# CS3844B

# Current Mode PWM Control Circuit with 50% Max Duty Cycle

## Description

The CS3844B provides all the necessary features to implement off-line fixed frequency current-mode control with a minimum number of external components.

The CS3844B incorporates a precision temperature-controlled oscillator to minimize variations in frequency. An internal toggle flip-flop, which blanks the output off every other clock cycle, limits the duty-cycle range to less than 50%. An undervoltage lockout ensures that $V_{REF}$ is stabilized within specification before the output stage is enabled. **The CS3844B has been optimized for lower start up current (600µA max).**

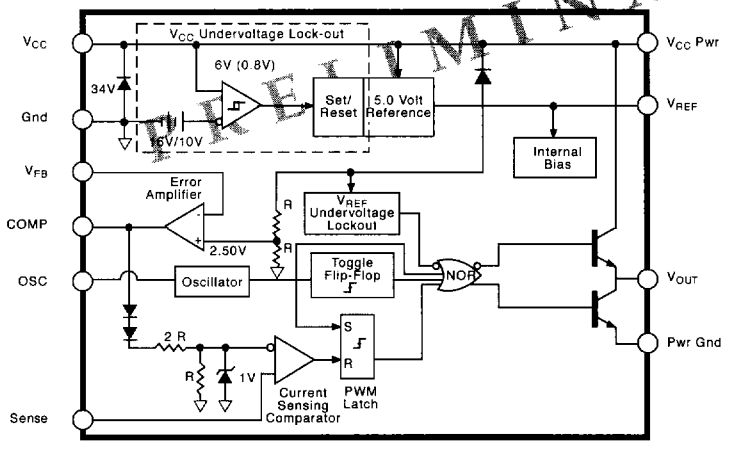
Other features include 1% trimmed band gap reference, pulse-by-pulse current limiting, and a high-current totem pole output for driving capacitive loads.



## Absolute Maximum Ratings

Supply Voltage ($I_{CC}$<30mA) ..... Self Limiting
Supply Voltage (Low Impedance Source) ..... 30V
Output Current ..... ±1A
Output Energy (Capacitive Load) ..... 5µJ
Analog Inputs ($V_{FB}$, $V_{Sense}$) ..... -0.3V to 5.5V
Error Amp Output Sink Current ..... 10mA
Lead Temperature Soldering
Wave Solder (through hole styles only) ..... 10 sec. max, 260°C peak
Reflow (SMD styles only) ..... 60 sec. max above 183°C, 230°C peak

## Block Diagram

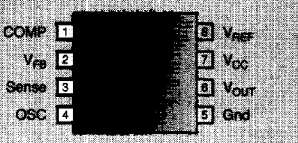



## Features

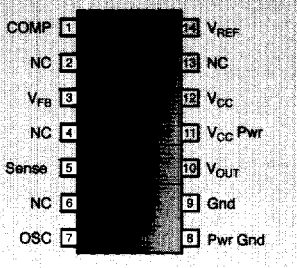
- Low Start Up Current (600µA max)
- Trimmed Oscillator Discharge Current for Precise Duty Cycle Control
- Current Mode Operation to 250kHz
- Automatic Feed Forward Compensation
- Latching PWM for Cycle-by-Cycle Current Limiting
- Internally Trimmed Bandgap Reference
- Undervoltage Lockout with Hysteresis
- High Current Totem Pole Output

## Package Options

**8 Lead PDIP & SO**




**14 Lead SO Narrow**

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Electrical Characteristics: $0° < T_A < 70°C$, $V_{CC}$ = 15V (Note 1); $R_T$ = 10kΩ, $C_T$ = 3.3nF for sawtooth mode, unless otherwise stated.

| PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNITS |
|---|---|---|---|---|---|
| **■ Reference Section** | | | | | |
| Output Voltage | $T_J$ = 25°C, $I_{REF}$ = 1mA | 4.90 | 5.00 | 5.10 | V |
| Line Regulation | 12 ≤ $V_{CC}$ ≤ 25V | | 6 | 20 | mV |
| Load Regulation | 1 ≤ $I_{REF}$ ≤ 20mA | | 6 | 25 | mV |
| Temperature Stability | (Note 1) | | 0.2 | 0.4 | mV/°C |
| Total Output Variation | Line, Load, Temp. (Note 1) | 4.82 | | 5.18 | V |
| Output Noise Voltage | 10Hz ≤ f ≤ 10kHz, $T_J$ = 25°C (Note 1) | | 50 | | µV |
| Long Term Stability | $T_A$ = 125°C, 1000 Hrs. (Note 1) | | 5 | 25 | mV |
| Output Short Circuit | $T_A$ = 25°C | -30 | -100 | -180 | mA |
| **■ Oscillator Section** | | | | | |
| Initial Accuracy | Sawtooth Mode, $T_J$ = 25°C | 47 | 52 | 57 | kHz |
| Voltage Stability | 12 ≤ $V_{CC}$ ≤ 25V | | 0.2 | 1.0 | % |
| Temperature Stability | Sawtooth Mode $T_{MIN}$ ≤ $T_A$ ≤ $T_{MAX}$ | | 5 | | % |
| Amplitude | $V_{OSC}$ (peak to peak) | | 1.7 | | V |
| **■ Error Amp Section** | | | | | |
| Input Voltage | $V_{COMP}$ = 2.5V | 2.42 | 2.50 | 2.58 | V |
| Input Bias Current | $V_{FB}$ = 0V | | -0.3 | -2.0 | µA |
| AVOL | 2 ≤ $V_{OUT}$ ≤ 4V | 65 | 90 | | dB |
| Unity Gain Bandwidth | (Note 2) | 0.7 | 1.0 | | MHz |
| PSRR | 12 ≤ $V_{CC}$ ≤ 25V | 60 | 70 | | dB |
| Output Sink Current | $V_{FB}$ = 2.7V, $V_{COMP}$ = 1.1V | 2 | 6 | | mA |
| Output Source Current | $V_{FB}$ = 2.3V, $V_{COMP}$ = 5V | -0.5 | -0.8 | | mA |
| $V_{OUT}$ HIGH | $V_{FB}$ = 2.3V, $R_L$ 15kΩ to Gnd | 5 | 6 | | V |
| $V_{OUT}$ LOW | $V_{FB}$ = 2.7V, $R_L$ = 15kΩ to $V_{REF}$ | | 0.7 | 1.1 | V |
| **■ Current Sense Section** | | | | | |
| Gain | (Notes 2&3) | 2.85 | 3.00 | 3.15 | V/V |
| Maximum Input Signal | $V_{COMP}$ = 5V (Note 2) | 0.9 | 1.0 | 1.1 | V |
| PSRR | 12 ≤ $V_{CC}$ ≤ 25V (Note 2) | | 70 | | dB |
| Input Bias Current | $V_{SENSE}$ = 0V | | -2 | -10 | µA |
| Delay to Output | $T_J$ = 25°C (Note 1) | | 150 | 300 | ns |
| **■ Output Section** | | | | | |
| Output Low Level | $I_{SINK}$ = 20mA | | 0.1 | 0.4 | V |
| | $I_{SINK}$ = 200mA | | 1.5 | 2.2 | V |
| Output High Level | $I_{SOURCE}$ = 20mA | 13.0 | 13.5 | | V |
| | $I_{SOURCE}$ = 200mA | 12.0 | 13.5 | | V |
| Rise Time | $T_J$ = 25°C, $C_L$ = 1nF (Note 2) | | 50 | 150 | ns |
| Fall Time | $T_J$ = 25°C, $C_L$ = 1nF (Note 2) | | 50 | 150 | ns |

## Electrical Characteristics: continued

| PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNITS |
|---|---|---|---|---|---|
| **■ Total Standby Current** | | | | | |
| Start-Up Current | | | 400 | 600 | µA |
| Operating Supply Current | $V_{FB} = V_{Sense} = 0V$, $R_T = 10k\Omega$, $C_T = 3.3nF$ | | 11 | 17 | mA |
| $V_{CC}$ Zener Voltage | $I_{CC} = 25mA$ | | 34 | | V |
| **■ PWM Section** | | | | | |
| Maximum Duty Cycle | | 46 | 48 | 50 | % |
| Minimum Duty Cycle | | | | 0 | % |
| **■ Under-Voltage Lockout Section** | | | | | |
| Start Threshold | | 14.5 | 16.0 | 17.5 | V |
| Min. Operating Voltage | After Turn On | 8.5 | 10.0 | 11.5 | V |



**Notes:** 1. These parameters, although guaranteed, are not 100% tested in production.

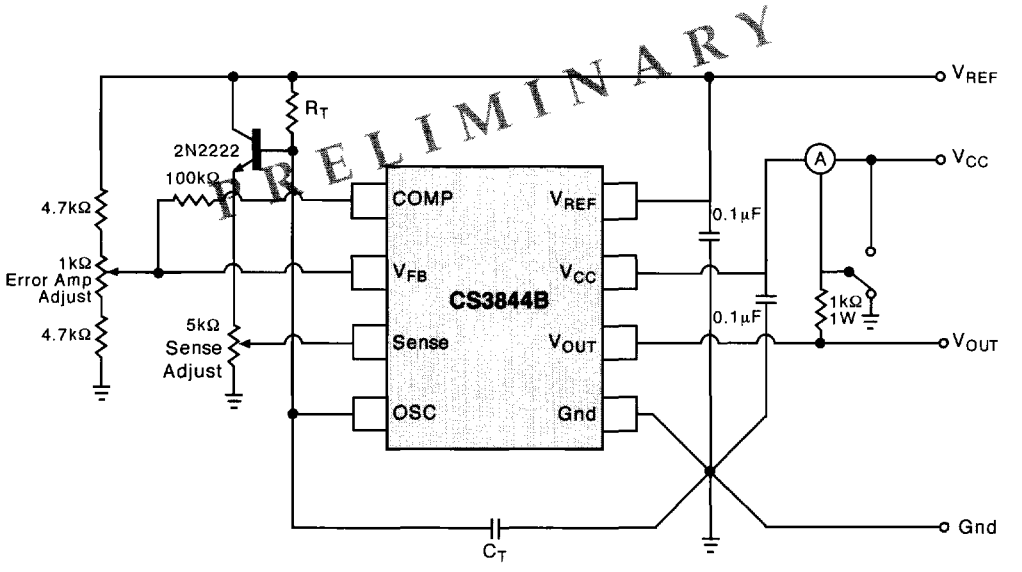
2. Parameter measured at trip point of latch with $V_{FB}$=0

3. Gain defined as: $A = \frac{\Delta V_{COMP}}{\Delta V_{Sense}}$; $0 \le V_{Sense} \le 0.8V$.

## Package Pin Description

| PACKAGE PIN # | | PIN SYMBOL | FUNCTION |
|---|---|---|---|
| 8L PDIP & SO | 14L SO | | |
| 1 | 1 | COMP | Error amp output, used to compensate error amplifier |
| 2 | 3 | $V_{FB}$ | Error amp inverting input |
| 3 | 5 | Sense | Noninverting input to Current Sense Comparator |
| 4 | 7 | OSC | Oscillator timing network with capacitor to Gnd, resistor to $V_{REF}$ |
| 5 | 9 | Gnd | Ground |
| 6 | 10 | $V_{OUT}$ | Output drive pin |
| 7 | 12 | $V_{CC}$ | Positive power supply |
| 8 | 14 | $V_{REF}$ | Output of 5V internal reference |
| | 8 | Pwr Gnd | Output driver Gnd |
| | 11 | $V_{CC}$ Pwr | Output driver positive supply |
| | 2, 4, 6, 13 | NC | No Connection |

## Test Circuit




## Circuit Description

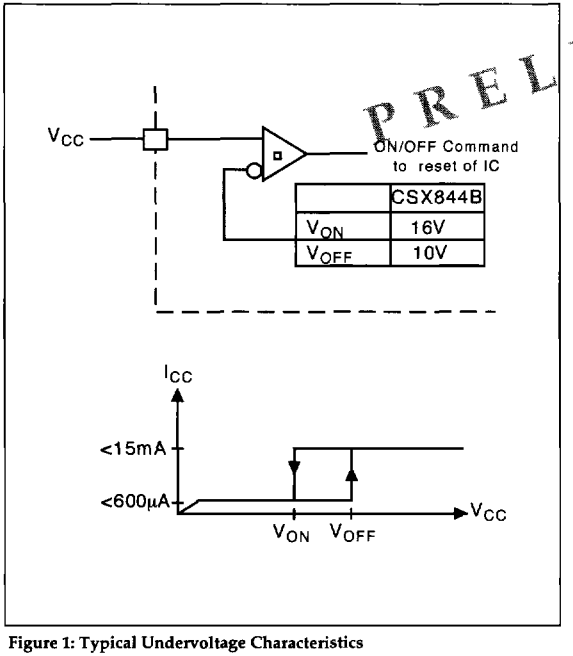



**Figure 1: Typical Undervoltage Characteristics**

### Undervoltage Lockout

During Undervoltage Lockout (Figure 1), the output driver is biased to sink minor amounts of current. The output should be shunted to ground with a resistor to prevent activating the power switch with extraneous leakage currents.

### PWM Waveform

To generate the PWM waveform, the control voltage from the error amplifier is compared to a current sense signal which represents the peak output inductor current (Figure 2). An increase in $V_{CC}$ causes the inductor current slope to increase, thus reducing the duty cycle. This is an inherent feed-forward characteristic of current mode control, since the control voltage does not have to change during changes of input supply voltage.

When the power supply sees a sudden large output current increase, the control voltage will increase allowing the duty cycle to momentarily increase. Since the duty cycle tends to exceed the maximum allowed to prevent transformer saturation in some power supplies, the internal oscillator waveform provides the maximum duty cycle clamp as programmed by the selection of OSC components.

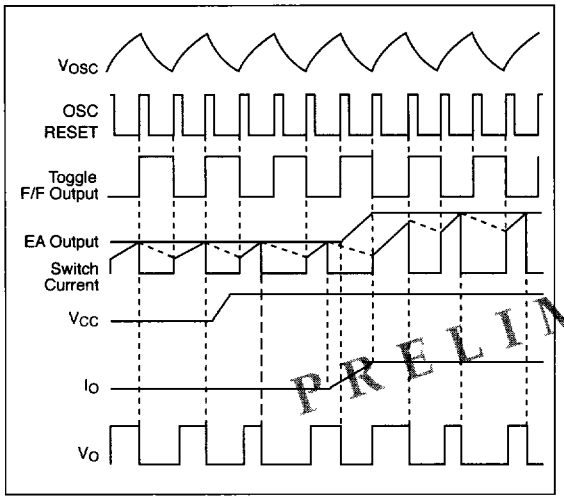



**Figure 2: Timing Diagram**

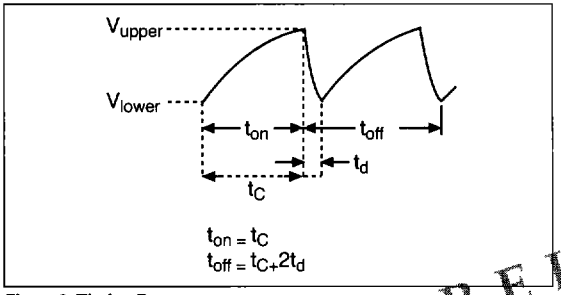



**Figure 3: Timing Parameters**

### Setting the Oscillator

The values for $t_c$ and $t_d$ are determined from the formulas:

$$t_c = R_T C_T \ln\left(\frac{V_{REF} - V_{lower}}{V_{REF} - V_{upper}}\right)$$

$$t_d = R_T C_T \ln\left(\frac{V_{REF} - I_d R_T - V_{lower}}{V_{REF} - I_d R_T - V_{upper}}\right)$$

Substituting in typical values for the parameters in:

$V_{REF} = 5.0V$, $V_{upper} = 2.7V$, $V_{lower} = 1.0V$, $I_d = 8.3mA$,

then

$$t_c \approx 0.5534 R_T C_T$$

$$t_d = R_T C_T \ln\left(\frac{2.3 - 0.0083\ R_T}{4.0 - 0.0083\ R_T}\right)$$

For better accuracy $R_T$ should be ≥10kΩ.

### Grounding

High peak currents associated with capacitive loads necessitate careful grounding techniques. Timing and bypass capacitors should be connected close to Gnd in a single point ground.

The transistor and 5kΩ potentiometer are used to sample the oscillator waveform and apply an adjustable ramp to Sense.

## Package Specification

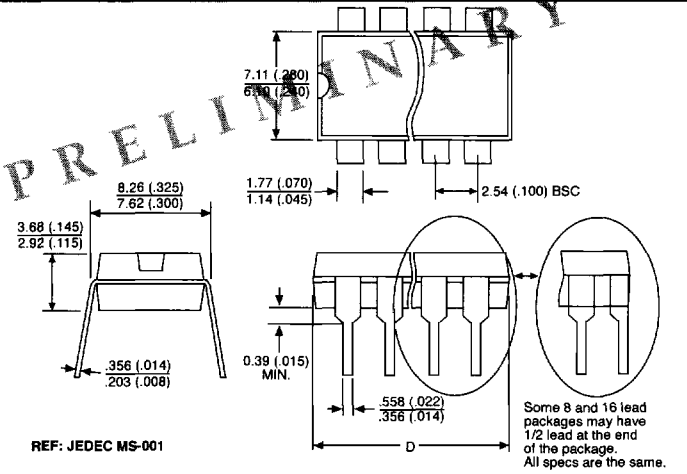
### PACKAGE DIMENSIONS IN mm (INCHES)

| Lead Count | D | | | |
|---|---|---|---|---|
| | Metric | | English | |
| | Max | Min | Max | Min |
| 8L PDIP | 10.16 | 9.02 | .400 | .355 |
| 8L SO Narrow | 5.00 | 4.80 | .197 | .189 |
| 14L SO Narrow | 8.75 | 8.55 | .344 | .337 |

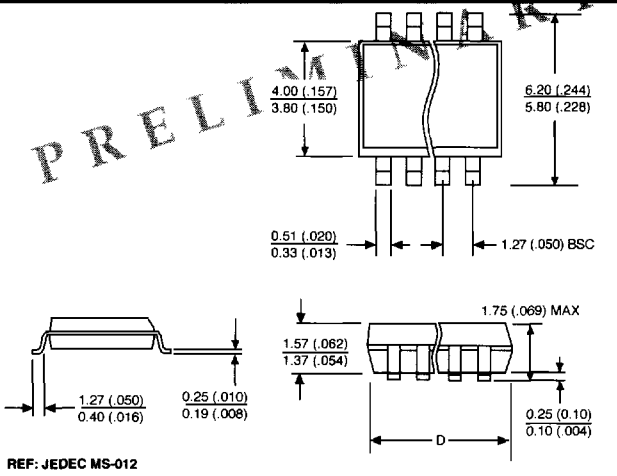
### PACKAGE THERMAL DATA

| Thermal Data | | 8 L PDIP | 8L SO | 14 L SO | |
|---|---|---|---|---|---|
| $RQ_{JC}$ | typ | 52 | 45 | 30 | °C/W |
| $RQ_{JA}$ | typ | 100 | 165 | 125 | °C/W |

### Plastic DIP (N); 300 mil wide




### Surface Mount Narrow Body (D); 150 mil wide




### Ordering Information

| Part Number | Description |
|---|---|
| CS3844BGN8 | 8L PDIP |
| CS3844BGD8 | 8L SO |
| CS3844BGDR8 | 8L SO *(tape & reel)* |
| CS3844BGD14 | 14L SO |
| CS3844BGDR14 | 14L SO *(tape & reel)* |

### Preliminary

This product is in the preproduction stages of the design process. The data sheet contains preliminary data. Cherry Semiconductor Corporation reserves the right to make changes to the specifications without notice. Please contact Cherry Semiconductor Corporation for the latest available information.