendations are a general guideline to be used for reducing external noise from coupling into the devices. The filter performance is designed for a wide range of noise frequencies. This low-pass filter starts to attenuate noise at approximately 10kHz. If a specific frequency noise component is known, such as switching power supplies frequencies, it is recommended that component values be adjusted and if required, additional filtering be added. Additionally, good general design practices for power plane voltage stability suggests adding bulk capacitance in the local area of all devices.

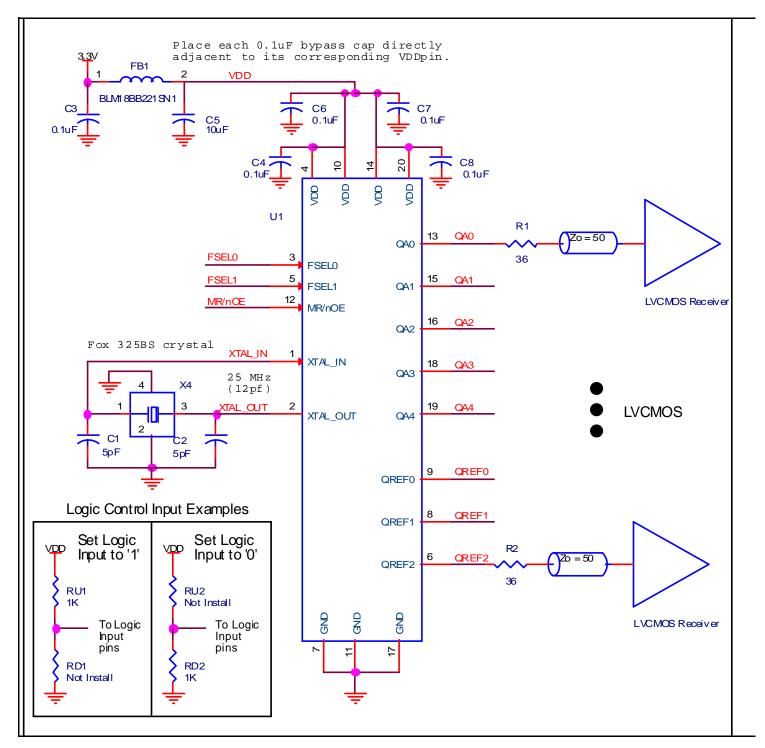


Figure 1. IDT8V40817 Application Schematic.

### **Power Considerations**

This section provides information on power dissipation and junction temperature for the IDT8V40817. Equations and example calculations are also provided.

#### 1. Power Dissipation.

The total power dissipation for the IDT8V40817 is the sum of the core power plus the analog power plus the power dissipated in the load(s). The following is the power dissipation for  $V_{DD} = 3.3V + 5\% = 3.465V$ , which gives worst case results.

Power (core)<sub>MAX</sub> = V<sub>DD MAX</sub> \* I<sub>DD</sub> = 3.465V \* 23mA) = 79.7mW

#### **Total Static Power:**

Power (core) $_{MAX} = 79.7 \text{mW}$ 

#### Dynamic Power Dissipation at F<sub>OUT</sub> (max.)

Total Power ( $F_{OUT\_MAX}$ ) = [( $C_{PD}$  \* N) \* Frequency \* ( $V_{DDO}$ )<sup>2</sup> = 8.5pF \* 66MHz \* (3.465V)<sup>2</sup> = **6.7mW per output N = number of outputs** 

= 53.844mW

#### **Total Power Dissipation**

- Total Power
  - = Total Static Power + Dynamic Power Dissipation
  - = 79.7 mW + 53.844 mW
  - = 133.58 mW

#### 2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj =  $\theta_{JA}$  \* Pd\_total + T<sub>A</sub>

Tj = Junction Temperature

 $\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

Pd\_total = Total Device Power Dissipation (example calculation is in section 1 above)

T<sub>A</sub> = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming no air flow and a multi-layer board, the appropriate value is 107.3°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is:

 $85^{\circ}\text{C} + 0.134\text{W} * 107.3^{\circ}\text{C/W} = 99.3^{\circ}\text{C}$ . This is below the limit of  $125^{\circ}\text{C}$ .

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 7. Thermal Resistance  $\theta_{JA}$  for 20 Lead QSOP, Forced Convection

$\theta_{JA}$ by Velocity					
Meters per Second	0	1	2.5		
Multi-Layer PCB, JEDEC Standard Test Boards	107.3°C/W	96.5°C/W	87.6°C/W		

# **Reliability Information**

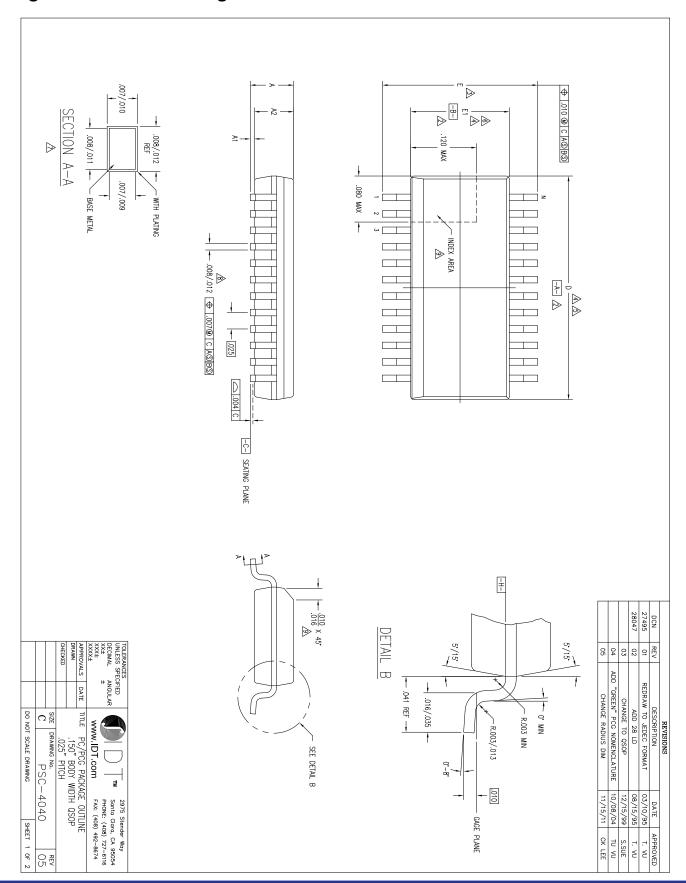
# Table 8. $\theta_{JB}$ vs. Air Flow Table for a 20 Lead QSOP

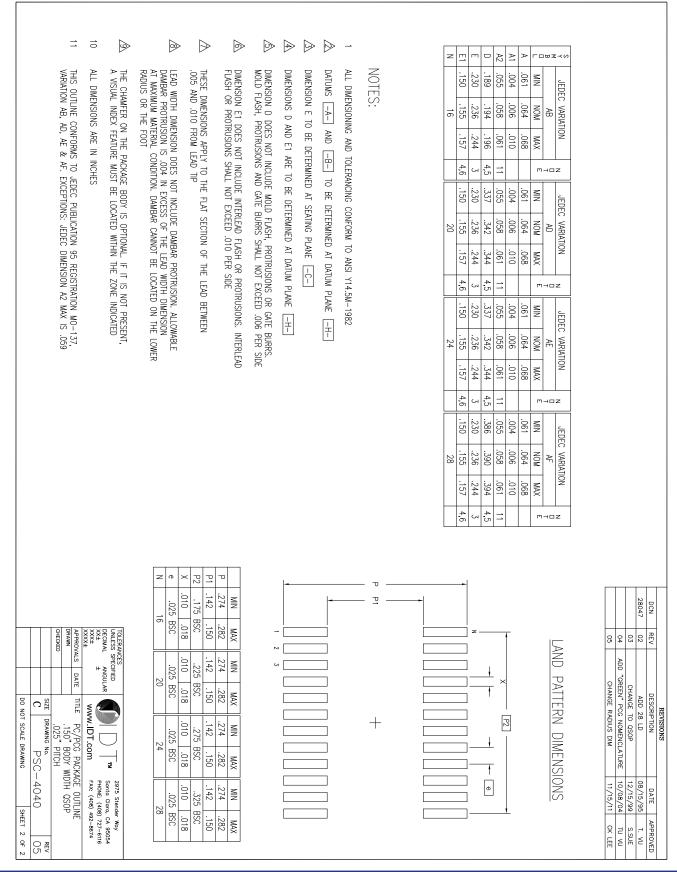
$\theta_{\sf JB}$ vs. Air Flow						
Meters per Second	0	1	2.5			
Multi-Layer PCB, JEDEC Standard Test Boards	107.3°C/W	96.5°C/W	87.6°C/W			

### **Transistor Count**

The transistor count for the IDT8V40817 is: 933

# **Package Outline and Package Dimensions**





# **Ordering Information**

## **Table 9. Ordering Information**

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
8V40817QG	IDT8V40817QG	"Lead-Free" 20 Lead QSOP	Tube	-40°C to 85°C
8V40817QG8	IDT8V40817QG	"Lead-Free" 20 Lead QSOP	Tape & Reel	-40°C to 85°C

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