

## CY7C1470BV33 CY7C1472BV33, CY7C1474BV33 72 Mbit (2M x 36/4M x 18/1M x 72) Pipelined SRAM with NoBL<sup>™</sup> Architecture

#### Features

- Pin-compatible and functionally equivalent to ZBT<sup>™</sup>
- Supports 250 MHz bus operations with zero wait states
   Available speed grades are 250, 200, and 167 MHz
- Internally self timed output buffer control to eliminate the need to use asynchronous OE
- Fully registered (inputs and outputs) for pipelined operation
- Byte Write capability
- Single 3.3V power supply
- 3.3V/2.5V IO power supply
- Fast clock-to-output time □ 3.0 ns (for 250 MHz device)
- Clock Enable (CEN) pin to suspend operation
- Synchronous self timed writes
- CY7C1470BV33, CY7C1472BV33 available in JEDEC-standard Pb-free 100-pin TQFP, Pb-free and non-Pb-free 165-ball FBGA package. CY7C1474BV33 available in Pb-free and non-Pb-free 209-ball FBGA package
- IEEE 1149.1 JTAG Boundary Scan compatible
- Burst capability—linear or interleaved burst order
- "ZZ" Sleep Mode option and Stop Clock option

### **Functional Description**

The CY7C1470BV33, CY7C1472BV33, and CY7C1474BV33 are 3.3V, 2M x 36/4M x 18/1M x 72 Synchronous pipelined burst SRAMs with No Bus Latency<sup>™</sup> (NoBL<sup>™</sup>) logic, respectively. They are designed to support unlimited true back-to-back read or write operations with no wait states. The CY7C1470BV33, CY7C1472BV33, and CY7C1474BV33 are equipped with the advanced (NoBL) logic required to enable consecutive read or write operations with data being transferred on every clock cycle. This feature dramatically improves the throughput of data in systems that require frequent read or write transitions. The CY7C1470BV33, CY7C1472BV33, and CY7C1474BV33 are pin compatible and functionally equivalent to ZBT devices.

All synchronous inputs pass through input registers controlled by the rising edge of the clock. All data outputs pass through output registers controlled by the rising edge of the clock. The clock input is qualified by the Clock Enable (CEN) signal, which when deasserted suspends operation and extends the previous clock cycle.

Three synchronous Chip Enables  $(\overline{CE}_1, CE_2, \overline{CE}_3)$  and an asynchronous Output Enable  $(\overline{OE})$  provide for easy bank selection and output tri-state control. To avoid bus contention, the output drivers are synchronously tri-stated during the data portion of a write sequence.

### **Selection Guide**

Description	250 MHz	200 MHz	167 MHz	Unit
Maximum Access Time	3.0	3.0	3.4	ns
Maximum Operating Current	500	500	450	mA
Maximum CMOS Standby Current	120	120	120	mA

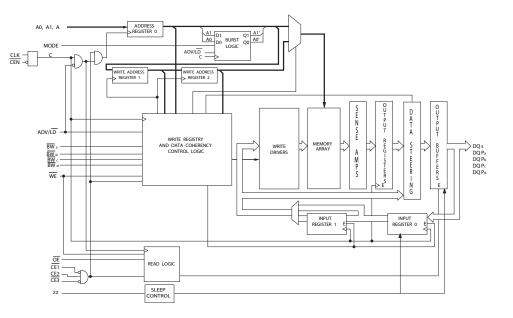
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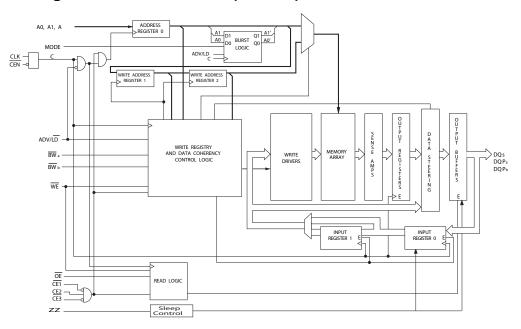
San Jose, CA 95134-1709 • 408-943-2600 Revised August 25, 2009



### Logic Block Diagram – CY7C1470BV33 (2M x 36)

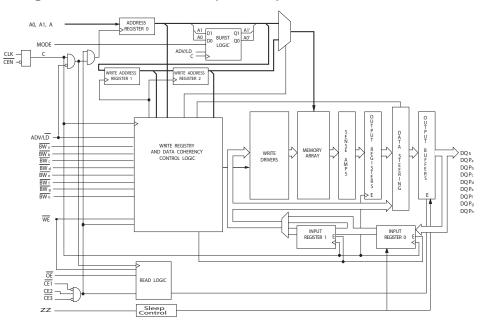


### Logic Block Diagram – CY7C1472BV33 (4M x 18)





### Logic Block Diagram – CY7C1474BV33 (1M x 72)

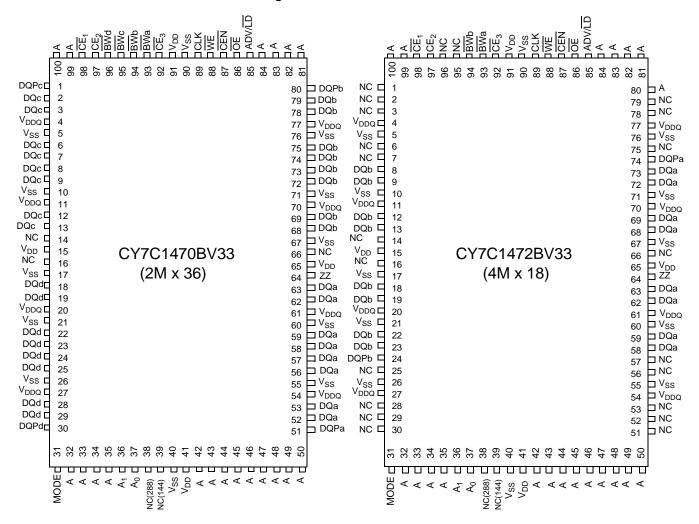




### CY7C1470BV33 CY7C1472BV33, CY7C1474BV33

### **Pin Configurations**

Figure 1. 100-Pin TQFP Pinout





### Pin Configurations (continued)

	CY7C1470BV33 (2M X 36)										
	1	2	3	4	5	6	7	8	9	10	11
Α	NC/576M	А	CE <sub>1</sub>	BWc	BWb	CE <sub>3</sub>	CEN	ADV/LD	А	А	NC
В	NC/1G	А	CE2	BWd	BWa	CLK	WE	OE	А	А	NC
С	DQP <sub>c</sub>	NC	V <sub>DDQ</sub>	V <sub>SS</sub>	V <sub>DDQ</sub>	NC	DQPb				
D	DQ <sub>c</sub>	DQ <sub>c</sub>	$V_{DDQ}$	$V_{DD}$	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>DD</sub>	$V_{DDQ}$	$DQ_b$	DQb
Е	DQ <sub>c</sub>	$DQ_{c}$	V <sub>DDQ</sub>	$V_{DD}$	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>DD</sub>	V <sub>DDQ</sub>	$DQ_b$	DQb
F	DQ <sub>c</sub>	$DQ_{c}$	V <sub>DDQ</sub>	$V_{DD}$	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>DD</sub>	V <sub>DDQ</sub>	$DQ_b$	DQb
G	DQ <sub>c</sub>	$DQ_{c}$	V <sub>DDQ</sub>	$V_{DD}$	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>DD</sub>	V <sub>DDQ</sub>	$DQ_b$	DQb
Н	NC	NC	NC	$V_{DD}$	V <sub>SS</sub>	V <sub>SS</sub>	$V_{SS}$	V <sub>DD</sub>	NC	NC	ZZ
J	DQd	$DQ_{d}$	V <sub>DDQ</sub>	V <sub>DD</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>DD</sub>	V <sub>DDQ</sub>	DQa	DQa
ĸ	DQ <sub>d</sub>	$DQ_{d}$	$V_{DDQ}$	$V_{DD}$	V <sub>SS</sub>	V <sub>SS</sub>	$V_{SS}$	$V_{DD}$	$V_{DDQ}$	$DQ_{a}$	DQa
L	DQ <sub>d</sub>	$DQ_{d}$	$V_{DDQ}$	$V_{DD}$	V <sub>SS</sub>	V <sub>SS</sub>	$V_{SS}$	$V_{DD}$	$V_{DDQ}$	$DQ_{a}$	DQa
М	DQd	$DQ_{d}$	V <sub>DDQ</sub>	$V_{DD}$	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>DD</sub>	V <sub>DDQ</sub>	DQa	DQa
Ν	DQP <sub>d</sub>	NC	$V_{DDQ}$	V <sub>SS</sub>	NC	NC	NC	V <sub>SS</sub>	$V_{DDQ}$	NC	DQPa
Р	NC/144M	А	А	А	TDI	A1	TDO	А	А	А	NC/288M
R	MODE	А	А	А	TMS	A0	TCK	A	А	А	A

### 165-Ball FBGA (15 x 17 x 1.4 mm) CY7C1470BV33 (2M x 36)

### CY7C1472BV33 (4M x 18)

	1	2	3	4	5	6	7	8	9	10	11
Α	NC/576M	А	CE <sub>1</sub>	BWb	NC	$\overline{CE}_3$	CEN	ADV/LD	А	А	A
В	NC/1G	А	CE2	NC	BWa	CLK	WE	OE	А	А	NC
С	NC	NC	V <sub>DDQ</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>DDQ</sub>	NC	DQPa
D	NC	$DQ_b$	V <sub>DDQ</sub>	$V_{DD}$	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>DD</sub>	$V_{DDQ}$	NC	DQa
E	NC	$DQ_b$	V <sub>DDQ</sub>	V <sub>DD</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>DD</sub>	V <sub>DDQ</sub>	NC	DQa
F	NC	$DQ_b$	V <sub>DDQ</sub>	V <sub>DD</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>DD</sub>	V <sub>DDQ</sub>	NC	DQa
G	NC	$DQ_b$	V <sub>DDQ</sub>	V <sub>DD</sub>	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>	V <sub>DD</sub>	V <sub>DDQ</sub>	NC	DQa
Н	NC	NC	NC	V <sub>DD</sub>	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>	V <sub>DD</sub>	NC	NC	ZZ
J	DQb	NC	V <sub>DDQ</sub>	V <sub>DD</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>DD</sub>	$V_{DDQ}$	DQa	NC
ĸ	DQb	NC	$V_{DDQ}$	$V_{DD}$	V <sub>SS</sub>	$V_{SS}$	$V_{SS}$	$V_{DD}$	$V_{DDQ}$	$DQ_{a}$	NC
L	DQb	NC	$V_{DDQ}$	$V_{DD}$	V <sub>SS</sub>	$V_{SS}$	$V_{SS}$	V <sub>DD</sub>	$V_{DDQ}$	$DQ_{a}$	NC
М	DQb	NC	V <sub>DDQ</sub>	V <sub>DD</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>DD</sub>	V <sub>DDQ</sub>	DQa	NC
Ν	DQPb	NC	V <sub>DDQ</sub>	V <sub>SS</sub>	NC	NC	NC	V <sub>SS</sub>	V <sub>DDQ</sub>	NC	NC
Р	NC/144M	А	А	А	TDI	A1	TDO	Α	А	А	NC/288M
R	MODE	А	А	А	TMS	A0	TCK	А	А	А	A



### Pin Configurations (continued)

### 209-Ball FBGA (14 x 22 x 1.76 mm)

	1	2	3	4	5	6	7	8	9	10	11
Α	DQg	DQg	А	CE <sub>2</sub>	А	ADV/LD	А	CE <sub>3</sub>	А	DQb	DQb
В	DQg	DQg	BWS <sub>c</sub>	BWSg	NC	WE	А	BWSb	BWS <sub>f</sub>	DQb	DQb
С	DQg	DQg	BWS <sub>h</sub>	BWSd	NC/576M	CE <sub>1</sub>	NC	BWSe	BWSa	DQb	DQb
D	DQg	DQg	V <sub>SS</sub>	NC	NC/1G	OE	NC	NC	$V_{SS}$	DQb	DQb
E	DQPg	DQPc	V <sub>DDQ</sub>	$V_{DDQ}$	V <sub>DD</sub>	V <sub>DD</sub>	$V_{DD}$	$V_{DDQ}$	V <sub>DDQ</sub>	DQPf	DQPb
F	DQc	DQc	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>	NC	$V_{SS}$	V <sub>SS</sub>	$V_{SS}$	DQf	DQf
G	DQc	DQc	V <sub>DDQ</sub>	$V_{DDQ}$	V <sub>DD</sub>	NC	$V_{DD}$	$V_{DDQ}$	V <sub>DDQ</sub>	DQf	DQf
н	DQc	DQc	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>	NC	$V_{SS}$	V <sub>SS</sub>	V <sub>SS</sub>	DQf	DQf
J	DQc	DQc	V <sub>DDQ</sub>	$V_{DDQ}$	V <sub>DD</sub>	NC	$V_{DD}$	$V_{DDQ}$	V <sub>DDQ</sub>	DQf	DQf
К	NC	NC	CLK	NC	$V_{SS}$	CEN	$V_{SS}$	NC	NC	NC	NC
L	DQh	DQh	V <sub>DDQ</sub>	$V_{DDQ}$	V <sub>DD</sub>	NC	$V_{DD}$	$V_{DDQ}$	V <sub>DDQ</sub>	DQa	DQa
М	DQh	DQh	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>	NC	$V_{SS}$	$V_{SS}$	$V_{SS}$	DQa	DQa
Ν	DQh	DQh	V <sub>DDQ</sub>	$V_{DDQ}$	V <sub>DD</sub>	NC	$V_{DD}$	$V_{DDQ}$	V <sub>DDQ</sub>	DQa	DQa
Р	DQh	DQh	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>	ZZ	$V_{SS}$	$V_{SS}$	V <sub>SS</sub>	DQa	DQa
R	DQPd	DQPh	V <sub>DDQ</sub>	$V_{DDQ}$	V <sub>DD</sub>	V <sub>DD</sub>	$V_{DD}$	$V_{DDQ}$	V <sub>DDQ</sub>	DQPa	DQPe
Т	DQd	DQd	V <sub>SS</sub>	NC	NC	MODE	NC	NC	V <sub>SS</sub>	DQe	DQe
U	DQd	DQd	NC/144M	А	А	А	А	А	NC/288M	DQe	DQe
V	DQd	DQd	А	А	А	A1	А	А	А	DQe	DQe
W	DQd	DQd	TMS	TDI	А	A0	А	TDO	ТСК	DQe	DQe

### CY7C1474BV33 (1M × 72)



### **Pin Definitions**

Pin Name	ІО Туре	Pin Description
A0 A1 A	Input- Synchronous	Address Inputs Used to Select One of the Address Locations. Sampled at the rising edge of the CLK.
BW <sub>a</sub> BW <sub>b</sub> BW <sub>c</sub> BW <sub>d</sub> BW <sub>e</sub> BW <sub>f</sub> BW <sub>g</sub> BW <sub>h</sub>	Input- Synchronous	<b>Byte Write Select Inputs, Active LOW</b> . Qualified with $\overline{WE}$ to conduct writes to the SRAM. Sampled on the rising edge of <u>CLK</u> . $\overline{BW}_a$ controls $DQ_a$ and $DQP_a$ , $\overline{BW}_b$ controls $DQ_b$ and $DQP_b$ , $\overline{BW}_c$ controls $DQ_c$ and $\underline{DQP}_c$ , $\overline{BW}_d$ controls $DQ_d$ and $\underline{DQP}_d$ , $\overline{BW}_e$ controls $DQ_e$ and $DQP_e$ , $\overline{BW}_f$ controls $DQ_f$ and $DQP_f$ , $\overline{BW}_g$ controls $DQ_g$ and $DQP_g$ , $\overline{BW}_h$ controls $DQ_h$ and $DQP_h$ .
WE	Input- Synchronous	Write Enable Input, Active LOW. Sampled on the rising edge of CLK if CEN is active LOW. This signal must be asserted LOW to initiate a write sequence.
ADV/LD	Input- Synchronous	Advance/Load Input Used to Advance the On-chip Address Counter or Load a New Address. When HIGH (and CEN is asserted LOW) the internal burst counter is advanced. When LOW, a new address can be loaded into the device for an access. After being deselected, ADV/LD must be driven LOW to load a new address.
CLK	Input- Clock	<b>Clock Input</b> . Used to cap <u>ture all synchronous inputs to the device.</u> CLK is qualified with $\overline{CEN}$ . CLK is only recognized if CEN is active LOW.
CE <sub>1</sub>	Input- Synchronous	<b>Chip Enable 1 Input, Active LOW</b> . Sampled on the rising edge of CLK. Used in conjunction with $CE_2$ and $\overline{CE}_3$ to select or deselect the device.
CE <sub>2</sub>	Input- Synchronous	<b>Chip</b> Enable <u>2</u> Input, Active HIGH. Sampled on the rising edge of CLK. Used in conjunction with $CE_1$ and $CE_3$ to select or deselect the device.
CE <sub>3</sub>	Input- Synchronous	<u>Chip Enable 3 Input, Active LOW</u> . Sampled on the rising edge of CLK. Used in conjunction with $CE_1$ and $CE_2$ to select or deselect the device.
ŌĒ	Input- Asynchronous	<b>Output Enable, Active LOW</b> . Combined with the synchronous logic block inside the device to control the direction of the IO pins. When LOW, the IO pins are enabled to behave as outputs. When deasserted HIGH, IO pins are tri-stated, and act as input data pins. OE is masked during the data portion of a write sequence, during the first clock when emerging from a deselected state and when the device has been deselected.
CEN	Input- Synchronous	<b>Clock Enable Input, Active LOW</b> . When asserted LOW the clock signal is recognized by the SRAM. When deasserted HIGH the clock signal is masked. Since deasserting CEN does not deselect the device, CEN can be used to extend the previous cycle when required.
DQ <sub>S</sub>	IO- Synchronous	<b>Bidirectional Data IO Lines</b> . As inputs, they feed into an on-chip data register that is triggered by the rising edge of CLK. As outputs, they deliver the data contained in the memory location specified by $A_{f1Z:0]}$ during the previous clock rise of the read cycle. The direction of the pins is controlled by OE and the internal control logic. When OE is asserted LOW, the pins can behave as outputs. When HIGH, $DQ_a$ - $DQ_d$ are placed in a tri-state condition. The outputs are automatically tri-stated during the data portion of a write sequence, during the first clock when emerging from a deselected state, and when the device is deselected, regardless of the state of OE.
DQP <sub>X</sub>	IO- Synchronous	<b>Bidirectional Data Parity IO Lines</b> . Functionally, these signals are identical to $DQ_X$ . During write sequences, $DQP_a$ is controlled by $BW_a$ , $DQP_b$ is controlled by $BW_b$ , $DQP_c$ is controlled by $BW_c$ , and $DQP_d$ is controlled by $BW_d$ , $DQP_e$ is controlled by $BW_e$ , $DQP_f$ is controlled by $BW_f$ , $DQP_g$ is controlled by $BW_g$ , $DQP_f$ is controlled by $BW_f$ , $DQP_g$ is controlled by $BW_g$ , $DQP_h$ is controlled by $BW_h$ .
MODE	Input Strap Pin	<b>Mode Input</b> . Selects the burst order of the device. Tied HIGH selects the interleaved burst order. Pulled LOW selects the linear burst order. MODE must not change states during operation. When left floating MODE defaults HIGH, to an interleaved burst order.
TDO	JTAG Serial Output Synchronous	Serial Data Out to the JTAG Circuit. Delivers data on the negative edge of TCK.
TDI	JTAG Serial Input Synchronous	Serial Data In to the JTAG Circuit. Sampled on the rising edge of TCK.



### Pin Definitions (continued)

Pin Name	Ю Туре	Pin Description
TMS	Test Mode Select Synchronous	This Pin Controls the Test Access Port State Machine. Sampled on the rising edge of TCK.
тск	JTAG Clock	Clock Input to the JTAG Circuitry.
V <sub>DD</sub>	Power Supply	Power Supply Inputs to the Core of the Device.
V <sub>DDQ</sub>	IO Power Supply	Power Supply for the IO Circuitry.
V <sub>SS</sub>	Ground	Ground for the Device. Should be connected to ground of the system.
NC	-	No Connects. This pin is not connected to the die.
NC(144M, 288M, 576M, 1G)	_	<b>These Pins are Not Connected</b> . They are used for expansion to the 144M, 288M, 576M, and 1G densities.
ZZ	Input- Asynchronous	<b>ZZ "Sleep" Input</b> . This active HIGH input places the device in a non-time critical "sleep" condition with data integrity preserved. During normal operation, this pin must be LOW or left floating. ZZ pin has an internal pull-down.



### **Functional Overview**

The CY7C1470BV33, CY7C1472BV33, and CY7C1474BV33 are synchronous-pipelined Burst NoBL SRAMs designed specifically to eliminate wait states during read or write transitions. All synchronous inputs pass through input registers controlled by the rising edge of the clock. The clock signal is qualified with the Clock Enable input signal (CEN). If CEN is HIGH, the clock signal is not recognized and all internal states are maintained. All synchronous operations are qualified with CEN. All data outputs pass through output registers controlled by the rising edge of the clock. Maximum access delay from the clock rise (t<sub>CO</sub>) is 3.0 ns (250 MHz device).

Accesses can be initiated by asserting all three Chip Enables  $(CE_1, CE_2, CE_3)$  active at the rising edge of the clock. If CEN is active LOW and ADV/LD is asserted LOW, the address presented to the device is latched. The access can either be a read or write operation, depending on the status of the Write Enable (WE). BW<sub>[X]</sub> can be used to conduct Byte Write operations.

Write operations are qualified by the Write Enable ( $\overline{\text{WE}}$ ). All writes are simplified with on-chip synchronous self timed write circuitry.

Three synchronous Chip En<u>ables</u> ( $\overline{CE}_1$ ,  $CE_2$ ,  $\overline{CE}_3$ ) and an asynchronous Output Enable ( $\overline{OE}$ ) simplify depth expansion. <u>All</u> operations (reads, writes, and deselects) are pipelined. ADV/ $\overline{LD}$  must be driven LOW after the device has been deselected to load a new address for the next operation.

#### Single Read Accesses

A read access is initiated when the following conditions are satisfied at clock rise: (1)  $\overline{CEN}$  is asserted LOW, (2)  $CE_{1}$ ,  $CE_{2}$ , and CE<sub>3</sub> are ALL asserted active. (3) the input signal WE is deasserted HIGH, and (4) ADV/LD is asserted LOW. The address presented to the address inputs is latched into the Address Register and presented to the memory core and control logic. The control logic determines that a read access is in progress and allows the requested data to propagate to the input of the output register. At the rising edge of the next clock the requested data is allowed to propagate through the output register and onto the data bus within 3.0 ns (250 MHz device) provided OE is active LOW. After the first clock of the read access the output buffers are controlled by OE and the internal control logic. OE must be driven LOW to drive out the requested data. During the second clock, a subsequent operation (read, write, or deselect) can be initiated. Deselecting the device is also pipelined. Therefore, when the SRAM is deselected at clock rise by one of the chip enable signals, its output tri-states following the next clock rise.

#### **Burst Read Accesses**

The CY7C1470BV33, CY7C1472BV33, and CY7C1474BV33 have an on-chip burst counter that enables the user to supply a single address and conduct up to four reads without reasserting the address inputs. ADV/LD must be driven LOW to load a new address into the SRAM, as described in the Single Read Accesses section. The sequence of the burst counter is determined by the MODE input signal. A LOW input on MODE selects a linear burst mode, a HIGH selects an interleaved burst sequence. Both burst counters use A0 and A1 in the burst sequence, and wraps around when incremented sufficiently. A

#### Single Write Accesses

Write accesses are initiated when the following conditions are satisfied at clock rise: (1)  $\overline{CEN}$  is asserted LOW, (2)  $\overline{CE_{1,}CE_{2}}$ , and  $\overline{CE_{3}}$  are ALL asserted active, and (3) the signal WE is asserted LOW. The address presented to the address inputs is loaded into the Address Register. The write signals are latched into the Control Logic block.

On the subsequent clock rise the data lines are automatically tri-stated regardless of the state of the  $\overline{OE}$  input signal. This allows the external logic to present the data on DQ and DQP (DQ<sub>a,b,c,d</sub>/DQP<sub>a,b,c,d</sub> for CY7C1470BV33, DQ<sub>a,b</sub>/DQP<sub>a,b</sub> for CY7C1472BV33, and DQ<sub>a,b,c,d,e,f,g,h</sub>/DQP<sub>a,b,c,d,e,f,g,h</sub> for CY7C1474BV33). In addition, the address for the subsequent access (read, write, or deselect) is latched into the Address Register (provided the appropriate control signals are asserted).

On the next clock rise the data presented to DQ and DQP  $(DQ_{a,b,c,d}/DQP_{a,b,c,d}$  for CY7C1470BV33,  $DQ_{a,b}/DQP_{a,b}$  for CY7C1472BV33, and  $DQ_{a,b,c,d,e,f,g,h}/DQP_{a,b,c,d,e,f,g,h}$  for CY7C1474BV33) (or a subset for byte write operations, see "Partial Write Cycle Description" on page 12 for details) inputs is latched into the device and the write is complete.

The data written during the Write operation is controlled by  $\overline{BW}$ ( $\overline{BW}_{a,b,c,d}$  for CY7C1470BV33,  $\overline{BW}_{a,b}$  for CY7C1472BV33, and  $\overline{BW}_{a,b,c,d,e,f,g,h}$  for CY7C1474BV33) signals. The CY7C1470BV33, CY7C1472BV33, and CY7C1474BV33 provides Byte Write capability that is described in "Partial Write Cycle Description" on page 12. Asserting the Write Enable input (WE) with the selected BW input selectively writes to only the desired bytes. Bytes not selected during a Byte Write operation remain unaltered. A synchronous self timed write mechanism has been provided to simplify the write operations. Byte Write capability has been included to greatly simplify read, modify, or write sequences, which can be reduced to simple Byte Write operations.

Because the CY7C1470BV33, CY7C1472BV33, and CY7C1474BV33 are common IO devices, data must <u>not</u> be driven into the device while the outputs are active. The OE can be deasserted HIGH before presenting data to the DQ and DQP ( $DQ_{a,b,c,d}/DQP_{a,b,c,d}$  for CY7C1470BV33,  $DQ_{a,b}/DQP_{a,b}$  for CY7C1472BV33, and  $DQ_{a,b,c,d,e,f,g,h}/DQP_{a,b,c,d,e,f,g,h}$  for CY7C1474BV33) inputs. Doing so tri-states the output drivers. As a safety precaution, DQ and DQP ( $DQ_{a,b,c,d}/DQP_{a,b,c,d}$  for CY7C1470BV33,  $DQ_{a,b,c,d,e,f,g,h}$  for CY7C1470BV33,  $DQ_{a,b}/DQP_{a,b}$  for CY7C1470BV33,  $DQ_{a,b}/DQP_{a,b,c,d}$  for CY7C1470BV33,  $DQ_{a,b,c,d,e,f,g,h}$  for CY7C1472BV33, and  $DQ_{a,b,c,d,e,f,g,h}/DQP_{a,b,c,d,e,f,g,h}$  for CY7C1474BV33) are automatically tri-stated during the data portion of a write cycle, regardless of the state of OE.

#### **Burst Write Accesses**

The CY7C1470BV33, CY7C1472BV33, and CY7C1474BV33 has an on-chip burst counter that enables the user to supply a single address and conduct up to four write operations without reasserting the address inputs. ADV/LD must be driven LOW to load the initial address, <u>as</u> described in "Single Write Accesses" on page 9. When ADV/LD is driven HIGH on the subsequent



clock rise, the Chip Enables ( $\overline{CE}_1$ ,  $CE_2$ , and  $\overline{CE}_3$ ) and  $\overline{WE}$  inputs are ignored and the burst counter is incremented. The correct BW ( $\underline{BW}_{a,b,c,d}$  for CY7C1470BV33,  $\underline{BW}_{a,b}$  for CY7C1472V33, and  $\underline{BW}_{a,b,c,d,e,f,g,h}$  for CY7C1474BV33) inputs must be driven in each cycle of the burst write to write the correct bytes of data.

#### Sleep Mode

The ZZ input pin is an asynchronous input. Asserting ZZ places the SRAM in a power conservation "sleep" mode. Two clock cycles are required to enter into or exit from this "sleep" mode. While in this mode, data integrity is guaranteed. Accesses pending when entering the "sleep" mode are not considered valid nor is the completion of the operation guaranteed. The device must be deselected before entering the "sleep" mode. CE<sub>1</sub>, CE<sub>2</sub>, and CE<sub>3</sub>, must remain inactive for the duration of t<sub>ZZREC</sub> after the ZZ input returns LOW.

# Table 1. Interleaved Burst Address Table (MODE = Floating or V<sub>DD</sub>)

First Address	Second Address	Third Address	Fourth Address
A1,A0	A1,A0	A1,A0	A1,A0
00	01	10	11
01	00	11	10
10	11	00	01
11	10	01	00

Table 2. Linear Burst Address Table (MODE = GND)

First Address	Second Address	Third Address	Fourth Address
A1,A0	A1,A0	A1,A0	A1,A0
00	01	10	11
01	10	11	00
10	11	00	01
11	00	01	10

### **ZZ Mode Electrical Characteristics**

Parameter	Description	Test Conditions	Min	Max	Unit
I <sub>DDZZ</sub>	Sleep mode standby current	$ZZ \ge V_{DD} - 0.2V$		120	mA
t <sub>ZZS</sub>	Device operation to ZZ	$ZZ \ge V_{DD} - 0.2V$		2t <sub>CYC</sub>	ns
t <sub>ZZREC</sub>	ZZ recovery time	ZZ <u>&lt;</u> 0.2V	2t <sub>CYC</sub>		ns
t <sub>ZZI</sub>	ZZ active to sleep current	This parameter is sampled		2t <sub>CYC</sub>	ns
t <sub>RZZI</sub>	ZZ Inactive to exit sleep current	This parameter is sampled	0		ns



#### Table 3. Truth Table

The truth table for CY7C1470BV33, CY7C1472BV33, and CY7C1474BV33 follows.<sup>[1, 2, 3, 4, 5, 6, 7]</sup>

Operation	Address Used	CE	ZZ	ADV/LD	WE	BW <sub>x</sub>	OE	CEN	CLK	DQ
Deselect Cycle	None	Н	L	L	Х	Х	Х	L	L-H	Tri-State
Continue Deselect Cycle	None	Х	L	Н	Х	Х	Х	L	L-H	Tri-State
Read Cycle (Begin Burst)	External	L	L	L	Н	Х	L	L	L-H	Data Out (Q)
Read Cycle (Continue Burst)	Next	Х	L	Н	Х	Х	L	L	L-H	Data Out (Q)
NOP/Dummy Read (Begin Burst)	External	L	L	L	Н	Х	Н	L	L-H	Tri-State
Dummy Read (Continue Burst)	Next	Х	L	Н	Х	Х	Н	L	L-H	Tri-State
Write Cycle (Begin Burst)	External	L	L	L	L	L	Х	L	L-H	Data In (D)
Write Cycle (Continue Burst)	Next	Х	L	Н	Х	L	Х	L	L-H	Data In (D)
NOP/Write Abort (Begin Burst)	None	L	L	L	L	Н	Х	L	L-H	Tri-State
Write Abort (Continue Burst)	Next	Х	L	Н	Х	Н	Х	L	L-H	Tri-State
Ignore Clock Edge (Stall)	Current	Х	L	Х	Х	Х	Х	Н	L-H	-
Sleep Mode	None	Х	Н	Х	Х	Х	Х	Х	Х	Tri-State

Notes

- X = "Don't Care", H = Logic HIGH, L = Logic LOW, CE stands for ALL Chip Enables active. BWx = 0 signifies at least one Byte Write Select is active, BWx = Valid signifies that the desired byte write selects are asserted, see "Partial Write Cycle Description" on page 12 for details.
   Write is defined by WE and BW<sub>[a:d]</sub>. See "Partial Write Cycle Description" on page 12 for details.
   When a write cycle is detected, all IOs are tri-stated, even during Byte Writes.
   The DQ and DQP pins are controlled by the current cycle and the OE signal.
   CEN = H inserts wait states.

- 6. 7.
- Device powers up deselected with the IOs in a tri-state condition, regardless of  $\overline{OE}$ .  $\overline{OE}$  is asynchronous and is not sampled with the clock rise. It is masked internally during Write cycles. During a read cycle DQ<sub>s</sub> and DQP<sub>[a:d]</sub> = tri-state when  $\overline{OE}$  is inactive or when the device is deselected, and DQ<sub>s</sub> = data when  $\overline{OE}$  is active.



#### Table 4. Partial Write Cycle Description

The partial write cycle description for CY7C1470BV33, CY7C1472BV33, and CY7C1474BV33 follows.<sup>[1, 2, 3, 8]</sup>

Function (CY7C1470BV33)	WE	BWd	BWc	BWb	BWa
Read	н	Х	Х	Х	Х
Write – No bytes written	L	Н	Н	Н	Н
Write Byte a – (DQ <sub>a</sub> and DQP <sub>a</sub> )	L	Н	Н	Н	L
Write Byte b – (DQ <sub>b</sub> and DQP <sub>b</sub> )	L	Н	Н	L	Н
Write Bytes b, a	L	Н	Н	L	L
Write Byte $c - (DQ_c and DQP_c)$	L	Н	L	Н	Н
Write Bytes c, a	L	Н	L	Н	L
Write Bytes c, b	L	Н	L	L	Н
Write Bytes c, b, a	L	Н	L	L	L
Write Byte d – (DQ <sub>d</sub> and DQP <sub>d</sub> )	L	L	Н	Н	Н
Write Bytes d, a	L	L	Н	Н	L
Write Bytes d, b	L	L	Н	L	Н
Write Bytes d, b, a	L	L	Н	L	L
Write Bytes d, c	L	L	L	Н	Н
Write Bytes d, c, a	L	L	L	Н	L
Write Bytes d, c, b	L	L	L	L	Н
Write All Bytes	L	L	L	L	L

Function (CY7C1472BV33)	WE	BWb	BWa
Read	Н	х	х
Write – No Bytes Written	L	Н	Н
Write Byte a – $(DQ_a \text{ and } DQP_a)$	L	Н	L
Write Byte b – (DQ <sub>b</sub> and DQP <sub>b</sub> )	L	L	Н
Write Both Bytes	L	L	L

Function (CY7C1474BV33)	WE	BW <sub>x</sub>
Read	н	х
Write – No Bytes Written	L	Н
Write Byte X – $(DQ_x and DQP_x)$	L	L
Write All Bytes	L	All BW = L

Note 8. Table lists only a partial listing of the Byte Write combinations. Any combination of  $\overline{BW}_{[a:d]}$  is valid. Appropriate Write is based on which Byte Write is active.



### IEEE 1149.1 Serial Boundary Scan (JTAG)

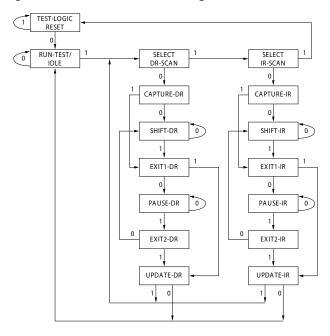
The CY7C1470BV33, CY7C1472BV33, and CY7C1474BV33 incorporates a serial boundary scan test access port (TAP). This port operates in accordance with IEEE Standard 1149.1-1990 but does not have the set of functions required for full 1149.1 compliance. These functions from the IEEE specification are excluded because their inclusion places an added delay in the critical speed path of the SRAM. Note that the TAP controller functions in a manner that does not conflict with the operation of other devices using 1149.1 fully compliant TAPs. The TAP operates using JEDEC-standard 3.3V or 2.5V IO logic levels.

The CY7C1470BV33, CY7C1472BV33, and CY7C1474BV33 contains a TAP controller, instruction register, boundary scan register, bypass register, and ID register.

#### **Disabling the JTAG Feature**

It is possible to operate the SRAM without using the JTAG feature. To disable the TAP controller, TCK must be tied LOW ( $V_{SS}$ ) to prevent clocking of the device. TDI and TMS are internally pulled up and may be unconnected. They may alternately be connected to  $V_{DD}$  through a pull up resistor. TDO must be left unconnected. During power up, the device comes up in a reset state, which does not interfere with the operation of the device.

#### Figure 2. TAP Controller State Diagram



The 0/1 next to each state represents the value of TMS at the rising edge of TCK.

#### Test Access Port (TAP)

#### Test Clock (TCK)

The test clock is used only with the TAP controller. All inputs are captured on the rising edge of TCK. All outputs are driven from the falling edge of TCK.

#### Test Mode Select (TMS)

The TMS input is used to give commands to the TAP controller and is sampled on the rising edge of TCK. It is allowable to leave this ball unconnected if the TAP is not used. The ball is pulled up internally, resulting in a logic HIGH level.

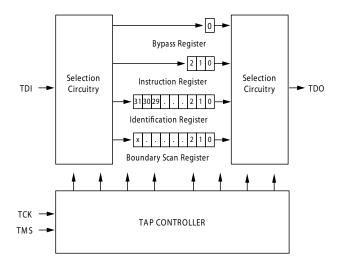
#### Test Data-In (TDI)

The TDI ball is used to serially input information into the registers and can be connected to the input of any of the registers. The register between TDI and TDO is chosen by the instruction that is loaded into the TAP instruction register. For information about loading the instruction register, see the TAP Controller State Diagram. TDI is internally pulled up and can be unconnected if the TAP is unused in an application. TDI is connected to the most significant bit (MSB) of any register. (See TAP Controller Block Diagram.)

#### Test Data-Out (TDO)

The TDO output ball is used to serially clock data-out from the registers. The output is active depending upon the current state of the TAP state machine. The output changes on the falling edge of TCK. TDO is connected to the least significant bit (LSB) of any register. (See TAP Controller State Diagram.)

#### Figure 3. TAP Controller Block Diagram



#### Performing a TAP Reset

A RESET is performed by forcing TMS HIGH ( $V_{DD}$ ) for five rising edges of TCK. This RESET does not affect the operation of the SRAM and may be performed while the SRAM is operating.

During power up, the TAP is reset internally to ensure that TDO comes up in a High Z state.

#### **TAP Registers**

Registers are connected between the TDI and TDO balls and scans data into and out of the SRAM test circuitry. Only one register can be selected at a time through the instruction register. Data is serially loaded into the TDI ball on the rising edge of TCK. Data is output on the TDO ball on the falling edge of TCK.



#### Instruction Register

Three bit instructions can be serially loaded into the instruction register. This register is loaded when it is placed between the TDI and TDO balls as shown in the "TAP Controller Block Diagram" on page 13. During power up, the instruction register is loaded with the IDCODE instruction. It is also loaded with the IDCODE instruction if the controller is placed in a reset state as described in the previous section.

When the TAP controller is in the Capture-IR state, the two least significant bits are loaded with a binary '01' pattern to enable fault isolation of the board-level serial test data path.

#### Bypass Register

To save time when serially shifting data through registers, it is sometimes advantageous to skip certain chips. The bypass register is a single bit register that can be placed between the TDI and TDO balls. This shifts data through the SRAM with minimal delay. The bypass register is set LOW ( $V_{SS}$ ) when the BYPASS instruction is executed.

#### Boundary Scan Register

The boundary scan register is connected to all the input and bidirectional balls on the SRAM.

The boundary scan register is loaded with the contents of the RAM IO ring when the TAP controller is in the Capture-DR state and is then placed between the TDI and TDO balls when the controller is moved to the Shift-DR state. The EXTEST, SAMPLE/PRELOAD and SAMPLE Z instructions can be used to capture the contents of the IO ring.

The Boundary Scan Order tables show the order in which the bits are connected. Each bit corresponds to one of the bumps on the SRAM package. The MSB of the register is connected to TDI and the LSB is connected to TDO.

#### Identification (ID) Register

The ID register is loaded with a vendor-specific, 32 bit code during the Capture-DR state when the IDCODE command is loaded in the instruction register. The IDCODE is hardwired into the SRAM and can be shifted out when the TAP controller is in the Shift-DR state. The ID register has a vendor code and other information described in "Identification Register Definitions" on page 18.

#### **TAP Instruction Set**

#### Overview

Eight different instructions are possible with the three bit instruction register. All combinations are listed in "Identification Codes" on page 18. Three of these instructions are listed as RESERVED and must not be used. The other five instructions are described in this section in detail.

The TAP controller used in this SRAM is not fully compliant to the 1149.1 convention because some of the mandatory 1149.1 instructions are not fully implemented.

The TAP controller cannot be used to load address data or control signals into the SRAM and cannot preload the IO buffers. The SRAM does not implement the 1149.1 commands EXTEST or INTEST or the PRELOAD portion of SAMPLE/PRELOAD; rather, it performs a capture of the IO ring when these instructions are executed. Instructions are loaded into the TAP controller during the Shift-IR state when the instruction register is placed between TDI and TDO. During this state, instructions are shifted through the instruction register through the TDI and TDO balls. To execute the instruction after it is shifted in, the TAP controller is moved into the Update-IR state.

#### EXTEST

EXTEST is a mandatory 1149.1 instruction which is executed whenever the instruction register is loaded with all 0s. EXTEST is not implemented in this SRAM TAP controller, and therefore this device is not compliant to 1149.1. The TAP controller does recognize an all-0 instruction.

When an EXTEST instruction is loaded into the instruction register, the SRAM responds as if a SAMPLE/PRELOAD instruction has been loaded. There is one difference between the two instructions. Unlike the SAMPLE/PRELOAD instruction, EXTEST places the SRAM outputs in a High Z state.

#### IDCODE

The IDCODE instruction loads a vendor-specific, 32 bit code into the instruction register. It also places the instruction register between the TDI and TDO balls and shifts the IDCODE out of the device when the TAP controller enters the Shift-DR state.

The IDCODE instruction is loaded into the instruction register during power up or whenever the TAP controller is in a test logic reset state.

#### SAMPLE Z

The SAMPLE Z instruction connects the boundary scan register between the TDI and TDO balls when the TAP controller is in a Shift-DR state. It also places all SRAM outputs into a High Z state.

#### SAMPLE/PRELOAD

SAMPLE/PRELOAD is a 1149.1 mandatory instruction. The PRELOAD portion of this instruction is not implemented, so the device TAP controller is not fully 1149.1 compliant.

When the SAMPLE/PRELOAD instruction is loaded into the instruction register and the TAP controller is in the Capture-DR state, a snapshot of data on the inputs and bidirectional balls is captured in the boundary scan register.

The user must be aware that the TAP controller clock can only operate at a frequency up to 20 MHz, while the SRAM clock operates more than an order of magnitude faster. Because there is a large difference in the clock frequencies, it is possible that during the Capture-DR state, an input or output may undergo a transition. The TAP may then try to capture a signal while in transition (metastable state). This does not harm the device, but there is no guarantee as to the value that is captured. Repeatable results may not be possible.

To guarantee that the boundary scan register captures the correct value of a signal, the SRAM signal must be stabilized long enough to meet the TAP controller's capture setup plus hold time ( $t_{CS}$  plus  $t_{CH}$ ).

The SRAM clock input might not be captured correctly if there is no way in a design to stop (or slow) the clock during a SAMPLE/PRELOAD instruction. If this is an issue, it is still



possible to capture all other signals and simply ignore the value of the CLK captured in the boundary scan register.

After the data is captured, it is possible to shift out the data by putting the TAP into the Shift-DR state. This places the boundary scan register between the TDI and TDO balls.

Note that since the PRELOAD part of the command is not implemented, putting the TAP to the Update-DR state while performing a SAMPLE/PRELOAD instruction has the same effect as the Pause-DR command.

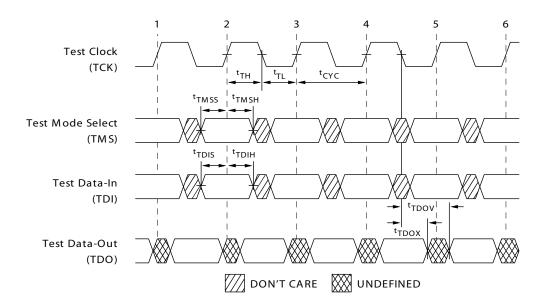
#### BYPASS

When the BYPASS instruction is loaded in the instruction register and the TAP is placed in a Shift-DR state, the bypass register is placed between the TDI and TDO balls. The advantage of the BYPASS instruction is that it shortens the boundary scan path when multiple devices are connected together on a board.

#### Reserved

These instructions are not implemented but are reserved for future use. Do not use these instructions.

#### Figure 4. TAP Timing





### **TAP AC Switching Characteristics**

Over the Operating Range<sup>[9, 10]</sup>

Parameter	Description	Min	Max	Unit
Clock	l			
t <sub>TCYC</sub>	TCK Clock Cycle Time	50		ns
t <sub>TF</sub>	TCK Clock Frequency		20	MHz
t <sub>TH</sub>	TCK Clock HIGH time	20		ns
t <sub>TL</sub>	TCK Clock LOW time	20		ns
Output Time	es	·		
t <sub>TDOV</sub>	TCK Clock LOW to TDO Valid		10	ns
t <sub>TDOX</sub>	TCK Clock LOW to TDO Invalid	0		ns
Setup Time	S	·		
t <sub>TMSS</sub>	TMS Setup to TCK Clock Rise	5		ns
t <sub>TDIS</sub>	TDI Setup to TCK Clock Rise	5		ns
t <sub>CS</sub>	Capture Setup to TCK Rise	5		ns
Hold Times	•			
t <sub>TMSH</sub>	TMS Hold after TCK Clock Rise	5		ns
t <sub>TDIH</sub>	TDI Hold after Clock Rise	5		ns
t <sub>CH</sub>	Capture Hold after Clock Rise	5		ns

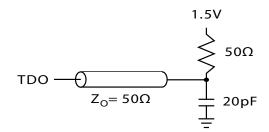
9. t<sub>CS</sub> and t<sub>CH</sub> refer to the setup and hold time requirements of latching data from the boundary scan register. 10. Test conditions are specified using the load in TAP AC Test Conditions. t<sub>R</sub>/t<sub>F</sub> = 1 ns.



### 3.3V TAP AC Test Conditions

Input pulse levels	V <sub>SS</sub> to 3.3V
Input rise and fall times	1 ns
Input timing reference levels	1.5V
Output reference levels	1.5V
Test load termination supply voltage .	1.5V

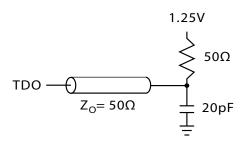
### 3.3V TAP AC Output Load Equivalent



### 2.5V TAP AC Test Conditions

Input pulse levels	V <sub>SS</sub> to 2.5V
Input rise and fall time	1 ns
Input timing reference levels	1.25V
Output reference levels	1.25V
Test load termination supply voltage	1.25V

### 2.5V TAP AC Output Load Equivalent



# TAP DC Electrical Characteristics And Operating Conditions $(0^{\circ}C < T_A < +70^{\circ}C; V_{DD} = 3.135V \text{ to } 3.6V \text{ unless otherwise noted})^{[11]}$

Parameter	Description	Test Conditions		Min	Max	Unit
V <sub>OH1</sub>	Output HIGH Voltage	$I_{OH} = -4.0 \text{ mA}, V_{DDQ} = 3.3 \text{V}$		2.4		V
		$I_{OH} = -1.0 \text{ mA}, V_{DDO}$	<sub>2</sub> = 2.5V	2.0		V
V <sub>OH2</sub>	Output HIGH Voltage	I <sub>OH</sub> = −100 μA	$V_{DDQ} = 3.3V$	2.9		V
			$V_{DDQ} = 2.5V$	2.1		V
V <sub>OL1</sub>	Output LOW Voltage	I <sub>OL</sub> = 8.0 mA	$V_{DDQ} = 3.3V$		0.4	V
		I <sub>OL</sub> = 1.0 mA	V <sub>DDQ</sub> = 2.5V		0.4	V
V <sub>OL2</sub>	Output LOW Voltage	I <sub>OL</sub> = 100 μA	$V_{DDQ} = 3.3V$		0.2	V
			$V_{DDQ} = 2.5V$		0.2	V
V <sub>IH</sub>	Input HIGH Voltage		$V_{DDQ} = 3.3V$	2.0	V <sub>DD</sub> + 0.3	V
			$V_{DDQ} = 2.5V$	1.7	V <sub>DD</sub> + 0.3	V
V <sub>IL</sub>	Input LOW Voltage		$V_{DDQ} = 3.3V$	-0.3	0.8	V
			$V_{DDQ} = 2.5V$	-0.3	0.7	V
I <sub>X</sub>	Input Load Current	$GND \leq V_{IN} \leq V_{DDQ}$		-5	5	μA



#### Table 5. Identification Register Definitions

Instruction Field	CY7C1470BV33 (2M x 36)	CY7C1472BV33 (4M x 18)	CY7C1474BV33 (1M x 72)	Description
Revision Number (31:29)	000	000	000	Describes the version number
Device Depth (28:24) <sup>[12]</sup>	01011	01011	01011	Reserved for internal use
Architecture/Memory Type(23:18)	001000	001000	001000	Defines memory type and archi- tecture
Bus Width/Density(17:12)	100100	010100	110100	Defines width and density
Cypress JEDEC ID Code (11:1)	00000110100	00000110100	00000110100	Enables unique identification of SRAM vendor
ID Register Presence Indicator (0)	1	1	1	Indicates the presence of an ID register

#### Table 6. Scan Register Sizes

Register Name	Bit Size (x36)	Bit Size (x18)	Bit Size (x72)
Instruction	3	3	3
Bypass	1	1	1
ID	32	32	32
Boundary Scan Order – 165 FBGA	71	52	-
Boundary Scan Order – 209 FBGA	-	-	110

#### Table 7. Identification Codes

Instruction	Code	Description
EXTEST	000	Captures IO ring contents. Places the boundary scan register between TDI and TDO. Forces all SRAM outputs to High Z state. This instruction is not 1149.1 compliant.
IDCODE	001	Loads the ID register with the vendor ID code and places the register between TDI and TDO. This operation does not affect SRAM operations.
SAMPLE Z	010	Captures IO ring contents. Places the boundary scan register between TDI and TDO. Forces all SRAM output drivers to a High Z state.
RESERVED	011	Do Not Use: This instruction is reserved for future use.
SAMPLE/PRELOAD	100	Captures IO ring contents. Places the boundary scan register between TDI and TDO. Does not affect SRAM operation. This instruction does not implement 1149.1 preload function and is therefore not 1149.1 compliant.
RESERVED	101	Do Not Use: This instruction is reserved for future use.



#### Table 8. Boundary Scan Exit Order (2M x 36)

Bit #	165-Ball ID	
1	C1	
2	D1	
3	E1	
4	D2	
5	E2	
6	F1	
7	G1	
8	F2	
9	G2	
10	J1	
11	K1	
12	L1	
13	J2	
14	M1	
15	N1	
16	K2	
17	L2	
18	M2	
19	R1	
20	R2	

Bit #	165-Ball ID
21	R3
22	P2
23	R4
24	P6
25	R6
26	R8
27	P3
28	P4
29	P8
30	P9
31	P10
32	R9
33	R10
34	R11
35	N11
36	M11
37	L11
38	M10
39	L10
40	K11

Bit #	165-Ball ID
41	J11
42	K10
43	J10
44	H11
45	G11
46	F11
47	E11
48	D10
49	D11
50	C11
51	G10
52	F10
53	E10
54	A9
55	B9
56	A10
57	B10
58	A8
59	B8
60	A7

Bit #	165-Ball ID
61	B7
62	B6
63	A6
64	B5
65	A5
66	A4
67	B4
68	B3
69	A3
70	A2
71	B2

#### Table 9. Boundary Scan Exit Order (4M x 18)

Bit #	165-Ball ID
1	D2
2	E2
3	F2
4	G2
5	J1
6	K1
7	L1
8	M1
9	N1
10	R1
11	R2
12	R3
13	P2

( )		
Bit #	165-Ball ID	
14	R4	
15	P6	
16	R6	
17	R8	
18	P3	
19	P4	
20	P8	
21	P9	
22	P10	
23	R9	
24	R10	
25	R11	
26	M10	

Bit #	165-Ball ID	
27	L10	
28	K10	
29	J10	
30	H11	
31	G11	
32	F11	
33	E11	
34	D11	
35	C11	
36	A11	
37	A9	
38	B9	
39	A10	

Bit # 165-Ball		
40	B10	
41	A8	
42	B8	
43	A7	
44	B7	
45	B6	
46	A6	
47	B5	
48	A4	
49	B3	
50	A3	
51	A2	
52	B2	



### Boundary Scan Exit Order (1M x 72)

Bit # 209-Ball ID		
1	A1	
2	A2	
3	B1	
4	B2	
5	C1	
6	C2	
7 D1		
8	D2	
9	E1	
10	E2	
11	F1	
12	F2	
13	G1	
14	G2	
15	H1	
16	H2	
17	J1	
18	J2	
19	L1	
20	L2	
21	M1	
22	M2	
23	N1	
24	N2	
25	P1	
26	P2	
27	R2	
28	R1	

	(2)
Bit #	209-Ball ID
29	T1
30	T2
31	U1
32	U2
33	V1
34	V2
35	W1
36	W2
37	T6
38	V3
39	V4
40	U4
41	W5
42	V6
43	W6
44	V5
45	U5
46	U6
47	W7
48	V7
49	U7
50	V8
51	V9
52	W11
53	W10
54	V11
55	V10
56	U11

57         U10           58         T11           59         T10           60         R11           61         R10           62         P11           63         P10           64         N11           65         N10           66         M11	
59         T10           60         R11           61         R10           62         P11           63         P10           64         N11           65         N10           66         M11	
60         R11           61         R10           62         P11           63         P10           64         N11           65         N10           66         M11	
61         R10           62         P11           63         P10           64         N11           65         N10           66         M11	
62         P11           63         P10           64         N11           65         N10           66         M11	
63         P10           64         N11           65         N10           66         M11	
64         N11           65         N10           66         M11	
65 N10 66 M11	
66 M11	
67 M10	
68 L11	
69 L10	
70 P6	
71 J11	
72 J10	
73 H11	
74 H10	
75 G11	
76 G10	
77 F11	
78 F10	
79 E10	
80 E11	
81 D11	
82 D10	
83 C11	
84 C10	

	209-Ball ID
85	B11
86	B10
87	A11
88	A10
89	A7
90	A5
91	A9
92	U8
93	A6
94	D6
95	K6
96	B6
97	K3
98	A8
99	B4
100	B3
101	C3
102	C4
103	C8
104	C9
105	B9
106	B8
107	A4
108	C6
109	B7
110	A3



### **Maximum Ratings**

Exceeding maximum ratings may impair the useful life of the device. These user guidelines are not tested.

Storage Temperature65°C to +150°C
Ambient Temperature with
Power Applied–55°C to +125°C
Supply Voltage on $V_{DD}$ Relative to GND–0.5V to +4.6V
Supply Voltage on V <sub>DDQ</sub> Relative to GND–0.5V to +V <sub>DD</sub>
DC to Outputs in Tri-State0.5V to $V_{DDQ}$ + 0.5V
DC Input Voltage –0.5V to $V_{DD}$ + 0.5V
Current into Outputs (LOW) 20 mA
Static Discharge Voltage > 2001V
(MIL-STD-883, Method 3015)
Latch Up Current > 200 mA

## **Neutron Soft Error Immunity**

Parameter	Description	Test Con- ditions	Тур	Max*	Unit
LSBU	Logical Single Bit Upsets	25°C	361	394	FIT/ Mb
LMBU	Logical Multi Bit Upsets	25°C	0	0.01	FIT/ Mb
SEL	Single Event Latch up	85°C	0	0.1	FIT/ Dev
* No LMBU or SEL events occurred during testing; this column represents a statistical $\chi^2$ , 95% confidence limit calculation. For more details refer to Application Note AN 54908 "Accelerated Neutron SER Testing and Calculation of Terrestrial Failure Rates"					

### **Operating Range**

Range	Ambient Temperature	V <sub>DD</sub>	V <sub>DDQ</sub>
Commercial	0°C to +70°C	3.3V -5%/+10%	
Industrial	-40°C to +85°C		to V <sub>DD</sub>

### **Electrical Characteristics**

Over the Operating Range<sup>[13, 14]</sup>

Parameter	Description	Test Conditions	Min	Мах	Unit
V <sub>DD</sub>	Power Supply Voltage		3.135	3.6	V
V <sub>DDQ</sub>	IO Supply Voltage	For 3.3V IO	3.135	V <sub>DD</sub>	V
		For 2.5V IO	2.375	2.625	V
V <sub>OH</sub>	Output HIGH Voltage	For 3.3V IO, I <sub>OH</sub> = -4.0 mA	2.4		V
		For 2.5V IO, I <sub>OH</sub> = -1.0 mA	2.0		V
V <sub>OL</sub>	Output LOW Voltage	For 3.3V IO, I <sub>OL</sub> = 8.0 mA		0.4	V
		For 2.5V IO, I <sub>OL</sub> = 1.0 mA		0.4	V
V <sub>IH</sub>	Input HIGH Voltage <sup>[13]</sup>	For 3.3V IO	2.0	V <sub>DD</sub> + 0.3V	V
	For 2.5V IO	1.7	V <sub>DD</sub> + 0.3V	V	
V <sub>IL</sub>	Input LOW Voltage <sup>[13]</sup>	For 3.3V IO	-0.3	0.8	V
		For 2.5V IO	-0.3	0.7	V
Ι <sub>X</sub>	Input Leakage Current except ZZ and MODE	$GND \le V_I \le V_{DDQ}$	-5	5	μΑ
	Input Current of MODE	Input = V <sub>SS</sub>	-30		μΑ
		Input = V <sub>DD</sub>		5	μΑ
	Input Current of ZZ	Input = V <sub>SS</sub>	-5		μΑ
		Input = V <sub>DD</sub>		30	μΑ
I <sub>OZ</sub>	Output Leakage Current	$GND \le V_I \le V_{DDQ}$ , Output Disabled	-5	5	μΑ

#### Notes

13. Overshoot:  $V_{IH}(AC) < V_{DD} + 1.5V$  (pulse width less than  $t_{CYC}/2$ ). Undershoot:  $V_{IL}(AC) > -2V$  (pulse width less than  $t_{CYC}/2$ ). 14.  $T_{power up}$ : assumes a linear ramp from 0V to  $V_{DD}$  (min.) within 200 ms. During this time  $V_{IH} < V_{DD}$  and  $V_{DDQ} \le V_{DD}$ .



### **Electrical Characteristics**

Over the Operating Range<sup>[13, 14]</sup> (continued)

Parameter	Description	escription Test Conditions			Max	Unit
I <sub>DD</sub> <sup>[15]</sup>	V <sub>DD</sub> Operating Supply	$V_{DD} = Max., I_{OUT} = 0 mA,$	4.0-ns cycle, 250 MHz		500	mA
		$f = f_{MAX} = 1/t_{CYC}$	5.0-ns cycle, 200 MHz		500	mA
			6.0-ns cycle, 167 MHz		450	mA
I <sub>SB1</sub>	Automatic CE	Max. V <sub>DD</sub> , Device Deselected,	4.0-ns cycle, 250 MHz		245	mA
	Power Down Current—TTL Inputs	$V_{IN} \ge V_{IH} \text{ or } V_{IN} \le V_{IL},$ f = f <sub>MAX</sub> = 1/t <sub>CYC</sub>	5.0-ns cycle, 200 MHz		245	mA
			6.0-ns cycle, 167 MHz		245	mA
I <sub>SB2</sub>	Automatic CE Power Down Current—CMOS Inputs	Max. V <sub>DD</sub> , Device Deselected, $V_{IN} \le 0.3V$ or $V_{IN} \ge V_{DDQ} - 0.3V$ , f = 0	All speed grades		120	mA
I <sub>SB3</sub>	Automatic CE Power Down Current—CMOS Inputs	Max. V <sub>DD</sub> , Device Deselected,	4.0-ns cycle, 250 MHz		245	mA
		$V_{IN} \le 0.3V$ or $V_{IN} \ge V_{DDQ} - 0.3V$ , f = $f_{MAX} = 1/t_{CYC}$	5.0-ns cycle, 200 MHz		245	mA
		MAXCTC	6.0-ns cycle, 167 MHz		245	mA
I <sub>SB4</sub>	Automatic CE Power Down Current—TTL Inputs	$ \begin{array}{l} \text{Max. } V_{DD} \text{, Device Deselected,} \\ V_{IN} \geq V_{IH} \text{ or } V_{IN} \leq V_{IL} \text{, } f = 0 \end{array} $	All speed grades		135	mA



### Capacitance

Tested initially and after any design or process changes that may affect these parameters.

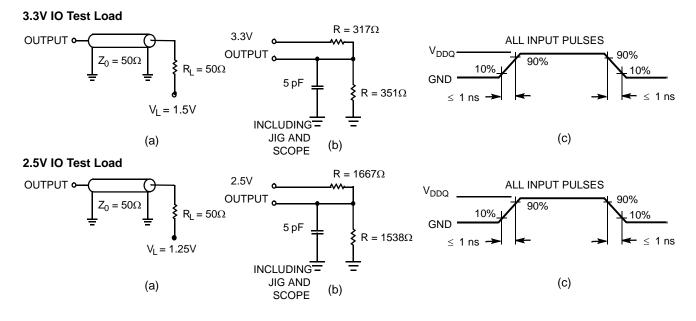
Parameter	Description	Test Conditions	100 TQFP Max	165 FBGA Max	209 FBGA Max	Unit
C <sub>ADDRESS</sub>	Address Input Capacitance	$T_{A} = 25^{\circ}C, f = 1 MHz,$	6	6	6	pF
C <sub>DATA</sub>	Data Input Capacitance	V <sub>DD</sub> = 3.3V V <sub>DDQ</sub> = 2.5V	5	5	5	pF
C <sub>CTRL</sub>	Control Input Capacitance	• DDQ = 2.5 •	8	8	8	pF
C <sub>CLK</sub>	Clock Input Capacitance		6	6	6	pF
C <sub>IO</sub>	Input/Output Capacitance		5	5	5	pF

### Thermal Resistance

Tested initially and after any design or process changes that may affect these parameters.

Parameters	Description	Test Conditions	100 TQFP Package	165 FBGA Package	209 FBGA Package	Unit
$\Theta_{JA}$	· · · · · · · · · · · · · · · · · · ·	Test conditions follow standard test methods and procedures for	24.63	16.3	15.2	°C/W
Θ <sub>JC</sub>	Thermal Resistance (Junction to Case)	measuring thermal impedance, per EIA/JESD51.	2.28	2.1	1.7	°C/W

### AC Test Loads and Waveforms





### **Switching Characteristics**

Over the Operating Range. Timing reference is 1.5V when  $V_{DDQ} = 3.3V$  and is 1.25V when  $V_{DDQ} = 2.5V$ . Test conditions shown in (a) of "AC Test Loads and Waveforms" on page 23 unless otherwise noted.

Denemeter	Description	-250		-200		-167		l lm it
Parameter	Description		Max	Min	Max	Min	Max	Unit
t <sub>Power</sub> <sup>[16]</sup>	V <sub>CC</sub> (typical) to the First Access Read or Write			1		1		ms
Clock								
t <sub>CYC</sub>	Clock Cycle Time	4.0		5.0		6.0		ns
F <sub>MAX</sub>	Maximum Operating Frequency		250		200		167	MHz
t <sub>CH</sub>	Clock HIGH	2.0		2.0		2.2		ns
t <sub>CL</sub>	Clock LOW	2.0		2.0		2.2		ns
Output Times								
t <sub>CO</sub>	Data Output Valid After CLK Rise		3.0		3.0		3.4	ns
t <sub>OEV</sub>	OE LOW to Output Valid		3.0		3.0		3.4	ns
t <sub>DOH</sub>	Data Output Hold After CLK Rise	1.3		1.3		1.5		ns
t <sub>CHZ</sub>	Clock to High Z <sup>[17, 18, 19]</sup>		3.0		3.0		3.4	ns
t <sub>CLZ</sub>	Clock to Low Z <sup>[17, 18, 19]</sup>	1.3		1.3		1.5		ns
t <sub>EOHZ</sub>	OE HIGH to Output High Z <sup>[17, 18, 19]</sup>		3.0		3.0		3.4	ns
t <sub>EOLZ</sub>	OE LOW to Output Low Z <sup>[17, 18, 19]</sup>			0		0		ns
Setup Times								
t <sub>AS</sub>	Address Setup Before CLK Rise	1.4		1.4		1.5		ns
t <sub>DS</sub>	Data Input Setup Before CLK Rise	1.4		1.4		1.5		ns
t <sub>CENS</sub>	CEN Setup Before CLK Rise	1.4		1.4		1.5		ns
t <sub>WES</sub>	WE, BW <sub>x</sub> Setup Before CLK Rise	1.4		1.4		1.5		ns
t <sub>ALS</sub>	ADV/LD Setup Before CLK Rise	1.4		1.4		1.5		ns
t <sub>CES</sub>	Chip Select Setup	1.4		1.4		1.5		ns
Hold Times								
t <sub>AH</sub>	Address Hold After CLK Rise	0.4		0.4		0.5		ns
t <sub>DH</sub>	Data Input Hold After CLK Rise	0.4		0.4		0.5		ns
t <sub>CENH</sub>	CEN Hold After CLK Rise	0.4		0.4		0.5		ns
t <sub>WEH</sub>	WE, BW <sub>x</sub> Hold After CLK Rise	0.4		0.4		0.5		ns
t <sub>ALH</sub>	ADV/LD Hold after CLK Rise	0.4		0.4		0.5		ns
t <sub>CEH</sub>	Chip Select Hold After CLK Rise	0.4		0.4	Ī	0.5	1	ns

#### Notes

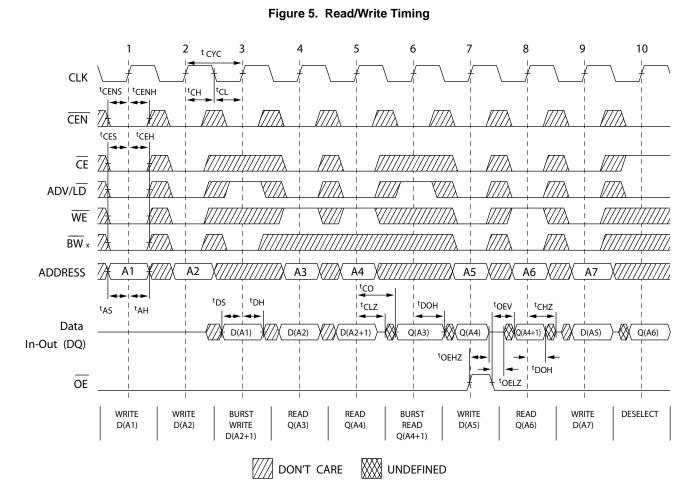
16. This part has an internal voltage regulator; t<sub>power</sub> is the time power is supplied above V<sub>DD</sub> minimum initially, before a read or write operation can be initiated.
 17. t<sub>CHZ</sub>, t<sub>CLZ</sub>, t<sub>EOLZ</sub>, and t<sub>EOHZ</sub> are specified with AC test conditions shown in (b) of "AC Test Loads and Waveforms" on page 23. Transition is measured ±200 mV from steady-state voltage.

At any voltage and temperature, t<sub>EOHZ</sub> is less than t<sub>EOLZ</sub> and t<sub>CHZ</sub> is less than t<sub>CLZ</sub> to eliminate bus contention between SRAMs when sharing the same data bus. These specifications do not imply a bus contention condition, but reflect parameters guaranteed over worst case user conditions. Device is designed to achieve High Z before Low Z under the same system conditions.
 This parameter is sampled and not 100% tested.



### **Switching Waveforms**

Figure 5 shows read-write timing waveform.<sup>[20, 21, 22]</sup>



Notes

20. For this waveform ZZ is tied LOW.

21. When  $\overline{CE}$  is LOW,  $\overline{CE}_1$  is LOW,  $CE_2$  is HIGH, and  $\overline{CE}_3$  is LOW. When  $\overline{CE}$  is HIGH,  $\overline{CE}_1$  is HIGH,  $CE_2$  is LOW or  $\overline{CE}_3$  is HIGH.

22. Order of the Burst sequence is determined by the status of the MODE (0 = Linear, 1= Interleaved). Burst operations are optional.

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### Switching Waveforms (continued)

Figure 6 shows NOP, STALL and DESELECT Cycles waveform.<sup>[20, 21, 23]</sup>

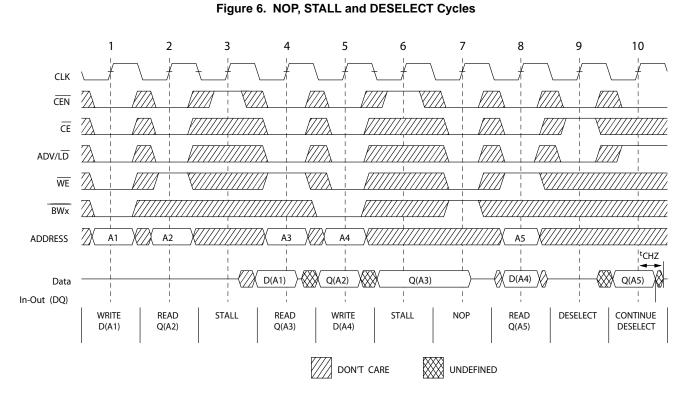
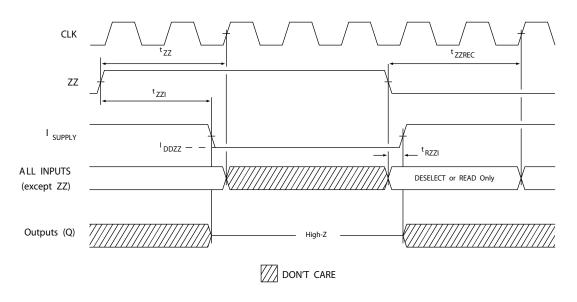


Figure 7 shows ZZ Mode timing waveform.<sup>[24, 25]</sup>

Figure 7. ZZ Mode Timing



#### Notes

23. The IGNORE CLOCK EDGE or STALL cycle (Clock 3) illustrated CEN being used to create a pause. A Write is not performed during this cycle. 24. Device must be deselected when entering ZZ mode. See "Truth Table" on page 11 for all possible signal conditions to deselect the device.

<sup>25.</sup> IOs are in High Z when exiting ZZ sleep mode.



### **Ordering Information**

The table below contains only the parts that are currently available. If you don't see what you are looking for, please contact your local sales representative. For more information, visit the Cypress website at www.cypress.com and refer to the product summary page at http://www.cypress.com/products

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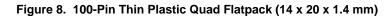
#### Table 10. Ordering Information

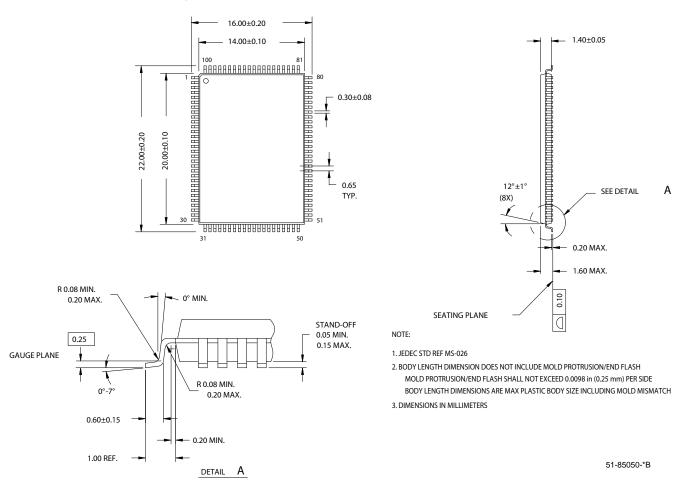
Speed (MHz)	Ordering Code	Package Diagram	Part and Package Type	Operating Range
167	CY7C1470BV33-167AXC	51-85050	100-pin Thin Quad Flat Pack (14 x 20 x 1.4 mm) Pb-free	Commercial
	CY7C1470BV33-167BZC	51-85165	165-ball Fine-Pitch Ball Grid Array (15 x 17 x 1.4 mm)	
	CY7C1472BV33-167BZC			
	CY7C1470BV33-167BZXC	51-85165	165-ball Fine-Pitch Ball Grid Array (15 x 17 x 1.4 mm) Pb-free	
	CY7C1474BV33-167BGC	51-85167	209-ball Fine-Pitch Ball Grid Array (14 x 22 x 1.76 mm)	
	CY7C1470BV33-167AXI	51-85050	100-Pin Thin Quad Flat Pack (14 x 20 x 1.4 mm) Pb-free	Industrial
	CY7C1472BV33-167AXI			
	CY7C1470BV33-167BZI	51-85165	165-ball Fine-Pitch Ball Grid Array (15 x 17 x 1.4 mm)	
	CY7C1470BV33-167BZXI	51-85165	165-ball Fine-Pitch Ball Grid Array (15 x 17 x 1.4 mm) Pb-free	
	CY7C1474BV33-167BGI	51-85167	209-ball Fine-Pitch Ball Grid Array (14 x 22 x 1.76 mm)	
200	CY7C1470BV33-200AXC	51-85050	100-pin Thin Quad Flat Pack (14 x 20 x 1.4 mm) Pb-free	Commercial
	CY7C1472BV33-200AXC			
	CY7C1470BV33-200BZC	51-85165	165-ball Fine-Pitch Ball Grid Array (15 x 17 x 1.4 mm)	
	CY7C1472BV33-200BZC			
	CY7C1472BV33-200BZXC	51-85165	165-ball Fine-Pitch Ball Grid Array (15 x 17 x 1.4 mm) Pb-free	
	CY7C1474BV33-200BGC	51-85167	209-ball Fine-Pitch Ball Grid Array (14 x 22 x 1.76 mm)	
	CY7C1474BV33-200BGXC		209-ball Fine-Pitch Ball Grid Array (14 x 22 x 1.76 mm) Pb-free	
	CY7C1470BV33-200BZI	51-85165	165-ball Fine-Pitch Ball Grid Array (15 x 17 x 1.4 mm)	Industrial
	CY7C1470BV33-200BZXI	51-85165	165-ball Fine-Pitch Ball Grid Array (15 x 17 x 1.4 mm) Pb-free	
	CY7C1474BV33-200BGI	51-85167	209-ball Fine-Pitch Ball Grid Array (14 x 22 x 1.76 mm)	



### CY7C1470BV33 CY7C1472BV33, CY7C1474BV33

### **Package Diagrams**

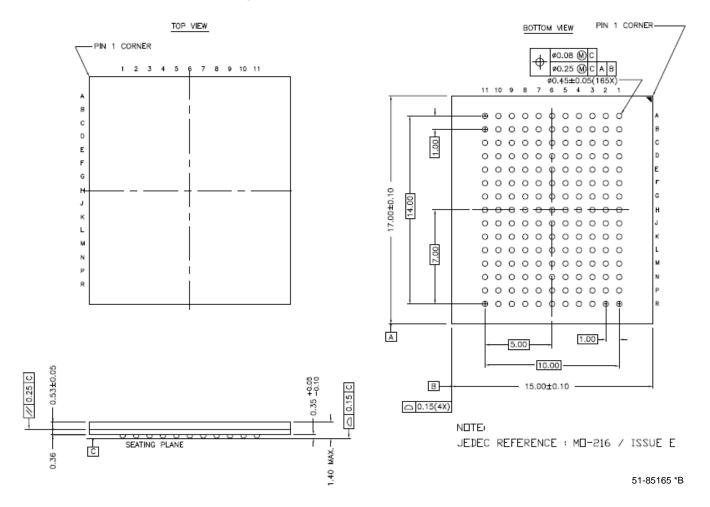






### Package Diagrams (continued)

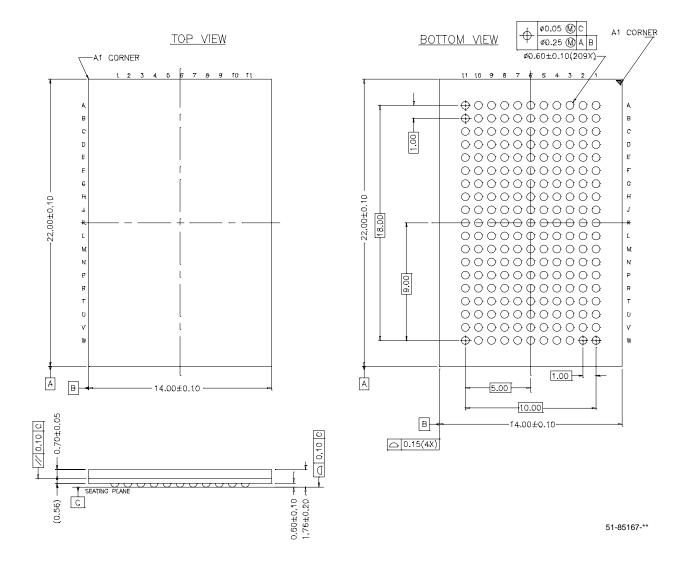
Figure 9. 165-Ball FBGA (15 x 17 x 1.4 mm)





### Package Diagrams (continued)







### **Document History Page**

# Document Title: CY7C1470BV33/CY7C1472BV33/CY7C1474BV33, 72 Mbit (2M x 36/4M x 18/1M x 72) Pipelined SRAM with NoBL™ Architecture Document Number: 001-15031

Documen		01 10001		
Revision	ECN	Orig. of Change	Submission Date	Description of Change
**	1032642	VKN/KKVTMP	See ECN	New Datasheet
*A	1897447	VKN/AESA	See ECN	Added footnote 15 related to IDD
*В	2082487	VKN	See ECN	Converted from preliminary to final
*C	2159486	VKN/PYRS	See ECN	Minor Change-Moved to the external web
*D	2755901	VKN	08/25/09	Included Soft Error Immunity Data Modified Ordering Information table by including parts that are available and modified the disclaimer for the Ordering information. Updated Package Diagram for spec 51-85165.

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